

FINAL REPORT

on

ENGINEERING MODEL OF ANTENNA CAPACITANCE PROBE

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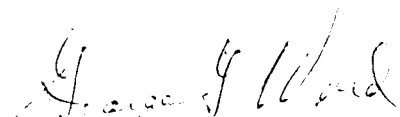
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Prepared for:
National Aeronautics and Space Administration
Goddard Space Flight Center
Greenbelt, Maryland


Washington Technological Associates, Inc.
979 Rollins Avenue
Rockville, Maryland

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Approved by:


Project Engineer
WTA


Chief, Electronic Engineer
WTA


Project Scientist
NASA/GSFC

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ABSTRACT

This report describes the development of the Engineering Model of the Antenna Capacitance Probe, part number 2158-R-434, which is the prototype of an instrument to be carried aboard the Radio Astronomy Explorer (RAE) Satellite in a forthcoming mission. The instrument is designed to measure the effective capacitance of the RAE antenna in two frequency ranges (750 kc and 1mc). Thus the capacitance not only can be determined directly, but also a function of frequency. The output to telemetry is binary digits rather than analog signals, hence the frequency as a function of capacitance has a high degree of accuracy.

The expected radiation dosage influenced the selection of the semiconductors and the circuit design was based on h_{FE} derating to insure operation should higher dosage be encountered. If further longevity is desired, certain transistor substitution is suggested based on subsequent information and investigation.

An overall system operation is followed by a detailed operation of the individual circuit explanation, including the appropriate waveforms.

A detailed reliability and failure analysis shows that the probability of complete success is .956 and the probability of complete failure is .044.

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I. INTRODUCTION

This report describes the Engineering Model of the Capacitance Probe produced under Contract NAS 5-3759. This Capacitance Probe, is the prototype of an instrument to be carried aboard the Radio Astronomy Explorer (RAE) Satellite in a forthcoming mission. The purpose of the probe is to measure the capacitance which occurs between each of two antenna booms connected as dipoles, and the spacecraft skin. The antenna capacitance will be measured at two frequencies widely separated from each other. The instrument is designed as the RAE Antenna Capacitance Probe part number 2158-R-434.

The basic principle of measurement of the Capacitance Probe consists in reflecting the capacitance which exists between the two antenna booms, and the spacecraft skin to an oscillator which will change its operating frequency in direct proportion to the capacitance seen by the antennas. Each antenna boom is fed by a coaxial cable, and in order for these booms to look like a dipole antenna which is balanced to ground from each boom, some sort of impedance matching device is necessary. For this reason balun transformers are used in the oscillator circuit, the secondary of which acts as part of the tuned circuit of the oscillator. Since capacitance is a function of frequency, two oscillators are used, so that capacitance as a function of frequency can also be determined.

A plot of capacitance versus oscillator frequency is provided for each oscillator. Prior to launch the capacitance of the antenna base, coaxial cable,

and the oscillator capacitance can be determined. Thus by subtracting out this total predetermined capacitance from the capacitance as measured by the oscillator the antenna capacitance can be determined.

Each oscillator is completely enclosed in a metal container which is temperature monitored. In addition, a calibrating capacitor is provided so that oscillator drift can be detected if it occurs and the readings compensated for. The overall system then produces frequencies which are directly related to the low frequency calibrate capacitance, high frequency capacitance, the antenna low frequency, and the antenna high frequency. These frequencies are then telemetered to ground so that a record is obtained of the antenna capacitance during flight.

The power to the Capacitance Probe will be turned off and on during the flight. This will reduce the power consumption and eliminate interference problems between other experiments.

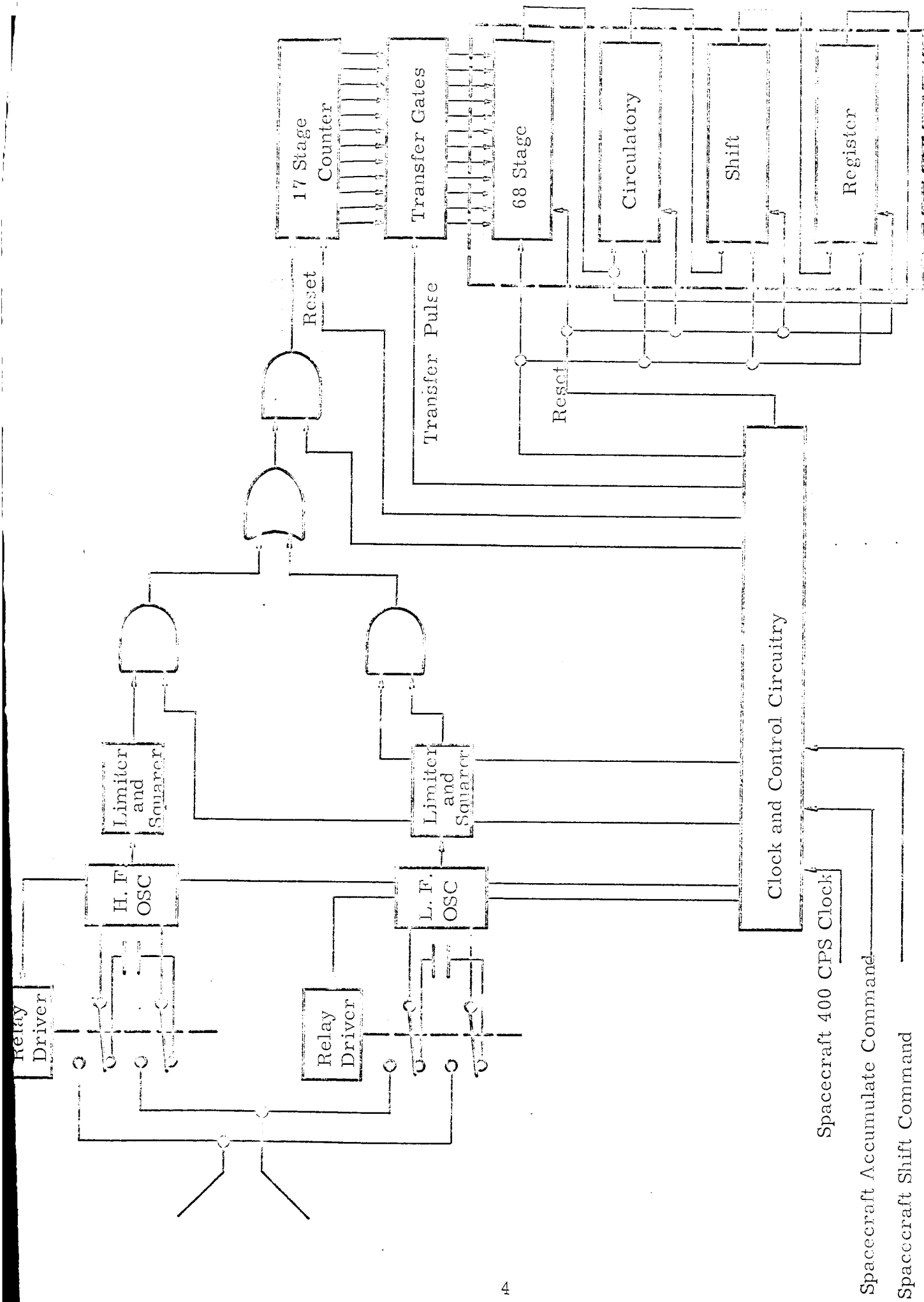
II. SYSTEM OPERATION

The general principle of operation of the Antenna Capacitance Probe has already been outlined in Section I above. This will now be extended by a more detailed functional description of the instrument, which will be followed by a discussion of the theory of operation of the system. Detailed description of some of the circuits will be presented in section IV.

2.1 GENERAL SYSTEM OPERATION

A block diagram of the Antenna Capacitance Probe is shown in Figure 1. The experiment is run synchronously from the 400 cps spacecraft clock. Upon receiving the power to the experiment, a minimum of 2.5 seconds is required for oscillator stability. At some time thereafter the programmer can initiate the accumulate or read command, which arrives as a pulse shifting from +6 volts (off) to 0 volts (on). The Capacitance probe, upon receipt of this command clears the counter, shift register and control circuitry by using this accumulate pulse as a reset.

Once the experiment is reset the oscillator frequencies are gated to the counter. The sequence of the frequencies going to the counter are: low frequency calibrate; 750 kcs; high frequency calibrate, 1 mcps; antenna low frequency; antenna high frequency. This sequence is internally controlled from the control logic circuitry by use of OR and NAND gates. The output of the oscillators are shaped for counter triggering by the squaring circuitry.



68 STAGE

Figure 1. 1-Block Diagram Antenna Capacitance Probe

After each input to the counter which lasts exactly for 100 milliseconds, the counter input is temporarily disabled while the parallel transfer of information from the counter to the shift register is made and the counter is reset. The counter then accepts the next frequency. During this count time the information in the shift register is serially shifted down 17 stages to make room for the following parallel transfer to the shift register.

After four inputs to the counter, which requires 480 milliseconds, the shift register is full and the control circuitry will allow it to accept no more information, by disabling the transfer and shift pulse lines.

The information now contained in the shift register in binary digits is ready to be sent to telemetry upon receipt of the Shift Command. Some convenient period of time after the 480 milliseconds required to fill the register, the shift command will occur in the form of a gating pulse to a NAND gate, shifting from a off-state potential of +6 V to a on potential of zero volts and lasting for at least some integral multiple of 168.75 milliseconds. This is to allow 68 pulses from the 400 cps spacecraft clock pulse, which is the other input of the NAND gate, to shift the information from the register out to telemetry. Since this is a non-destruct circulatory type of shift register, at the end of the 68th pulse the information will have been completely shifted out, but also will be repositioned in the register identically to its position before the arrival of the shift command.

The information to telemetry is in the form of binary digits rather than analog signals, thus enabling each bit of frequency to be read. However some bit may be dropped or lost due to noise in transmission and for this reason it may be desirable to shift the information out twice, since no parity code has been provided.

Upon the completion of the shift pulse, when all the information stored in the register has been sent to telemetry the programmer may shut off the power to the Capacitance Probe, and the antennas can now be time shared with another experiment.

III. RADIATION EFFECTS ANALYSIS

3.1 INTRODUCTION

The requirement for radiation hardening is due to the fact that the RAE Satellite is to be flown in an orbit lying in the Van Allen belt. This is a circular orbit at an altitude of 6000 km inclined 50° to the equator. Because the intended service life of the satellite is at least one year, this means that the instruments will be in a radiation environment for 10,000 hours or more. Electronic components, such as resistors, capacitors, coils and insulators having been tested repeatedly in various radiation environments, have shown that they are orders of magnitudes less sensitive to radiation than most semiconductors. Therefore, it is reasonable to conclude that the maximum radiation tolerant level, that is the threshold failure, will be primarily determined by the semiconductors. In addition, the circuit design based on reduced gain will have a direct bearing on circuit failure due to radiation dosage. Since semiconductor devices are the most sensitive components to space radiation, this analysis is limited to the semiconductor devices of the Capacitor Probe.

3.2 TYPE OF NUCLEAR RADIATION EXPECTED

A combination of nuclear radiation environment may be encountered by a space vehicle passing through the Van Allen belts. Some of the types of radiation are, the steady state trapped radiation, cosmic radiation, auroral radiation and solar flare radiation. The greatest electronic failure damage to space electronic system in earth orbit is due to trapped particles which are the Van Allen electrons

and protons plus the Starfish explosion. Although the Starfish electron levels at the higher altitudes have not decayed as quickly, the electron levels in the lower portion of the electron belt have decayed by approximately an order of magnitude.

Cosmic radiation consist mainly of hydrogen nuclear (protons) having very high speed and whose energy is in the range of 1×10^9 to 1×10^{10} electron volts. It is impractical to shield against cosmic particles of such great penetration.

Since the cosmic-particle flux is roughly about two particles per square centimeter per second. The ionization dose rate which can be attributed to cosmic radiation is in the vicinity of 10^{-4} rads per hour and that attributable to secondarily produced particles is about 10^{-3} rads per hour. Most materials and electronic components are uneffected until an ionization dose of 10^{-4} rads has been absorbed. Consequently cosmic radiation does not pose a severe threat to the performance of electronic equipment, especially when the materials and packaging are such as to reduce the secondary radiation or ionization effects. That is, complete potting of package to reduce any air space, and the restriction of materials such as teflon, Kel-F and the like. Although certain amounts of radiation effects can be annealed out by subjecting the electronic components to high temperatures, the temperature environment of the spacecraft is for too low for any efficient annealing to take place.

The innear Van Allen belt beginning somewhere between 399 to 1199 kilometers depending upon the latitude, and extending from about 45° north magnetic latitude

to 45° south magnetic latitude is made up mainly of a large number of electrons and protons of various kinetic energies all trapped by the earth's magnetic field.

Since radiation energy may be supplied in two forms: electromagnetic radiation, such as X- and gamma rays; or as particle radiation. The particles comprising the second basic form of radiation energy can be positively charged protons, negatively charged electrons or uncharged neutrons which are moving at speeds slower than that of light.

For charged particles, the interaction between the particle and an atom of the struck material produces a displacement which is caused by the electrostatic force between the nuclear charge and the moving particle. This displacement will occur if the charge particle has sufficient energy and if it passes within the radius of the inner most electron orbit. Uncharged particles such as neutrons, however, must pass within the much smaller radius of the atomic nucleus. If electrons have sufficient kinetic energy that is energies greater than 200 kev, they will create displacement. Since electrons of energies greater than 200 kev comprise less than 10% of the total in the Van Allen belts, displacements by electrons are not a major concern. However, whether the electrons originate in the Van Allen belt or as secondary electrons produced by gamma radiation they create the same effect.

Secondary electrons produced by gamma radiation processes can attain energies in excess of 200 kev. For primary electrons with energies ranging from 40 to 50 million electron volts, the energy is expended in the form of braking radiation, which is a complete conversion from the electron into gamma radiation. The threat from electrons in space radiation environment is due more from

contributions to the total ionization dose than from displacement damages.

Although the average energy of electrons in the belt allows shielding, the secondary radiations they create is more troublesome. Protons are the most penetrating type of charged particle radiation in the Van Allen belts because of their higher average energy and their larger mass. Shielding against protons, therefore is difficult.

3.3 RADIATION ESTIMATES

Due to the spacecrafts orbit it has been estimated that the radiation dosage will be: 2.2×10^{12} electrons per cm^2 per year having kinetic energies greater than .5 mev; and 10^4 protons per cm^2 per second having kinetic energies greater than 30 mev. Further, it has been estimated that the spacecraft will provide a shielding density of .63 grams per cm^2 and the package will provide a shielding of .4 grams per cm^2 . If it is assumed that this shielding will be done with aluminum then all the constant troublesome radiation effects will be due totally to protons having kinetic energies greater than 20 mev, as the shielding will be sufficient to effectively shield against all but the most energetic electrons (3 mev)³.

Definite conclusions and expectations concerning nuclear radiation damage are not possible at this time, due partly to insufficient investigations, results and methods which are not totally compatible. Although it is well known that the radiation resistance of semiconductors can vary between batch lots of the same type for any given manufacturer it has also been generally agreed that some methods

of manufacturers are better than others for reducing failure due to nuclear radiation. Germanium it would seem is better in general than silicon, however, trade-offs exist when the circuit use is considered.

Although the transistors selected for use in the Capacitance Probe do not necessarily represent ones having the highest possible radiation resistant characteristics; they were chosen consistent with the current available information as to preferred type, previous performance, and radiation testing results. They

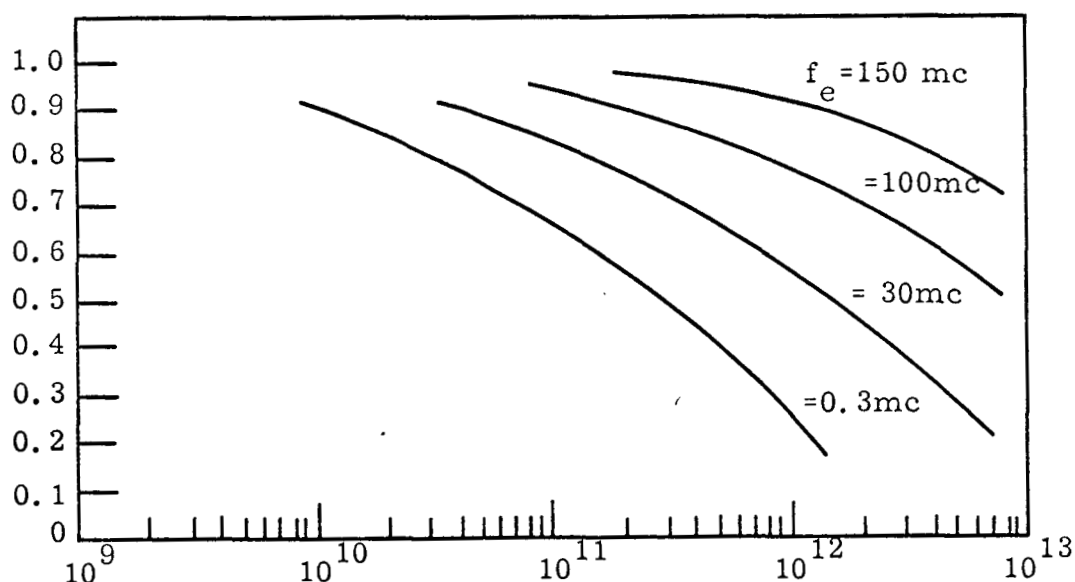


Figure 3.1-NEUTRON EXPOSURE (NVT, $E > 0.1$ MEV)

are of the high f_T and low breakdown voltage type. The low breakdown voltage is also a requirement in the design of other radiation tolerant semiconductor devices, namely diodes. Semiconductor radiation dosage in laboratory testing have, because of convenience, been exposed mainly to neutrons rather than electrons or protons, therefore in order to reduce all data to common terms, the proton and electron rates can be reduced to equivalent neutron dose rates. For protons the average figure which can be used for a close equivalence is to multiply the proton

rate by 4.2; for electrons, multiply the rate by 0.01 in order to get the equivalent neutron rate.

The susceptibility of transistors to radiation is a function of the transit time across the base region and, thus is a function of the transistor alpha cutoff frequency.⁵ Degradation in transistor alpha as a function of neutron flux for various alpha cutoff transistors is shown in Figure 3.1. By converting the expected dosage of electrons and protons to equivalent total dosage of neutrons, the expected dosage would be 11.2×10^4 neutrons per square cm per second.

The alpha cutoff frequency of the 2N930 is 30 Mc. From Figure 3.1 a total dose of 41×10^{11} n/cm² would be required for the alpha to decrease to a value 0.7 of its initial value, which means that if no derating is done in the circuit design, the 2N930 could remain in orbit 1.16 years before its alpha degrades to 0.7 of its initial value. Actually in every discrete circuit of the Capacitance Probe the circuits have been designed to work with the common emitter gain lower than 0.5. The alpha cutoff frequency of the 2N2604 and 2N915 are 100 Mc and 400 Mc respectively and therefore have greater lifetimes than the 2N930.

An effective increase in the lifetime can be obtained by the substitution of 2N918 transistor for the 2N929 and 2N930, and substituting 2N3251 for the 2N871. This would mean that the lowest alpha would be 200 Mc; that is, the life would be extended to better than 8.3 years before the lowest alpha cutoff transistor would be degraded to 0.7 of its original value.

Similarly, the selection of the integrated circuits used in the Capacitance Probe are those whose past performance in space and radiation test results indicate that they would survive the expected exposure in the Van Allen belt. There have not been many studies performed in integrated circuits since they are relatively new. However, those which have been performed conclude that the integrated circuits will be as radiation resistant as the same circuit makeup with compatible discrete components. Those Texas Instrument (series 51) circuits which were tested⁶ showed that first failures occurred at about 1.8×10^{14} n/cm² which is an order of magnitude above that discussed above. One possible means of improving the radiation resistance of these units, which could easily be incorporated in later units if deemed necessary, would be to decrease fan-out loading, since it has been found that there is a direct correlation between fan-out and failure dosage.⁴

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IV. CIRCUIT DESCRIPTION

The individual part of the RAE Antenna Capacitance Probe will be discussed in some detail in this section of the report. The progression to be followed is:

- 4. 1) Counter, transfer gates and shift register
- 4. 2) Purpose of control logic
- 4. 3) Functional operation of control gates
- 4. 4) Control gates input and output timing signals
 - 4. 4. 1 Oscillator gating
 - 4. 4. 2 Transfer pulse
 - 4. 4. 3 Relay Driver pulse
 - 4. 4. 4 Reset pulse
 - 4. 4. 5 Shift pulse
- 4. 5) Development of control signals
- 4. 6) Relay Driver circuitry
- 4. 7) Oscillator circuitry
- 4. 8) Limiting and squaring circuitry
- 4. 9) Output buffer and send shifter

4. 1 COUNTER, TRANSFER GATES AND SHIFT REGISTER

The frequency counting and storage circuitry, comprising the counter parallel transfer gates and circulatory shift register is shown in the overall schematic. The counter is made up of 17 conventional ripple through divide by two- binary frequency divider stages; stage 17 of which registers the most

significant bit. In using 17 binary stages to count 2 megacycle by accurately counting for 100 milliseconds the counter has the capability of counting a decimal number as high as 131,071. Since the counting time is .1 second or 100 milliseconds the conversion factor between the actual frequency of the oscillator and the number in the counter is 10, therefore the highest frequency capability of the counter is 1,310 710 cps. Although the calibration frequency of the high frequency oscillator is one megacycle or 100,000 as read by the counter, it is not anticipated that the capability of the counter will be exceeded when the antenna is substituted for the calibrating capacitor. If however this should occur, the counter would begin counting over at a frequency of 131,172; hence the maximum frequency would be added to the frequency read out of the counter.

The last 15 stages of the counter are Texas Instrument SN5112 integrated circuit Flip-Flops, which are rated at a maximum of one megacycle. To insure the reliability of this counter, two high speed flip-flop, Signetics SE124G's were used for the first 2 stages, whereby the SN5112 will now see a maximum of 250 kcs. By using different I. C. components for the first two stages of the counter however, required that the logic levels be compatible, hence the inverter between the 2nd and 3rd stages. This compatibility requirement also caused an additional circuitry to be added to the reset line driving only the signetic stages. While all 17 stages of the counter reset from the same reset pulse, the last 15 T. I. stages only require one inverter, however, the first two high frequency Signetic stages require 2 inverters in the reset line for logic compatibility.

The transfer gates are Texas Instrument SN516A operated as Nand Gates. If the side of the counter stage driving a transfer gate is a "one" when the transfer pulse occurs, then this "one" is transferred to the appropriate shift register. The transfer of information from the counter to the shift register is done in parallel with the occurrence of the transfer pulse, however serial transfer is performed once the information is in the register with the occurrence of shift pulses.

The first 17 stages of the shift register are T. I.'s SN5111, the 51 following are SN510's, the reason for the difference being that the first 17 stages accept the information from the counter, therefore must have the capability of set, reset, and shift pulse, while the 51 following stages only need the capabilities of reset and shift pulse.

The register being a uni-directional circulating type of shift register necessitates that the output of the 68th stage be connected to the first stage which is shown in the overall schematic diagram. Thus shifting one bit at a time, requires 68 shift pulses for all information to be fed to telemetry. This shift rate occurs at a 400 cps rate which would thus require 170 milliseconds for the total transfer of information to be shifted out, and the information to be placed in its initial position within the register.

All stages of the register are essentially of the RS type and the interconnections between stages are identical. Since a shift register requires temporary storage between two adjacent flip-flops. The state of one flip-

flop being changed is stored and used as the input to the subsequent flip-flop stage. That is the steering diodes or circuitry of the n th stage are tied to the n th-1 stage and the output of the n th stage drives the n th+1 stages steering circuitry.

4.2 CONTROL LOGIC

The purpose of the control logic circuitry is to generate timing pulses at the correct time and of the correct length which will:

- (a) control which oscillator is on
- (b) control the mode of the oscillator (antenna or calibrate)
- (c) control the length of time an oscillator is on
- (d) shift the register
- (e) transfer the information in the counter to the register
- (f) reset the counter
- (g) control the counting time of the counter
- (h) prevent the register from over flowing before shift out command
- (i) provide shift out upon shift command

The only input to the control logic circuitry are the 400 cps clock pulse, the read or accumulate command and the shift command. Hence the proper control pulses are generated internally using these three input signals and Texas Instrument and Signetics integrated circuits; specifically Flip-Flops, Nand Gates, OR Gates, oneshot multivibrators and "clock" drivers.

4.3 FUNCTIONAL OPERATION OF CONTROL GATES

In the interest simplicity and clarity a step-by-step operational description will be presented, followed by the timing signals generated internally and applied to the control gates and interface circuitry between the gates and counter-memory circuitry. Subsequently the generation of the timing signals will be discussed.

Assume that the information contained in the register has been sent to telemetry and that the power has been shut-off, then at some time t_n , the power is reapplied. A minimum delay of 2.5 seconds ($t_n + 2.5 \text{ sec.}$) is required for oscillator stability before the Read Command can be applied; hence the read command occurs at $t_n + 2.5 + t$ acting as a reset pulse. The reset or Read command pulse resets the counter, logic circuitry and clears the shift register.

Immediately after the reset, the low frequency oscillator in the calibrate mode is allowed through the NAND gate A HG1/2 (terminal 2 of gate A HG1) and through the OR gate A HG1/2, 9, 10 (terminals 2, 9 and 10 connected together form one output); the low frequency calibrate is prevented from reaching the counter for 20 milliseconds however, by Nand gate A HG2/9, 2.

During this 20 millisecond delay the following things occur: (2) 5 milliseconds after the reset or read command, the transfer pulse is applied to the transfer gates, (b) 10 milliseconds after reset, the counter is reset.

At the completion of the 20 millisecond delay the Nand gate A HG2/92 is opened for exactly 100 milliseconds and the low frequency calibrate signal is allowed to pass to the counter. During this 100 milliseconds interval while the counter accepts and counts the low frequency calibrate signal, there are 17 shift pulses applied to the shift register, at a rate of 200 cps. The shift pulses occur 27.5 milliseconds after the counter begins counting and since it takes 82.5 milliseconds for the 17 pulses to be applied, the shift pulses last until the counter is gated off.

At the same instant the counter is gated off the low frequency calibrate signal is gated off and the high frequency calibrate oscillator signal is allowed to pass through the Nand gate A HG1/10 and the OR gate A HG1/9, 2. 10 but is delayed in passing to the counter again for 20 milliseconds while the transfer pulse and counter reset pulse is applied as previously explained. At the end of the 20 millisecond delay the Nand gate A HG1/92 is again opened for exactly 100 milliseconds and the counter accepts this frequency. Again while this is happening the information now contained in the first 17 stages of the shift register (low frequency calibrate) is shifted to the second 17 stages of the shift register by the 17 shift pulses passing through the Nand Gate C-G3/8, and OR gate C-G3/8, as previously explained.

At the completion of the second 100 millisecond count the high frequency calibrate is prevented from reaching the counter by the Nand gates, and the following things occur: the Nand gate B-G7/8 applies a positive pulse which

energizes the relay driver for the low frequency oscillator; the Low frequency Nand gate A HG1/2 is opened. Thus the frequency as seen from the antenna capacitance is fed to the Nand gate A HG2/92. Again this gate is prevented from opening for the 20 millisecond period while the high frequency calibrate signal, still contained in the counter, is dumped into the first 17 shift register stages with the transfer pulse which occurs 5 milliseconds after the start of the 20 millisecond delay, and the counter is reset again at 10 milliseconds in this 20 millisecond period. Following this 20 millisecond delay the counter accepts the antenna determining low frequency signal while the previous 2 signals are shifted down the register 17 stages.

Upon completion of the 3rd count the signal from the low frequency oscillator is disabled from reaching the counter and the low frequency relay driver is disabled.

However, at the instant the low frequency relay driver is de-energized, the high frequency relay driver is energized by a positive pulse from the output of Nand gate B-G6/8 and the read and shift cycle is completed identically to that of the low frequency cycle above. Thus at the completion of the 4th 100 millisecond count period there is information in the counter and in shift register stage 18 through 68.

If there was nothing to shut off this total read-in cycle the whole procedure would repeat at this point- in fact it starts too. However, during the 5th 20 millisecond delay period, not only is the information in the counter put in the

first 17 stages of the shift register, filling it, and the counter cleared, but at the completion of the 20 millisecond delay, the state of flip flop B-FF14 changes so that the Nand gate B-G7/4 will pulse flip flop C-G2/8 and B-G4/9 is disabled until another read command occurs resetting B-FF11. While the counter will continue to count there will be no further transfer pulses occurring to dump the counter information into the shift register; the only shift pulses which can now occur are generated by the shift command which would apply 68 pulses from the spacecraft 400 cps clock through Nand gate C-G3/4 and OR gate C-G3/8 to shift the information contained in the shift register to telemetry at the same time re-circulating the information within the register.

4.4 CONTROL GATES INPUT AND OUTPUT TIMING SIGNALS

The inputs and outputs to the above gates and the counter memory interface circuitry will be shown consistent with the assumptions made previously. It is important to note at the outset that all T. I.'s Nand gates use logic levels which are complimentary to those of the Signetic Nand gates; failure to do so will result in improper synchronization of the relay drivers.

4.4.1 Oscillator Gating

As shown in Figure 4.1A, the gating signal to the Nand gates controlling output of the oscillator shapes are developed from opposite sides of a flip-flop; hence when one gate is "open" the other is shut. Further gating is provided by another Nand gate controlled by a flip flop whose normal mode

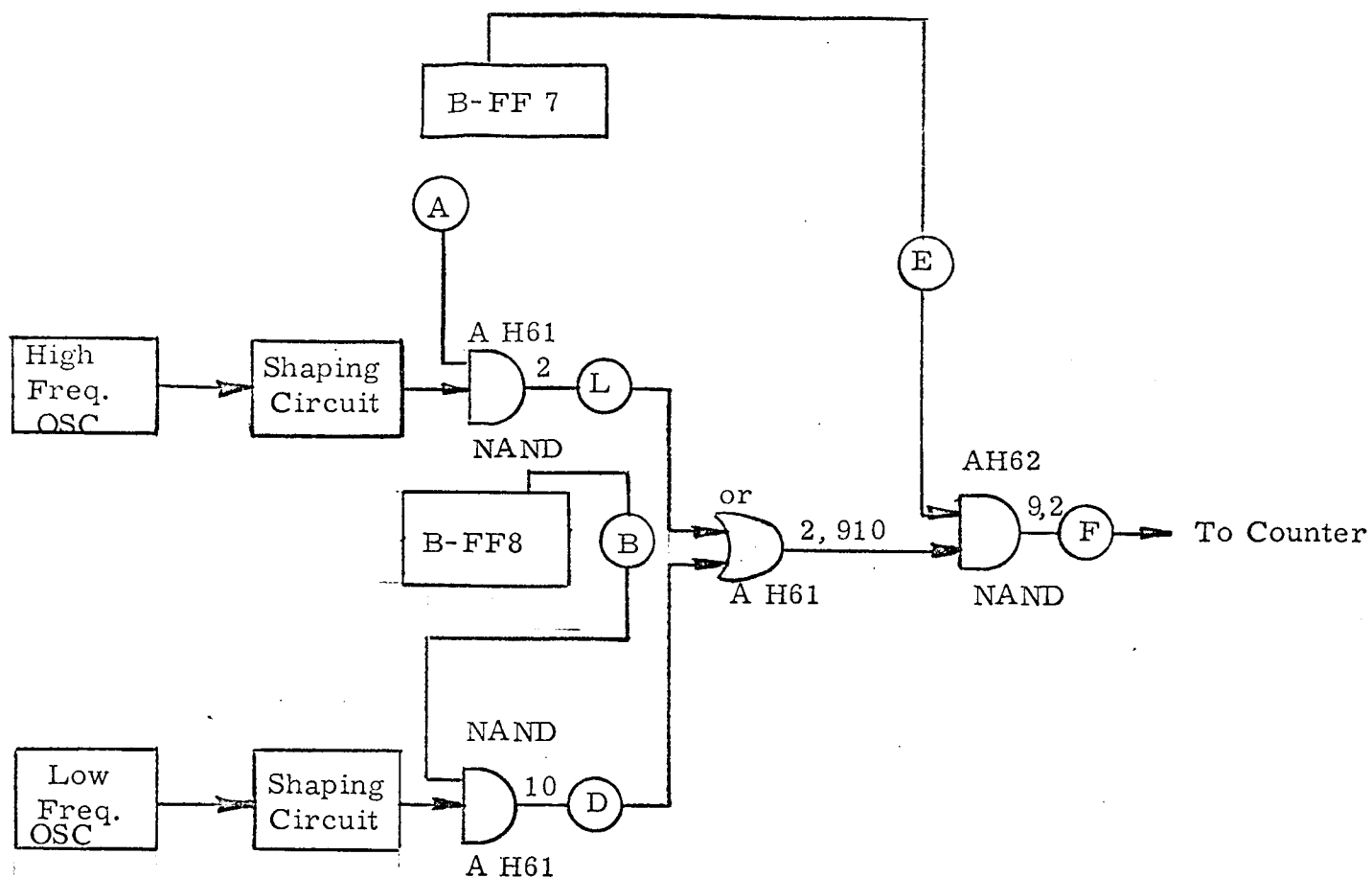


Figure 4.1 A Count Control Circuit

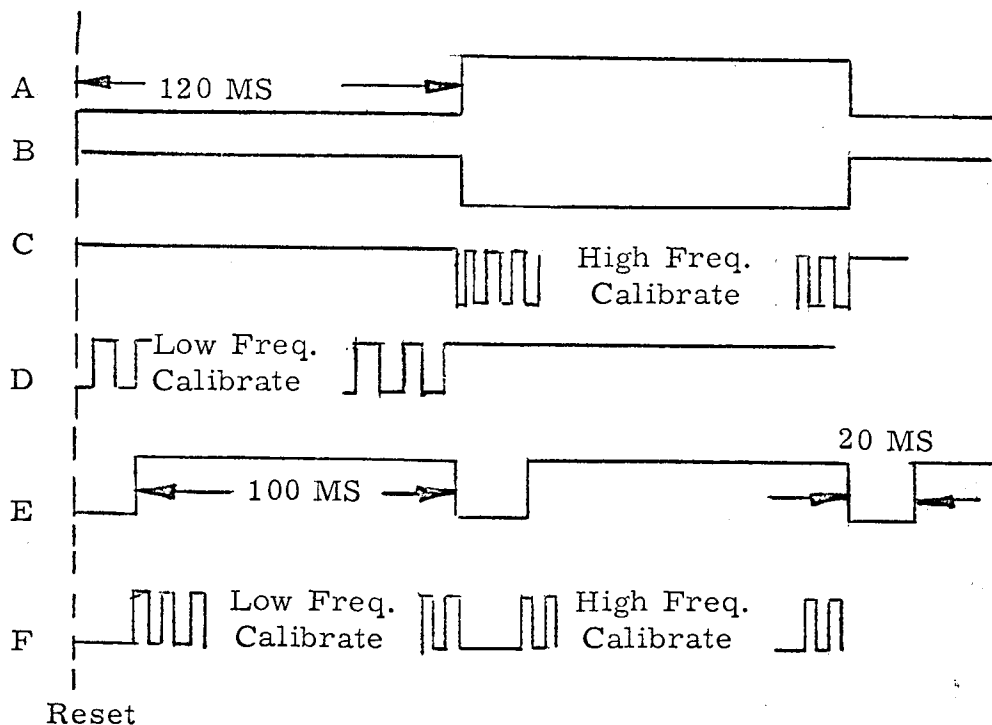


Figure 4.1 B Input & Output to Count Control Circuitry

is disrupted giving the output (signal E) as shown in figure 4.1B. It is this gate and its control which provides the 20 millisecond delay in which the transfer and reset pulses are generated. The output of the last Nand gate goes directly to the counter with no interface circuitry.

4.4.2 Transfer Pulse

5 milliseconds after the start of the 20 millisecond delay of oscillator information to the counter, the transfer pulse is generated. The associated circuitry for the generator is shown in figure 4.2A. At this point it should be noted that signal E of figure 4.1B is signal A in this presentation, being developed from C-FF7. The flip flop counter circuitry B-FF12, 13 and 14 is being driven from this signal, and in effect count the number of times this signal appears at the input. As long as signals D and F are down (low), Nand gate B-G4/9 will pass the pulse generated by signal E. Signal E is passed through one shot multivibrator having a pulse duration of approximately 11 usec. Before going to the transfer gate it is passed through an inverter for logic compatibility. Like C-FF7, the flip flop C-FF10 is a binary frequency divider which has had its regular output modified; the method of accomplishing this will be shown later.

4.4.3 Relay Driver Pulse

The method of generating the relay driver pulses are also shown in figure 4.2A. The 3 flip flop binary counter which drives Nand gate B-G4/8 also drives 2 additional Nand gates. The output of these gates go directly to the

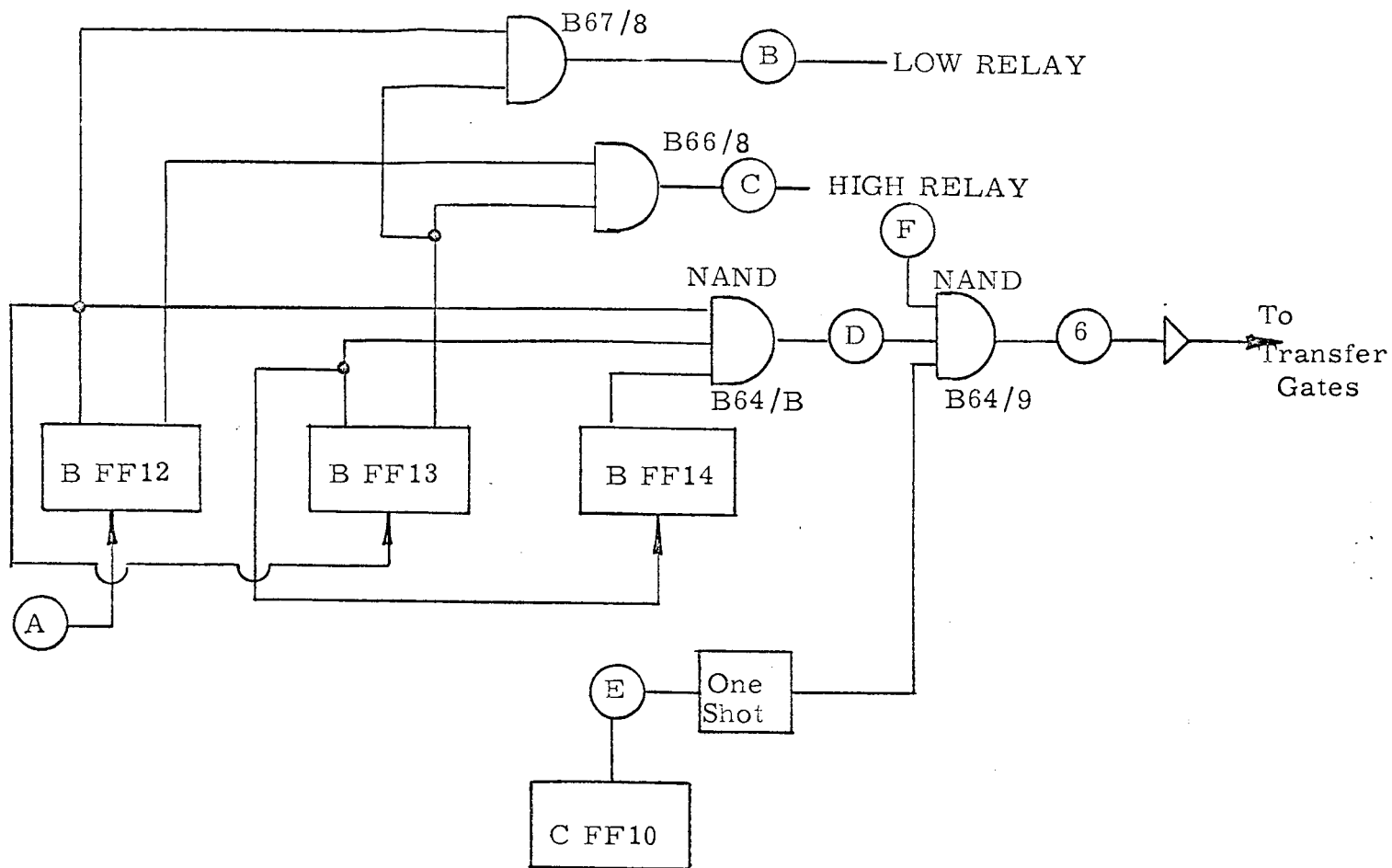


Figure A. Relay Driver and Transfer Pulse Circuitry

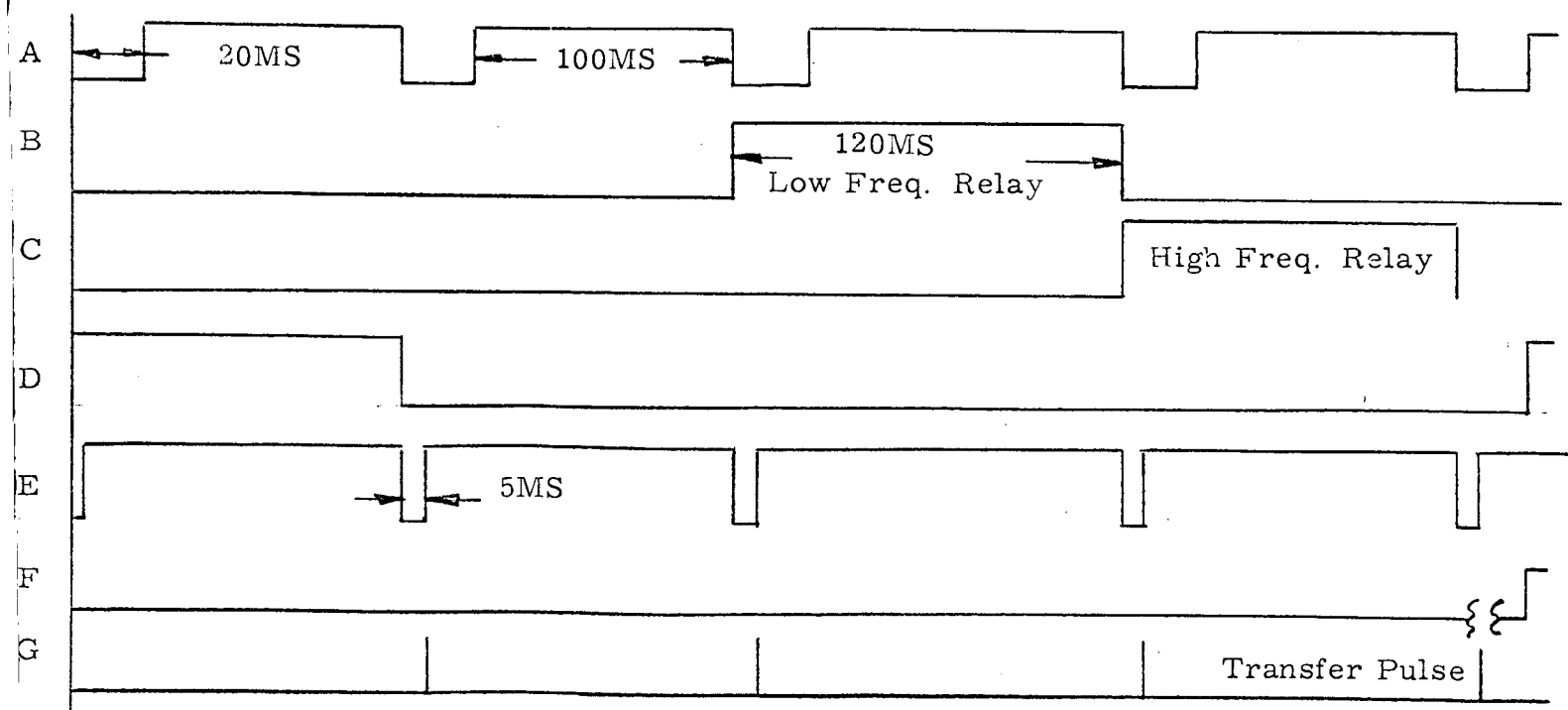


Figure 4.2B. Timing Pulses for Relay Drivers and Transfer Pulses

inputs of the relay driver. Hence when the carry side of BFF12 and BFF13 are high there is an output from Nand-gate B G6/8 which will energize the high frequency relay driver. The output sequency of these gates is shown in figure 4.2B.

4.4.4 Reset Pulse

As mentioned previously the read command is really a reset pulse, reseting the entire instrument. The counter on the other hand can be reset periodically from internal circuitry as shown in figure 4.3A. Because the reset signal to the shift register is a DC level rather than a pulse the R-C integrating interface circuitry shown is to present stray pulses from perhaps accidently reseting a portion or stage of the shift register.

The binary frequency divider flip flop C-FF9 has had its normal mode interrupted similar to CFF7 and CFF10; its output is shown in figure 4.3B. The output of this flip flop, sent through a non-inverting one-shot multivibrator with a pulse duration of approximately 11 usec., forms the counter reset pulse which occurs during the 20 millisecond delay of oscillator information to the counter. Signal C of figure 4.3B is not used in the generation of the counter reset pulse, but is included only to show the occurance of the reset pulse in reference to it.

The dual inverters driving the first two stages of the counter are only for logic compatibility for the high frequency stages as previously explained.

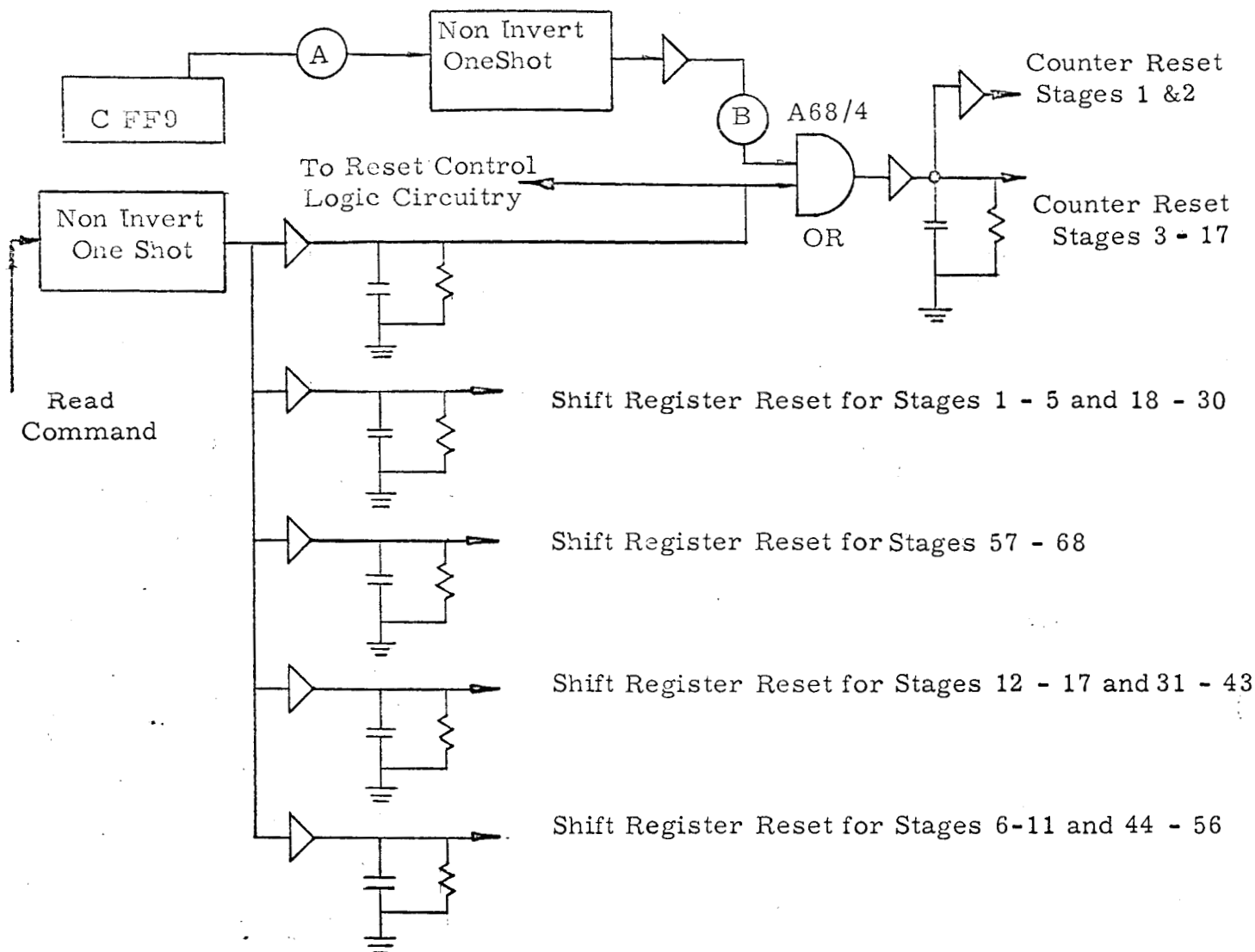


Figure A. Counter and Shift Register Reset Circuitry

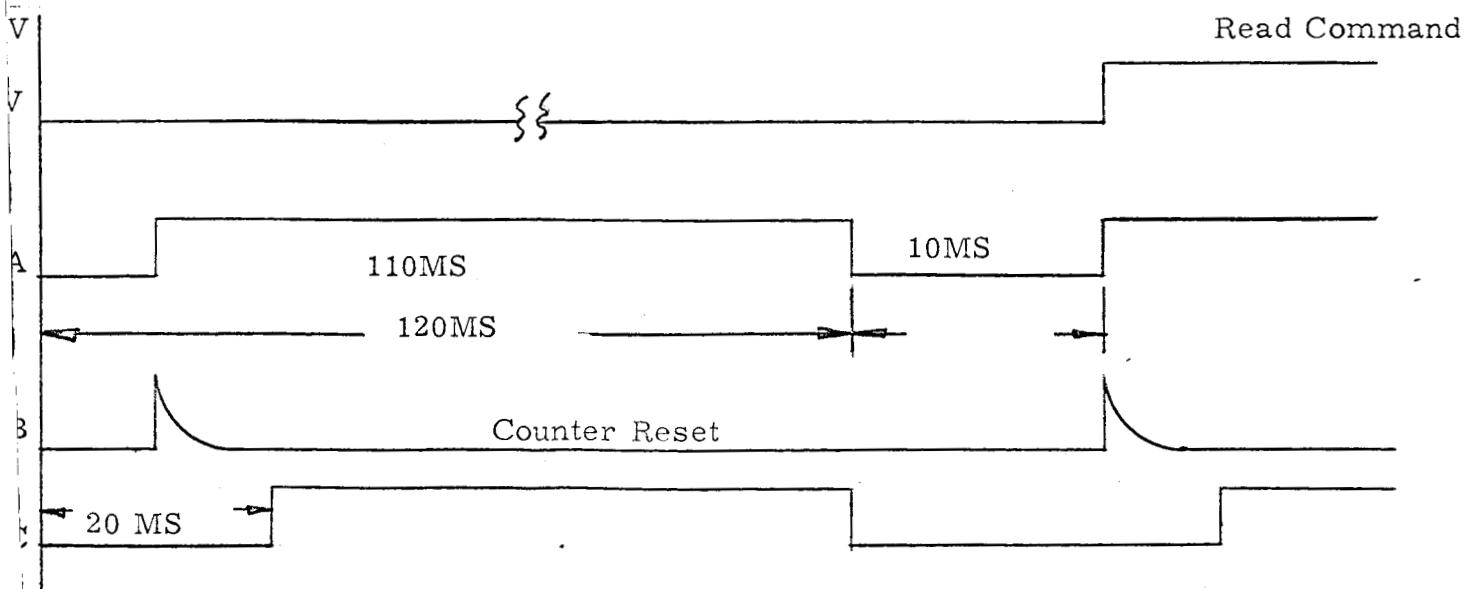


Figure 4. 3B. Read Command and Counter Reset Pulses

4.4.5 Shift Pulse Circuitry

The shift pulses to the register occur at two rates; 17 pulses at the rate of 200 per second after each transfer pulse lasting for 82.5 milliseconds, and 68 pulses occurring at the rate of 400 pulses per second gated in coincidence with the shift command, lasting for 168.75 milliseconds for read out to telemetry. The circuitry and signals generating these pulses are shown in figure 4.4A and B respectively.

The 68 shift pulses occur whenever the shift command opens Nand Gate C-G3/4 allowing the 400 cycle clock to be applied to the shift pulse lines. The 17 shift pulses are generated internally yet derived from the 400 cps clock pulse. As long as signals G and H are such that Nand gate C-G2/8 is open the shift pulses actually are the pulses coming from C-FF1, which is a binary frequency divider being driven from the 400 cps clock pulses; hence the output of this flip flop is at the rate of 200 cps.

The read command will reset B-FF11 such that signal H would open Nand gate C-G2/8, and this signal will not change until 4 sets of 17 shift pulses have been applied to the shift register. Actually B-FF14 will set BFF11 after counting 4 input signals from C-FF7.

This then leaves signal G, which is gated by signals D & E as the largest controlling factor of gate C-G2/8. Signal E is dependent upon the same as signal A. Signal D on the other hand is generated from a divide by three circuit which is "up" once for every 3 times signal C is up. The output

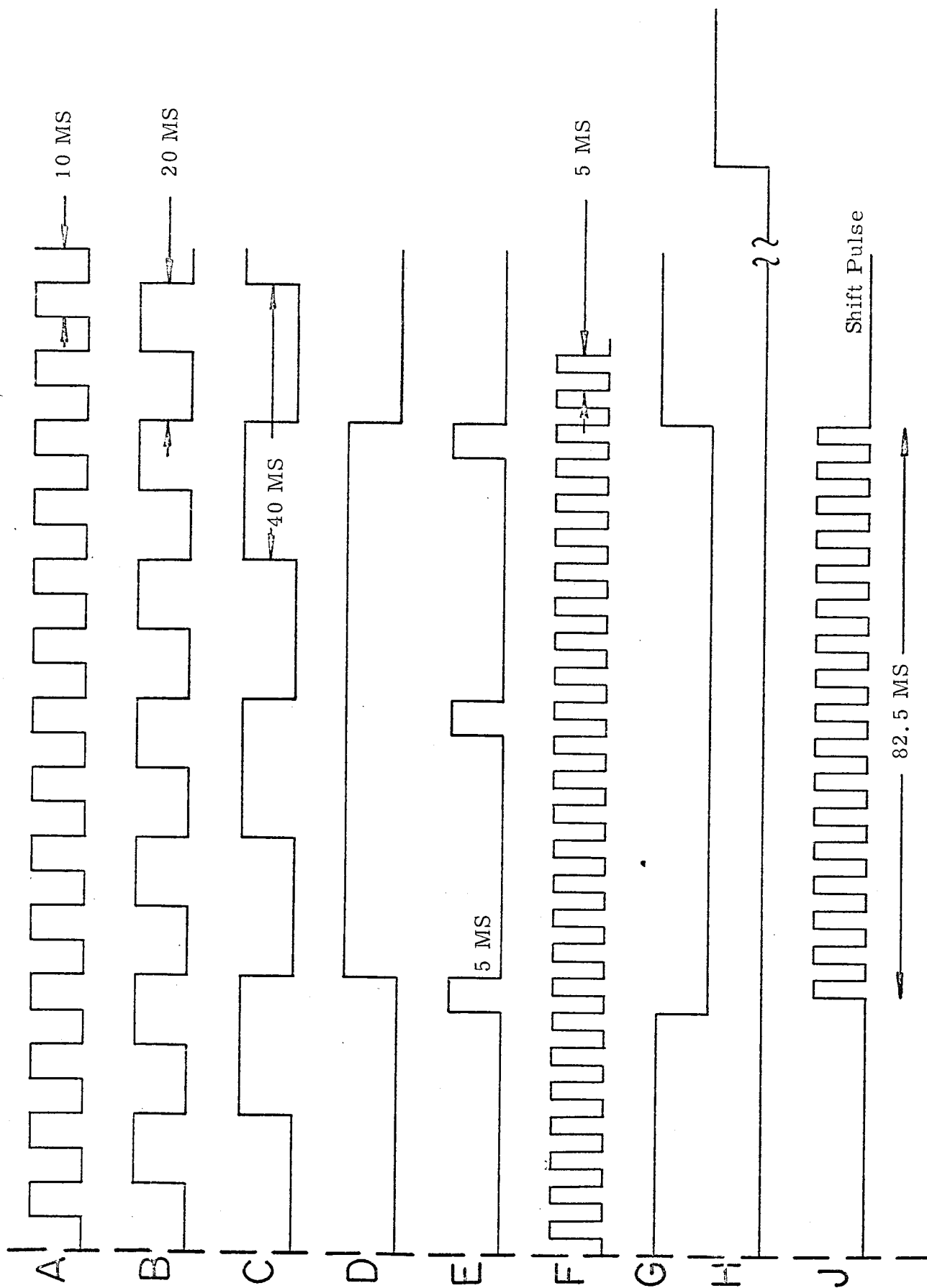


Figure 4. 4B Input And Output Signals To Control Shift Pulses

of Nand gate C-G1/8 will be up, thereby shutting off Nand gate C-G2/8, anytime signals D and E are both down. Whenever either or both signals D and E are up however, signal G will be down, and until signal H changes will open Nand gate C-G2/8 to pass the output of C-FF1.

The clock drivers in the shift pulse lines are present to insure sufficient drive capabilities to all 68 stages of the register, since each driver is only capable of driving 20 stages.

4.5 DEVELOPMENT OF CONTROL SIGNALS

The circuitry necessary to develop the proper control signals to be applied to the control gates is shown in figure 4.5 A, while the outputs of each stage is shown in figure 4.5B. All ten flip flops used in this circuitry are basically of the RST type. While each flip flop has the capability of its regular outputs, that is Q and \bar{Q} , they also have additional capability of those outputs from a low impedance emitter follower. In the interest of simplicity however only one side of the outputs have been shown in figure .

The first stage is driven by the spacecraft 400 cps clock pulse; signal labled A. All ten stages are reset by the arrival of the read command pulse to the reset input. Each of the first four stages act as a standard ripple through divide by two binary counter stage. The second stage while driving the T input of the third stage also applies its output to the set input of the tenth stage. Similarly the output of the 3rd stage drives the T input of the fourth stage and in addition provides an input to the set input of the ninth stage. The fifth and

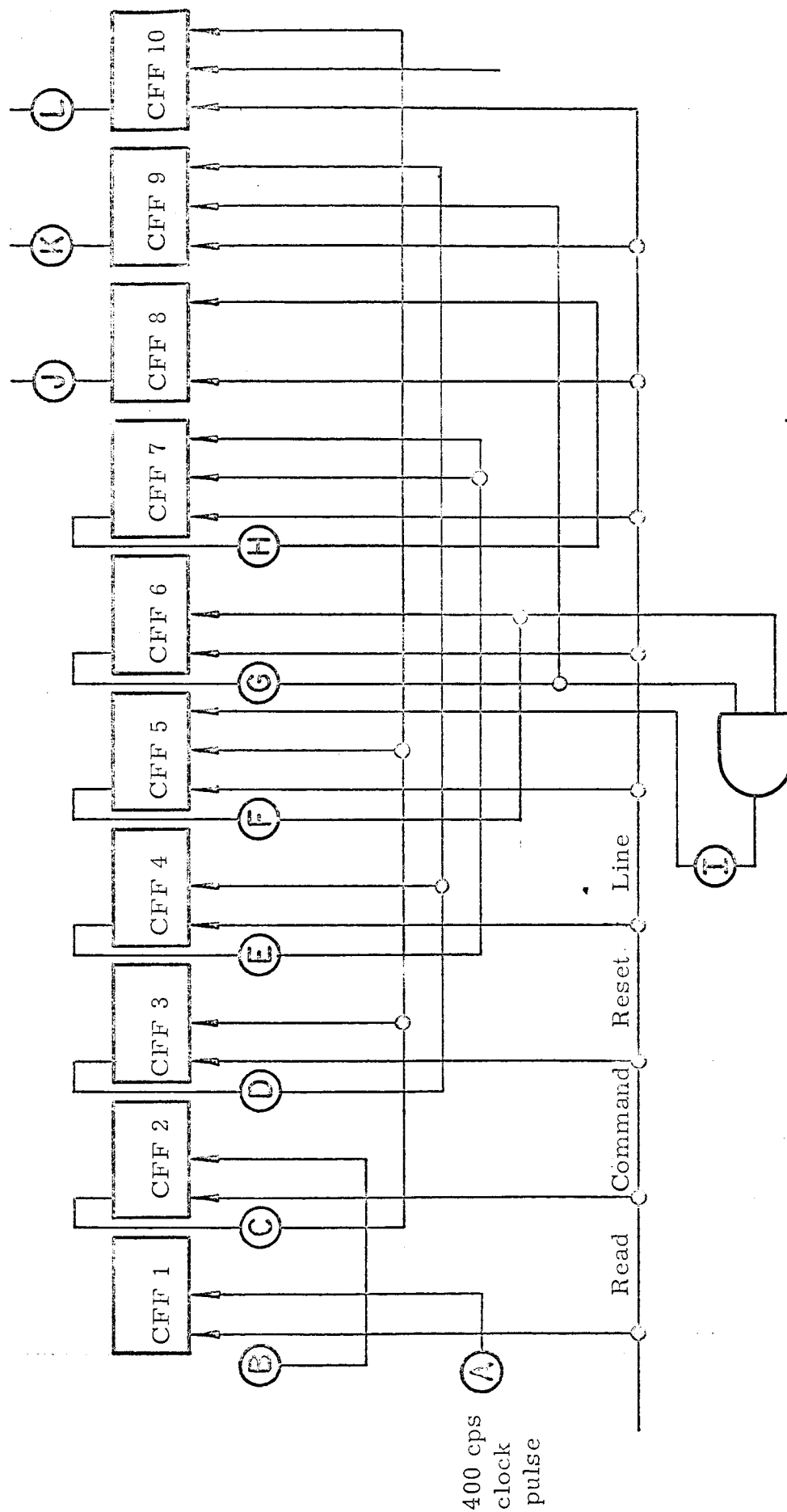


Figure 4.5A Control Signal Development Circuitry

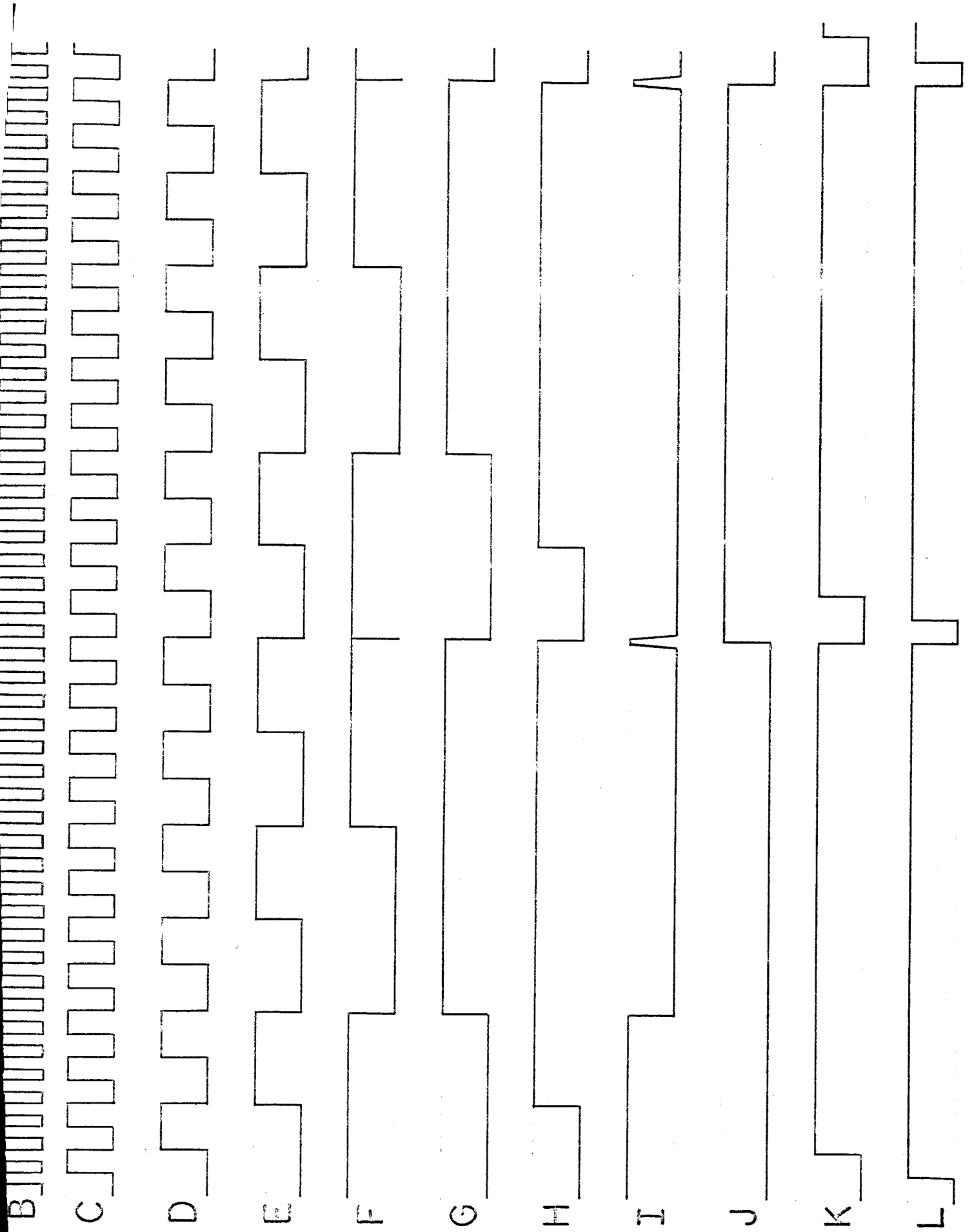


Figure 4.5B Signals Developed For Control Circuitry

sixth stages together with the associated Nand gate provide a divide-by-three circuit; the output of the sixth stage driving the T inputs of the seventh, ninth and tenth stages.

4.6 RELAY DRIVER OPERATION

Because the low frequency oscillator relay pulls in first, the relay driver will be described in terms of this driver, however both relay drivers are identical. The circuitry for these drives are shown in the overall schematic.

Before the arrival of the actuating pulse the base of Q_3 is at ground potential. Thus forward biasing this emitter follower; hence at full saturation approximately one-half volt is developed across it. Because of the low positive voltage at the junction of the 3.9K ohm and 1.5K ohm resistors the base of Q_4 is biased off at approximately -5 volts, hence the relay is de-energized.

With the arrival of a positive 3.5 volt pulse to the base of Q_3 the bias on the PNP transistor Q_3 is now such as to apply reverse bias to the base-to-emitter junction, thereby cutting off this stage. With the emitter follow cut-off the base potential of the NPN Q_4 becomes positive, putting this transistor into full saturation, and the appropriate relay will be energized thus switching the appropriate oscillator from the calibrate capacitor to the antenna and its capacitance.

The positive pulse (+3.5V) which is sent to the relay drivers will last for 120 millisecond and is developed from the control logic circuitry. As can be seen from Figure 4.2B during the first two 120 millisecond periods the

Capacitance Probe is on and neither relay is energized, then the low frequency is energized for the 3rd 120 millisecond intervals upon completion of which the low frequency relay is again de-energized and the high frequency relay driver is energized for 120 milliseconds during the 4th interval.

4.7 OSCILLATOR CIRCUIT

The oscillator circuit is shown in figure 4. 6A, while the AC equivalent circuit for this oscillator is shown in figure 4. 6B. Both the high and low frequency oscillator circuits are identical with the exception of some values of resistors, capacitors and the baluns; thus the description will be made in terms of the high frequency circuit. The calibration frequencies are one megacycle for the high frequency and 750 kc for the low frequency oscillator.

Resistors R33, R34, R35 and R36 determine the bias point and hence the quiescent operating point of transistor Q_7 . Similarly for the emitter follower resistors R37, R38, R39 and R40 determine the quiescent point for Q_8 . The oscillator output signal is developed across R36, the emitter AC load resistor and coupled through C18 to the emitter follower buffer stage.

The operation of the oscillator is more simply seen if the parallel capacitors are grouped into one equivalent capacitance as shown in figure 4. 6B. The signal developed across R35 is fed back to the junction of C_2 and C_3 sustaining oscillations at the frequency determined by L'_1 , C'_1 , C'_2 , C'_3 and the reflected capacitance from the primary of the transformer. Capacitors C_{16} and C_{17} offer

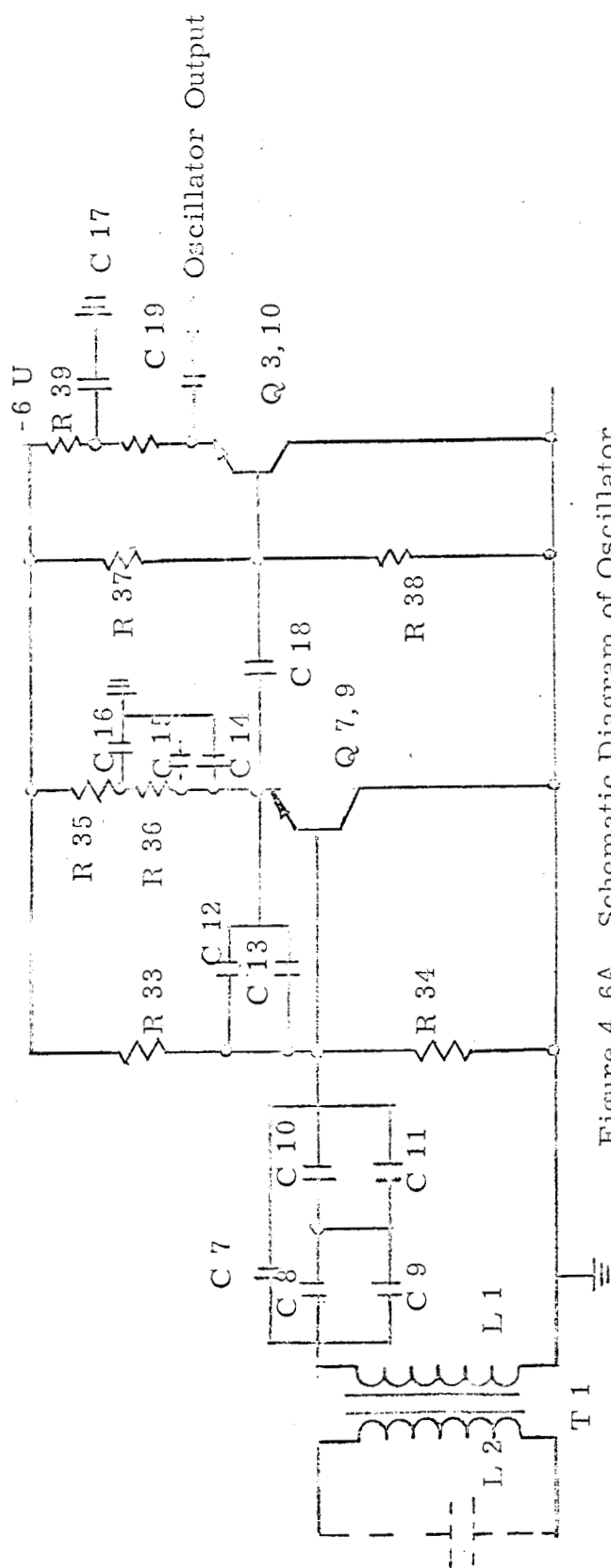


Figure 4.6A Schematic Diagram of Oscillator

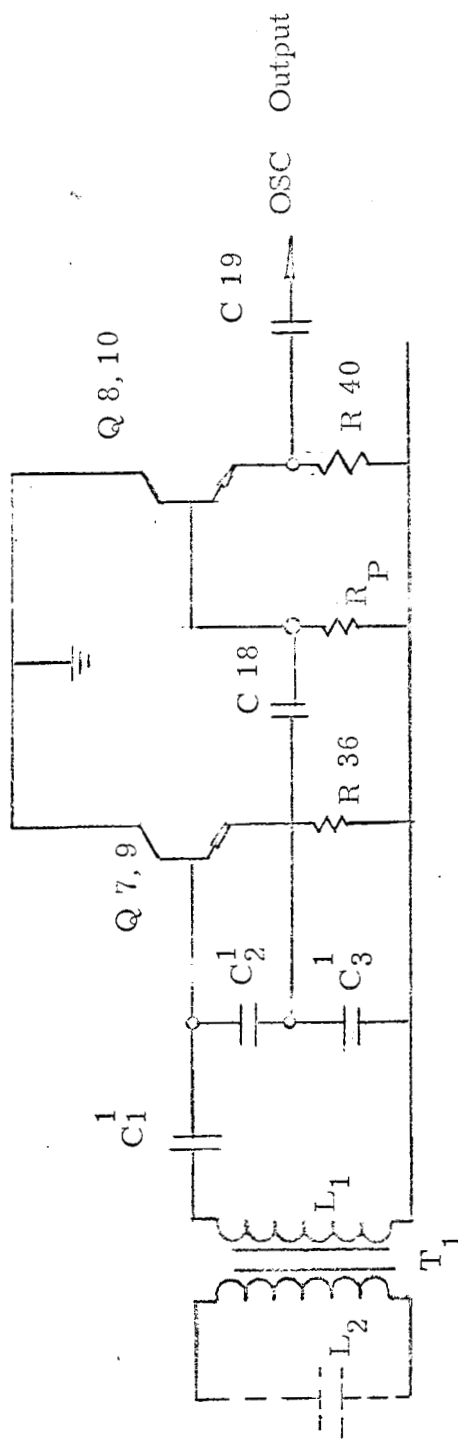


Figure 4.6B. A.C. Equivalent Circuit of Oscillator

a low impedance path for AC thereby effectively grounding R35 and R39 hence the AC load impedance is R36 and R40 for each stage.

4.8 LIMITING AND SQUARING CIRCUIT

The output of each oscillator is fed to separate squaring circuitry, both of which are identical; this circuitry is shown in figure 4.7A. The output of the oscillators are sinusoidal, however, to be fed to the counter the waveform must be squared, or formed into pulses which can be counted by the conventional ripple through binary counter.

The 10K ohm resistor at the input to the schmitt trigger serves as a isolation resistance and also forms part of the gain relationship for the Fairchild F-710C Amplifier. The ratio of the 200K ohm feedback resistor to the 10K ohm input resistance determines the gain of the amplifier, which is 20. The diodes at the input to the Fairchild amplifier will limit the sinewave input to approximately $\pm .6$ volts to prevent damage to the amplifier.

Since the input signal to the F-710 amplifier is fed to the inverting input, as the input signal goes positive, the output will go negative and because of the gain, the amplifier will go into saturation, giving a -3 volt output until the input signal again goes through zero whereupon the output is again saturated in the opposite direction to +3 volts. The 10K resistor from the non-inverting input (+) of the Fairchild F-710C amplifier to ground is selected so that both inputs to the amplifier will see the same impedance. The signal has now been transformed from a sinewave to a square wave of the same frequency.

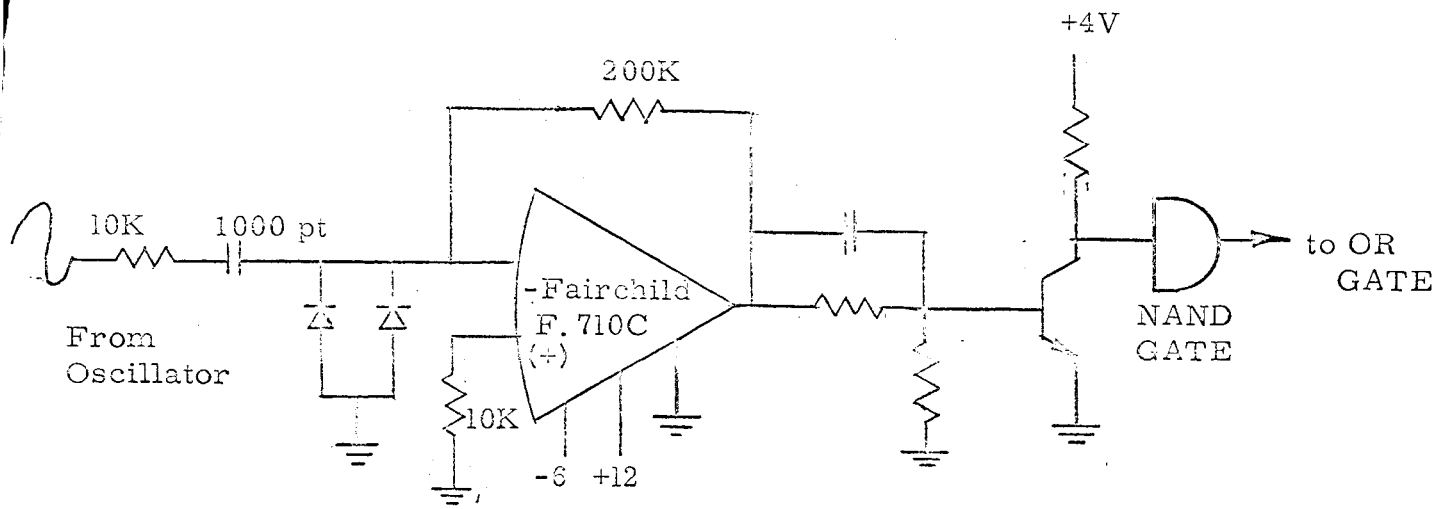


Figure 4.7 Limiting And Squaring Circuit

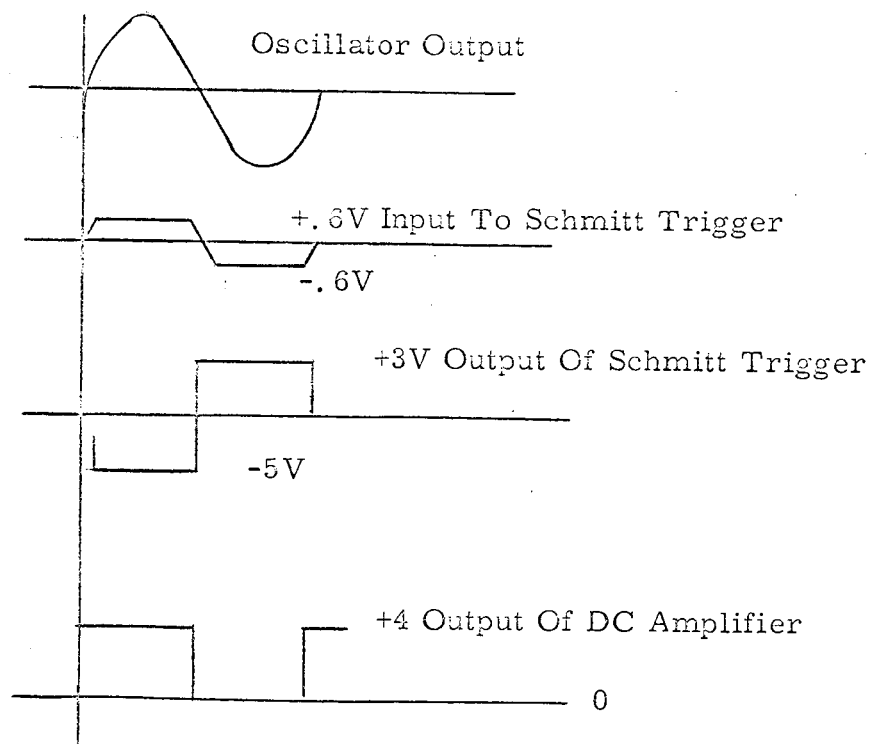


Figure 4.7B Waveforms Of Limiting And Squaring Circuit

The NPN transistor amplifier following the schmitt trigger is used mainly to change the level of the square wave shifting from +4 volts to ground, so that the square wave is compatible to drive the A HG1 Nand gate. When the input to this stage is negative the transistor Q is cut off and the output to the Nand gate is + 4 volts. However, when the input becomes positive the base to emitter becomes forward biased causing the transistor to conduct hard and the output to the Nand gate approaches ground.

4.9 OUTPUT BUFFER AND LEVEL SHIFTER

The output of the 68th stage of the shift register goes to the input of the first stage by trigger steering previously explained, and the output is also to telemetry through a buffer and level shifter stage as seen in the overall schematic.

With the output of the 68th stage at ground potential transistor Q_{13} is cut-off with approximately -.64 volts of back bias applied to the base. This means that the base of Q_{14} is forward biased and this stage is in hard conduction, hence approximately 4.35 volts would be applied across R 76; thus the output is following the input minus V BE. The diode, CR8, prevents more than .6V reverse bias to be applied to transistor Q_{14} and also any capacitance discharge loading from the output is fed through the diode and then through the low impedance saturated transistor.

When the positive pulse arrives at the input, Q_{13} becomes forward biased and conducts, placing the junction of R74 and R75 very close to ground

Now Q_{14} is cut-off and the output zero. Again, should there be any capacitance loading try to discharge through R76 the diode would reduce the impedance with a parallel path. The output voltage levels to telemetry then are ground and 4.3 volts.

V. RELIABILITY AND FAILURE ANALYSIS

5.1 SUMMARY

In accordance with specifications, a reliability analysis of the Antenna Capacitance Probe has been conducted. The specifications require a probability of complete success of at least 0.8 and a probability of complete failure not greater than 0.05. The reliability analysis indicates a probability of complete success of 0.94937 and a probability of complete failure of 0.044012.

5.2 GUIDELINES FOR ANALYSIS

As suggested by the Military Standardization Handbook "Reliability Stress and Failure Rate Data for Electronic Equipment," MIL-HDBK-217, this analysis assumes that the reliability of complex electronic equipment depends primarily upon the reliability of the parts used in the equipment. The part failure rate is the parameter on which the estimate is made. The part failure rates given in MIL-HDBK-217 are primarily applicable to standard parts subjected to electrical stresses between 0.1 (10%) and 1.0 (100%) of their rated values. Failure data relative to stresses below 0.1 were either statistically insignificant or indicated that 0.1 was the threshold below which little or no reduction in failure rate could be anticipated.

Although a complete reliability analysis should include both catastrophic (random) and performance degradation failures, most degradation failures are eliminated through conservative design. Therefore, equipment failure is

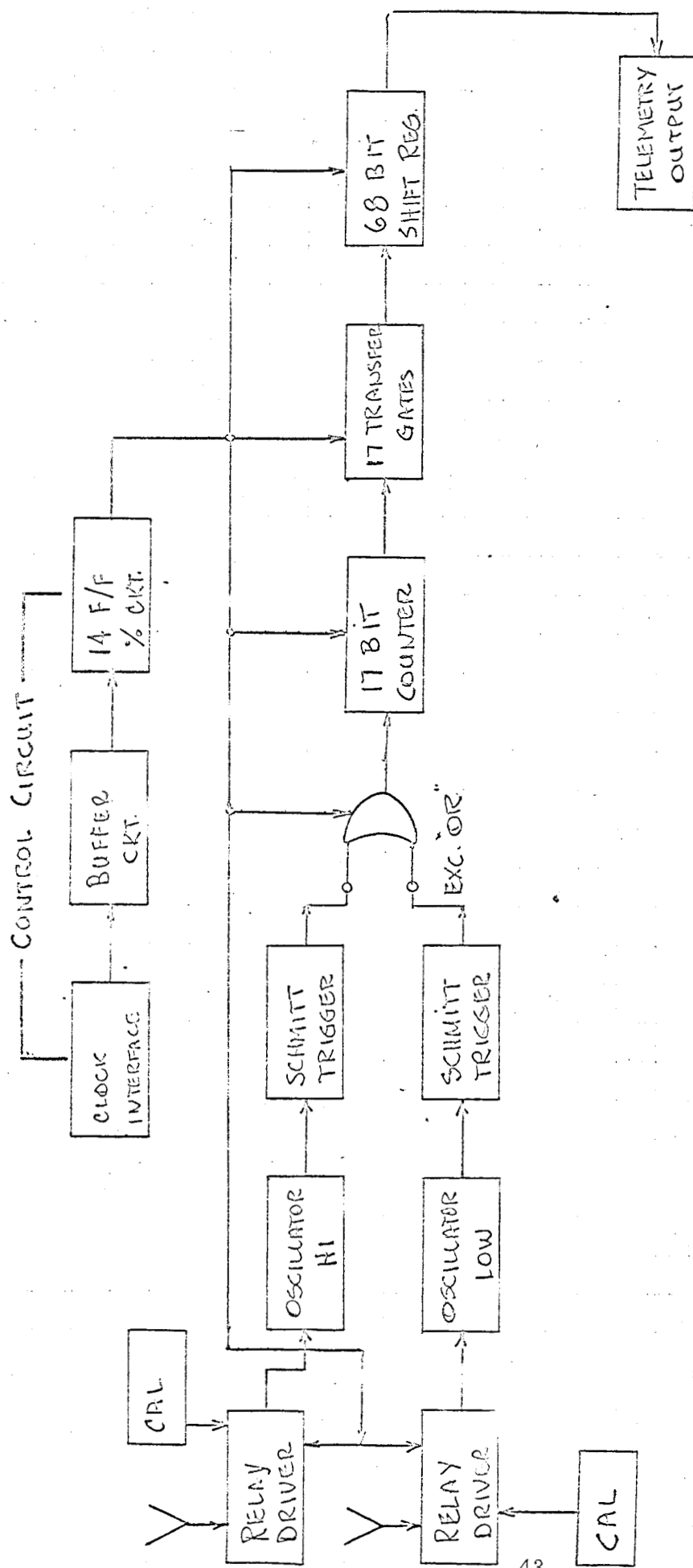
primarily dependent upon the number of catastrophic failures that occur as a function of time.

The following reliability report covers the Radio Astronomy Experiment and the procedure for this report is based on the following assumptions: (a) any failure that occurs will be considered a catastrophic failure; (b) catastrophic part failures occur randomly and exhibit a constant failure rate; (c) all component failures are to be considered catastrophic failures with certain exceptions.

5.3 SYSTEM RELIABILITY

The system reliability shall be computed on the basis of using most recent, manufactures reliability reports and MIL HDBR 217 to establish those reliability data not available from manufacturers specifications.

The system under consideration is shown in Figure 5. 1. All components are operated at less than 50% of their rated maximum operational specifications. The system can be considered a complete failure if the nexus between input and output is broken due to an open or short failure. The control circuit is not part of the series path for data flow however, its prime function is to give the data to the output and therefore a failure in this function is catastrophic.



SYSTEM BLOCK DIAGRAM

Figure 5.1

Circuit Tabulation:

"17 Bit Counter"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
SN 5112	TI	15	0.016%/1000HR*	0.24%/1000HR
SE 1243	Signetics	2	0.04%/1000HR**	0.08%/1000HR
				0.32%/1000HR

"Transfer Gates"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
SN 5161	TI	5	0.016%/1000HR	0.08%/1000HR
SN 516A	TI	1	0.016%/1000HR	0.016%/1000HR
				0.096%/1000HR

"Shift Register"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
SN 5111	TI	17	0.016%/1000HR	0.272%/1000HR
SN 510	TI	51	0.016%/1000HR	0.816%/1000HR
				1.088%/1000HR

** All Signetics failure data based on interpolated data taken from Signetics memo "Present Reliability Status" 28 February 1964. 60% CL at 55°C

* All TI Failure report data obtained from 1964 supplement to 1963 Failure Report and is established for 60% CL at 55°C.

"Control Circuit"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
SN 511 A	TI	9	0.016%/1000HR	0.144%/1000HR
SN 5111	TI	5	0.016%/1000HR	0.080%/1000HR
SN 512A	TI	3	0.016%/1000HR	0.048%/1000HR
SN 517A	TI	4	0.016%/1000HR	0.064%/1000HR
SN 514A	TI	2	0.016%/1000HR	0.032%/1000HR
SN 513 A	TI	8	0.016%/1000HR	0.128%/1000HR
Transistor 2N929	TI	2	0.020%/1000HR	0.040%/1000HR
Resistor	Welwyn	15	0.001%/1000HR	0.015%/1000HR
Capicator MC80	Aerovox	9	0.001%/1000HR	0.009%/1000HR
				0.560%/1000HR

"Exclusive Or Gate"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
SE 115G	Signetics	2	0.04%/1000HR	0.08%/1000HR
				0.08%/1000HR

"Schmitt Trigger"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
F 710 IC	FCHD	1	0.020%/1000HR	0.020%/1000HR
Transistor 2N929	TI	1	0.020%/1000HR	0.020%/1000HR
Diode 1N916	FCHD	2	0.010%/1000HR	0.020%/1000HR
Capicator MC 80	Aerovox	2	0.001%/1000HR	0.002%/1000HR
Resistor F-25 1/4W	Welwyn	6	0.001%/1000HR	0.006%/1000HR
				0.068%/1000HR

"Relay Driver"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
Transistor 2N 871	FCHD	2	0.020%/1000HR	0.040%/1000HR
Diode 1N916	FCHD	1	0.010%/1000HR	0.010%/1000HR
Relay BRJ12C1P6AS	Filtors	1	0.011%/1000HR	0.011%/1000HR
Resistor F-25 1/4W	Welwyn	3	0.001%/1000HR	<u>0.003%/1000HR</u> 0.064%/1000HR

"Oscillator"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
Transistor 2N 915	TI	2	0.020%/1000HR	0.040%/1000HR
Transformer	WTA	1	0.050%/1000HR	0.050%/1000HR
Capicator DM 15	ARCO	13	0.001%/1000HR	0.013%/1000HR
Resistor F-25 1/4W	Welwyn	8	0.001%/1000HR	<u>0.008%/1000HR</u> 0.111%/1000HR

"Calibrate Network"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
Capicator DM 15	ARCO	2	0.001%/1000HR	<u>0.002%/1000HR</u> 0.002%/1000HR

"Telemetry Output"

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
Transistor 2N930	TI	2	0.020%/1000HR	0.040%/1000HR
Diode 1N916	FCHD	1	0.010%/1000HR	0.010%/1000HR
Capicator DM 15	ARCO	1	0.001%/1000HR	0.001%/1000HR
Resistor F-25 1/4W	Welwyn	6	0.001%/1000HR	<u>0.006%/1000HR</u> .057%/1000HR

Miscellaneous

<u>Component Type</u>	<u>Manufacturer</u>	<u>Quantity</u>	<u>Failure Rate</u>	<u>Total Failure Rate</u>
Filter Capicator (Tantalum)	Sprague	8	0.012%/1000HR	0.096%/1000HR
Connectors	Cannon	2-1a	0.011%/1000HR	0.011%/1000HR
		1a	0.004%/1000HR	<u>0.004%/1000HR</u>
				0.021%/1000HR

$$\lambda_T = \sum_{i=1}^{i=N} \lambda_i + \lambda_{i+1} \dots \lambda_N$$

λ_i = Part Failure Rate (%/1000HR)

λ_T = Composite Failure Rate for the Part Group Under Study

Item	Total Failure Rate (%/1000HR)	
17 Bit Counter	0.32	
Transfer Gates	0.096	
Shift Register	1.008	
Control Circuit	0.560	
Exclusive Or Gate	0.080	
Schmitt Trigger	2 a 0.068	.136
Relay Driver	2 a 0.064	0.128
Oscillator	2 a 0.111	0.222
Calibrate Network	2 a 0.002	0.004
Telemetry Network		<u>0.057</u>
		2.611 %/1000HRS

$\lambda_T = 2.611$ - The total is in accordance with a 60% confidence level
and TA= +55°C (Max thermal stress)

The probability of complete success of the entire Experiment

Package:

$$P_s = e^{-\lambda_T t}$$

where t = total operating time

$$t = 2000 \text{ HRS}$$

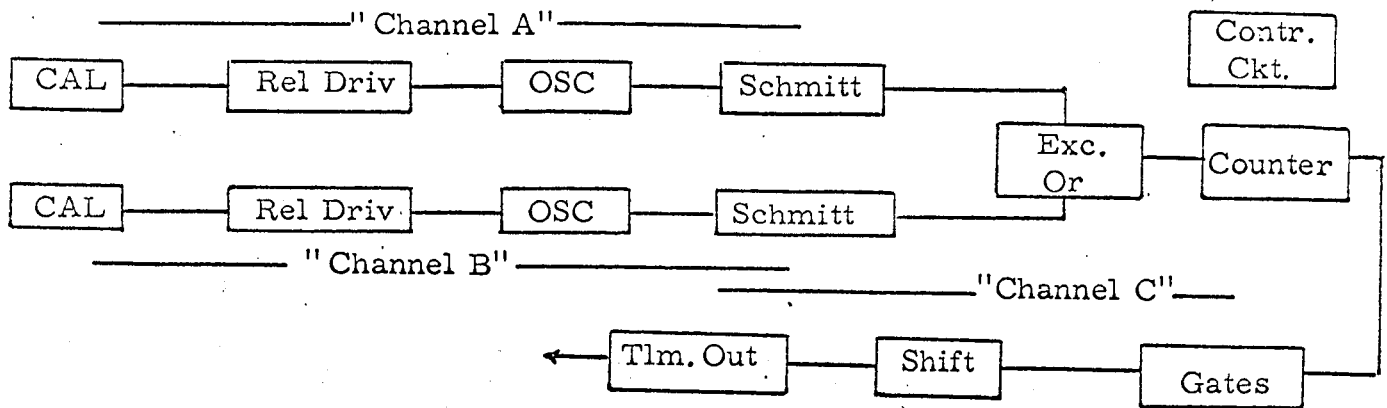
$$\therefore P_s = e^{-(2.611\%/1000\text{HRS})(2000 \text{ HRS})}$$

$$P_s = 0.94937$$

$$\text{Mean Time Between Failure} \quad \text{MTBF} = \frac{1}{\lambda_T} = 38.250 \times 10^3 \text{ HRS}$$

$$P_s \text{ including } 400 \text{ hrs. test time} = 0.93950$$

Probability of Total Failure



P_a = Probability of success of Channel A

P_b = Probability of success of Channel B

P_c = Probability of success of Channel C

Failure Rate of Channel "A"

Cal. Network	.002%/1000HR
Oscillator	.111%/1000HR
Relay Driver	.064%/1000HR
Schmitt Trigger	<u>.068%/1000HR</u>

.245%/1000HR

-.245%/KHR(2000HR)

$$P_a = e$$

$$P_a = 0.99512$$

Failure Rate of Channel "B"

$$P_b = P_a = 0.99512$$

Failure Rate of Channel "C"

17 Bit Counter	0.32%/1000HR
Transfer Gates	0.096%/1000HR
Shift Register	1.008%/1000HR
Control Circuit	0.560%/1000HR
Exclusive Or	0.080%/1000HR
Telemetry Network	<u>0.057%/1000HR</u>
	1.121%/1000HR

$$P_c = e^{-1.121\%/KHR} \quad (2000HR)$$

$$P_c = \frac{1}{1.046} = 0.95602$$

$$P_{c \text{ Fail}} = 1 - P_c = .04398$$

Probability Table

Channel A	Channel B	Channel C
B	G	G
G	G	G
B	B	G✓
G	B	G
B	G	B✓
G	G	B✓
B	B	B✓
G	B	B✓

✓ = Criteria for total failure
B = Bad
G = Good

$$P_f = \left[P_c (1-P_a) (1-P_b) + P_b (1-P_a) (1-P_c) + P_a P_b (1-P_c) \right. \\ \left. + (1-P_a) (1-P_b) (1-P_c) + P_a (1-P_b) (1-P_c) \right]$$

$$Pb = Pa$$

$$Pf = Pc (1-Pa)^2 + (Pa-Pa^2) (1-Pc) + Pa^2 (1-Pc) + (1-Pa)^2 (1-Pc) \\ + (Pa-Pa^2) (1-Pc)$$

$$= Pc (1-Pa)^2 + 2 (1-Pc) (Pa-Pa^2) + Pa^2 (1-Pc) + (1-Pa)^2 (1-Pc)$$

$$= 23.698 \times 10^{-6} + 433.364 \times 10^{-6} + .04355 + 1.0473 \times 10^{-6}$$

$$Pf = 0.044012$$

APPENDIX A
SPECIFICATION FOR
ENGINEERING MODEL OF THE
CAPACITANCE PROBE

1. SCOPE

The unit supplied will be used in engineering integration tests of the electronic system of the RAE satellite. The purpose of these integration tests is to make possible the writing of an accurate specification for the prototype and flight models of this unit for the satellite. This satellite will be a non-scaled satellite with a circular orbit inclined 50 degrees to the equator at an altitude of 6000 kilometers.

2. APPLICABLE DOCUMENTS

The provisions of the following specifications, standards, drawings and publications of date of current issue will apply as detailed in this specification. In the event of a conflict, the detailed provisions of this specification shall govern.

MSFC Std. 154	Printed circuit design and construction.
MS 33586	Metals, definition of dissimilar electric.
MIL-I-6181D	Interface control requirements, air-craft equipment.
S-320-D1	General environmental test specification for Spacecraft and components.
GSFC TID-S-100	Specification for contractor-prepared monthly, periodic, and final project reports.
GSFC X623-63-147	Specification drawings, engineering and associated lists.

MIL-JDBK-217	Reliability stress and failure rate data for electronic equipment.
GSFC-PPL-3	Preferred parts list.
FED-STD-102	Preservation packaging, and packing levels.
FED-STD-123	Marking for domestic shipment.
2.1 <u>Mechanical Drawings</u>	
GB-RAE-1155-001	Standoff
GB-RAE-1155-002	Insulator Board
GB-RAE-1155-003	Spacer
GB-RAE-1155-004	Can
GB-RAE-1155-005	Encapsulation Procedure
GC-RAE-1155-009	Retaining Strip
GC-RAE-1155-011	Rear Panel
GD-RAE-1155-500	Container Cover
GD-RAE-1155-505	Cover
GD-RAE-1155-504	Right Side Panel
GD-RAE-1155-507	Container
GD-RAE-1155-511	Left Side Panel
GF-RAE-1155-512	Capacitance Probe Electronic Enclosure Assembly
GF-RAE-1155-513	Front Panel

2.2 Electrical Drawings

GC-RAE-1155-006	Printed Wiring Assy Control #4
GC-RAE-1154-007	Printed Wiring Board Control #4
GC-RAE-1155-008	Printed Wiring Main Board
GC-RAE-1155-010	Printed Wiring Board Test Limiter
GR-RAE-1155-101	Printed Wiring Assy Shift Register
GD-RAE-1155-203	Welded Module T & R Driver
GD-RAE-1155-204	Printed Wiring Assy Test Limiter
GD-RAE-1155-205	Printed Wiring Assy RAE Oscillator
GR-RAE-1155-206	Schematic Diagram RAE Capacitance Probe Experiment
GD-RAE-1155-402	Printed Wiring Assy Counter Shift Register
GD-RAE-1155-403	Printed Wiring Assy Control #1
GD-RAE-1155-404	Printed Wiring Assy H. S. Counter Shift Register
GD-RAE-1155-405	Printed Wiring Assy Control #2
GD-RAE-1155-406	Printed Wiring Assy Control #3
GD-RAE-1155-501	Printed Wiring Board Control # 3
GD-RAE-1155-502	Printed Wiring Board H. S. Counter Shift Register
GD-RAE-1155-503	Printed Wiring Board Shift Register

GD-RAE-1155-506	Printed Wiring Board Control #2
GD-RAE-1155-508	Printed Wiring Board Control Circuit#1
GD-RAE-1155-509	Printed Wiring Board Counter Shift Register
GD-RAE-1155-510	Printed Wiring Board RAE Oscillator
GR-RAE-1155-514	Printed Wiring Assy Main Wiring

2.3 Other Documents

Report No. NAS 9175-1	Final Report on "Engineering Model of Antenna Capacitance Probe", Part No. 2158-R-434, under Contract No. NAS 5-9175
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3.1 REQUIREMENTS

The capacitance probe shall be designed and constructed to satisfy all of the requirements specified herein.

3.1 General

3.1.1 Materials

Materials and processes used during manufacture of items covered herein shall be of high quality, suitable for the purpose, and conform to applicable government and contractor's specifications whenever possible. Materials and components of proven reliability shall be employed to the greatest practicable extent.

3.1.1.1 Metals

All metals which are used in the construction of the items covered by this specification that are subject to deterioration when exposed to the climate and environmental conditions specified in U. S. Government Spec. S-320-D1 shall be protected in a manner that will in no way prevent compliance with this specification. The use of protective coatings that will crack, chip or scale with age or extremes of climatic and environmental conditions shall be avoided.

Dissimilar metals as defined by MS 33586 shall not be used in intimate contact unless protected against electrolytic corrosion.

3.1.2 Interchangeability

All parts having the same manufacturers parts numbers shall be directly and completely interchangeable with each other with respect to installation and performance, except as approved specifically by the purchaser.

3.1.3 Workmanship

"The units including all parts and accessories shall be manufactured in a thorough workmanlike manner in accordance with best commercial practice. The provisions of MSFC 154, 270, 271 and NPC 200-4 shall be applicable. Component interconnections shall be soldered."

3.2 Electrical Requirements

The capacitance probe shall monitor a dipole antenna in accordance with the operational program of spacecraft and shall develop a signal which indicates the capacitive reactance of the antenna. This signal shall be processed and stored in a format which is compatible with the spacecraft telemetry system.

The instrumentation for the antenna capacitance measuring system shall be divided into two packages. One package shall be located near the antennas to be measured and shall contain the capacitance oscillators, the calibration capacitors, and associated switching relays and gates. The main package shall be located in the main body electronics stack and will contain power conversion, counter circuit, transfer gates, shift register, clock frequency conversion logic and control logic.

3.2.1 Antenna Capacitance Oscillator

Two closely coupled oscillators shall be located in the package near the antenna terminations. These oscillators will operate at 750 kc and 1 mc, respectively. The oscillator design shall be basically LC units with a short term

stability of $\pm 0.05\%$ from 2.5 seconds after the application of power until the removal of power. The basic oscillator when connected to the antenna will cause a maximum excitation of 5 volts peak to peak to be applied to the antenna.

The respective oscillators shall assume a calibration frequency of 750 kc and 1 mc when their capacitance standard is switched into the circuit. If the antenna capacitance changes, then the basic oscillator frequency will shift. Therefore, f becomes proportional to c . A DC bias can be applied to the antenna if a desired potential is required between the antenna and the spacecraft.

The oscillators shall assume a calibration frequency of 500 kc when a capacitance standard is switched into the circuit. These calibration circuits shall allow a check to be made on the basic oscillator frequency to determine if a drift is occurring.

Upon command from satellite control circuitry, the oscillators shall be gated with logic gating so that their signals shall pass alternately to the counter circuitry.

3.2.2 Counter

A seventeen stage counter shall be used to count the 4 respective oscillator frequencies, two measuring frequencies and two calibrate frequencies. These seventeen stages shall allow a maximum frequency of one megacycle to be recorded. The counter circuit shall have a sample period of 100 milliseconds.

The basic accuracy of the capacitance measuring system shall be achieved by having a sample period of sufficient time to record the frequency to the desired accuracy and by holding the sampling period very accurately.

3.2.3 Transfer Gates

The transfer gates shall be activated by a control signal at the end of the 100 millisecond sample period. These gates shall transfer the binary information from the counter to the shift register.

3.2.4 Shift Register

The shift register shall be composed of 68 stages which shall allow the storage of four, seventeen bit binary words. After the transfer of each seventeen bits and be ready to accept the next binary word into the register. The shift time shall be less than 100 milliseconds. The four digital words shall be accumulated and stored in the following order: high frequency calibration; low frequency calibration; high frequency; and low frequency.

Upon command from spacecraft control circuitry, the shift register will accept the spacecraft shift pulses so that the data can be transferred serially. The output of the last shift register stage will be monitored by telemetry. The rate of shift out will be 400 bits per second. The output of the last stage will be buffered to conform to interface requirements as specified by GSFC.

The shift register shall be capable of being cycled at speeds up to 250,000 bits/second.

3.2.5 Experiment Clock

The experiment clock shall operate from the basic satellite clock pulse of 400 cps square wave, and the basic sampling time of the instrument shall be derived from this clock, which has a frequency stability of ± 0.1 percent.

The experiment clock shall provide reset signals for the counter and the shift register. The clock shall provide transfer signals for the transfer gates, commands for the gates which control the counter input, signals to drive the oscillator calibrate relay, and experiment shift pulses to shift each word in the register during the data accumulation cycle.

3.2.6 Control Signals

The unit shall operate in conjunction with the external control signals listed below.

3.2.6.1 Power Control

Power shall be externally switched on and off to the unit so that power shall be available only during operating periods.

3.2.6.2 Accumulate Command

An accumulate command shall be provided to the unit to start the data accumulation cycle. It shall be provided at not less than 2.5 seconds after power is provided to the unit.

3.2.6.3 Clock Pulses

Clock pulses shall be provided to the unit as discussed in paragraph 3.2.5.

3.2.7 Accuracy

The overall accuracy of the instrument shall be ± 0.15 percent or better, including the basic satellite clock error.

3.2.8 Power

The unit shall operate from +6 vdc, -6 vdc, and +3 vdc. These voltages are regulated to $\pm 1\%$ maximum power consumption shall not exceed the values given below.

+6 volts 40 MA	0.24 watt
-6 volts 20 MA	0.12 watt
+3 volts 300 MA	<u>0.90 watt</u>

Total Power Dissipation	1.36 watt
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Not more than 1.36 watts shall be dissipated by the unit for a period of 3.6 seconds every 10 minutes.

3.2.9 Equipment Grounds

Power supply and signal returns shall be isolated from each other and from the case.

3.1.10 Telemetry

The unit shall be suitable for operation with a PCM telemetry system at a rate of 400 to 800 bits per second.

3.3 Mechanical Requirements

3.3.1 Configuration

The instrument shall be packaged in two sections. The main body shall be a standard RAE instrument can 5 x 7 inches with a height of 1-1/4 inches. The other package to be located at the antenna base shall not exceed 9 cubic inches. The actual shape shall be specified by GSFC.

3.3.2 Weight

The total weight of this unit shall not exceed 1.75 pounds. Reduction in weight shall be a prime design consideration.

3.3.3 Packaging

The unit shall be packaged modularly using either printed circuit board or welded module techniques. Two contacts shall be provided on the power supply connector (s) for each power supply connection. All exterior surfaces shall be protected from the environment specified.

3.3.4 Marking

The marking must not deface or damage the equipment or affect its functional use. The marking should be legible with the item mounted in normal position. Each item shall be legibly marked with the following information as a minimum.

Specification Number including latest Revision Number

Part Number and Name

Serial Number

Date of Manufacture (month and year)

Contractor's name and trademark

This information shall preferably be applied to a nameplate (or tag on items too small for a nameplate) securely fastened to the item. It may be applied directly to the surface of the item by acid or electric etching, engraving, silk screen process branding, embossing, casting, molding or by the metal-cal process. The permanency of the marking shall be sufficient to withstand the environmental conditions specified. Decals and stencils are not acceptable.

3.4 Environmental Requirements

The unit shall meet all electrical and mechanical specifications throughout the range of environmental conditions specified in this section.

3.4.1 Altitude

The unit shall be capable of operation at a pressure of 10^{-13} torr. Hermetic sealing of parts and assemblies shall be provided to the extent required.

3.4.2 Temperature

The operating temperature range shall be -20°C to $+70^{\circ}\text{C}$, and the highest rate of temperature cycling from one extreme to the other shall be 15 minutes. The storage temperature range shall be -30°C to $+70^{\circ}\text{C}$.

3.4.3 Humidity

The unit shall operate relative humidity range of 0 to 98 percent.

3.4.4 Vibration

The unit shall meet the sinusoidal and random vibration levels as specified in paragraph 4.2.2.

3.4.5 Acceleration

The unit shall meet the acceleration levels as specified in paragraph 4.3.5 of specification S-320-D-1.

3.4.6 Thermal-Vacuum

The equipment shall operate within specification while exposed to 10^{-13} torr over the range of -20°C to $+70^{\circ}\text{C}$.

3.4.7 Radio Frequency Interference

The unit shall conform to the following:

3.4.7.1 Generation

a) Radio Frequency Conducted

The equipment shall be designed to meet the requirements specified in MIL-I-6181D para. 4.3.1 in the frequency range 0.15 to 25 Mc.

b) Low Frequency Conducted

Interface voltages and currents in the frequency range 35 cps to 150 kc in excess of 30 mv and 10 ma peak-to-peak shall not appear on any conductor external to the equipment, which could conduct interference to other equipment.

c) Radiated

Equipment is to meet the requirements specified in MIL-I-6181D para. 4.3.2 in the frequency range 0.15 to 200 Mc.

3.4.7.2 Susceptibility

a) Radio Frequency Conducted

Equipment is to meet the requirements specified in para. 4.3.4.1.1 of MIL-I-6181D. Tests shall be made over 0.15 to 400Mc.

b) Audio Frequency Conducted

No change in indication, malfunction or degradation of performance shall be produced in any equipment when a sine wave signal of 100 mv rms or when a step function or low frequency square wave of 300 mv amplitude having a rise time of less than 1 microsec is applied as shown in Fig. 19 of MIL-I-6181D. Tests shall be made over the frequency range 35 cps to 150 kc.

in performance) during a mission of 10,000 hours (2,000 hours operating time) greater than 0.8 (confidence level 60 percent).

Both of these figures refer to operation in the space environment.

4. SAMPLING, INSPECTION, AND TEST PROCEDURES

4.1 Quality Assurance

The contractor shall conform to specification NASA Quality Publication NPC 200-3, April 1962 entitled, "Inspection System Provisions for Suppliers of Space Materials, Parts Components and Services". Wherever possible, parts shall be selected from the following documents:

- 1) GSFC-PPL-3, 1 Jan. 65, Preferred Parts List,
- 2) JPL Preferred Parts List for Spacecraft Scientific Experiments,
- 3) OGO Electronic Parts Specification Development Guide.

All part ratings shall be derated for design use as necessary to achieve, in conjunction with simplified design and redundancy where necessary, the specified reliability for the unit.

The reliability analysis required as a part of the final report shall be based on MIL-HDBK-217, 8 Aug. 62, "Reliability Stress and Failure Rate Data for Electronic Equipment".

All work on this order is subject to inspection and test by NASA or its designated representative at all times (including period of performance) and places.

Final acceptance will be performed at GSFC.

4.2 Test Program

The contractor shall fully test each unit to insure full compliance with all electrical and mechanical requirements of the specification.

Although each unit shall be bench-tested as above, only one unit shall be tested for compliance with the environmental requirements, and only the following tests shall be performed (in the order listed):

- 1) temperature,
- 2) radio frequency interference,
- 3) vibration
- 4) thermal-vacuum

The above tests shall be performed prior to delivery of the units to GSFC.

All measurements shall be made with instruments which have had their accuracy verified periodically against standards traceable to the National Bureau of Standards.

Data shall be provided to GSFC as specified in para. 4.3 of this specification.

4.2.1 Temperature

4.2.1.1 Cold Storage

- a) Thermocouples to indicate temperature stabilization shall be attached to the component in sufficient number and at such locations as to measure the highest and lowest component temperature. Stabilization shall be considered achieved when no thermocouple varies by more than 0.5°C/hr .
- b) The instrumented component, while non-operative, shall be placed in a test chamber which shall be stabilized at $-30^{\circ}\text{C} \pm 2^{\circ}\text{C}$. The

component shall remain so exposed for six hours.

- c) Following this exposure, the temperature of the test chamber shall be increased to $+25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ and the component operated typical of orbital flight for sufficient duration to determine compliance with its performance specification.

4.2.1.2 Hot Storage

- a) The component shall be placed in a nonoperative state and the temperature of the test chamber increased in one hour or less to $+70^{\circ}\text{C} \pm 2^{\circ}\text{C}$. After stabilization of chamber temperature, the component shall remain exposed to $+70^{\circ}\text{C}$ for six hours.
- b) Following this exposure the temperature of the test chamber shall be reduced to $+25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ and the component operated typical of orbital flight. The duration of operation shall be sufficient to determine compliance of the component with its performance.

4.2.1.3 Operational Temperature

- a) The instrumented component, while non-operative, shall be placed in a test chamber. The temperature of the chamber shall be adjusted such that the temperature of the component becomes stabilized at $-20^{\circ}\text{C} \pm 2^{\circ}\text{C}$. After temperature stabilization, the component shall be operated for one hour minimum and its performance checked. For cyclically operated components ("on-off" orbital operation), cold start capability shall be demonstrated at least three times. Each cycle of operation shall be

preceded by reverting to stabilized conditions at $-20^{\circ}\text{C} \pm 2^{\circ}\text{C}$. The total duration of operation shall be one hour minimum.

- b) Immediately following the cold temperature exposure, the component shall be turned off and the temperature of the test chamber increased to $+70^{\circ}\text{C} \pm 2^{\circ}\text{C}$. After temperature stabilization, the component shall be operated.
- c) Following this exposure, the temperature of the test chamber shall be reduced to $+25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ and the component operated for sufficient duration to determine compliance with its performance specification.

4.2.2 Vibration

For this test, with the exception of test levels and recording on magnetic tape (recording accelerometer signals shall be recorded continuously; however, the use of magnetic tape is not mandatory) all the requirements of paragraphs 4.1.8 through 4.1.8.5 of specification S-320-D-1 shall apply. Since the tests must be performed in the absence of both a representative spacecraft structure and transmissibility measurements, the component shall be attached to the vibration excitor via a rigid fixture. Sinusoidal vibration test levels shall be as specified in paragraph 4.2.2.1. Random vibration test levels shall be as specified in paragraph 4.2.2.2.

4.2.2.1 Sinusoidal Vibration

This test shall be conducted by sweeping the applied frequency once through each range specified in the schedule. The rate of change of frequency shall be 2 octaves per minute.

SINUSOIDAL VIBRATION SCHEDULE

Axis	Frequency Range cps	Test Duration Min.	Acceleration g, 0-to-peak
Thrust (Z-Z)	10-50	1.17	3.8*
	50-110	.57	1.524 mm(.06 inch)** double amplitude constant displacement
	110-200	.43	37.0**
	200-500	.66	7.3
	500-2000	1.00	21.0
	2000-3000	1.00	15.0
	Total	4.83 Min.	

SINUSOIDAL VIBRATION SCHEDULE (con't)

Axis	Frequency Range cps	Test Duration Min.	Acceleration g, 0-to-peak
Lateral (X-X)	10-50	1.17	3.0*
and Lateral (Y-Y)	50-96	.44	.584 mm (.023 inch)** double amplitude con- stant displacement
	96-250	.72	11.0*
	250-500	.50	2.3
	500-2000	1.00	3.8
	2000-3000	1.00	15.0
	Total	4.83 min.	
	(each axis)		
	Grand Total:	13.49 Minc.	

* When the specified accelerations cannot be attained due to armature displacement limitations, the input may be a constant displacement not less than 0.5 in. double amplitude.

** In these frequency ranges, the test levels include the effects of transmissibilities through the spacecraft structure as explained in paragraph 4.3.4 of S-320-D-1.

4.2.2.2 Random Motion Vibration

Gaussian random vibration shall be applied with the "g-peaks" clipped at three times the root-mean-square acceleration specified in the schedule. With the component installed, the control accelerometer response shall be equalized such that the specified power spectral density (PSD) values are within +3 db throughout the frequency band as determined by the analysis specified in 4.1.8.3 of specification S-320-D-1. The filter roll-off characteristic above 200 cps shall be at a rate of 40 db/octave or greater.

RANDOM VIBRATION SCHEDULE

Axis	Frequency Range cps	Test Duration Min.	PSD Level g ² /cps	Acceleration g-rms
Thrust (Z-Z)	20-2000	4 (each axis)	0.07	11.8
Lateral (X-X)				
Lateral (Y-Y)				
Total: 12 Min.				

4.2.3 Thermal-Vacuum

The component shall be mounted in such a manner that it is thermally isolated from its mounting fixture. The mounting fixture for this purpose shall be designed to minimize interference of radiation paths from the radiant sources to the component. The component shall be installed in the chamber in such a way that it will not be exposed to any abnormally hot or cold source. Thermocouples shall be attached to the component in sufficient number and at such locations as to measure the highest and lowest component temperature. The instrumented component shall be placed in the test chamber, and an operational check shall be performed prior to chamber evacuation. With the component operative, typical of boost phase, the chamber shall be evacuated to 10^{-12} TORR. Corona effects shall be monitored throughout the evacuation period.

4.2.3.1 Low-Temperature Soak

Immediately following the corona check, the component shall be turned

off and the chamber adjusted such that stable conditions are established at 10^{-13} torr and $-20^{\circ}\text{C} \pm 2^{\circ}\text{C}$. Stabilization shall be assumed when indication from each thermocouple attached to the component does not change by more than 0.5°C per hour. The chamber control shall be held fixed, and the component operated throughout the exposure period of 24 hours. For cyclically-operated components ("on-off" orbital operation), cold start capability shall be demonstrated at least three times during the 24-hour exposure. Each cycle of operation shall be preceded by reverting to stabilized conditions at $-20^{\circ}\text{C} \pm 2^{\circ}\text{C}$. The temperature of the component shall be monitored continuously during the exposure. Performance of the component shall be monitored at least at the beginning and end of the exposure period. For cyclically-operated components, the performance of the component shall be checked periodically, consistent with the test duty cycle.

4.2.3.2 High-Temperature Soak

At the conclusion of the low-temperature vacuum exposure, the chamber conditions shall be altered such that stable conditions are re-established at 10^{-13} torr and $+70^{\circ}\text{C} \pm 2^{\circ}\text{C}$. Stabilization shall be assumed when indication from each thermocouple does not change by more than 0.5°C per hour. When stable conditions have been achieved, the chamber temperature controls and its temperature held fixed. The component shall be operated and its temperature monitored throughout the exposure period of 24 hours. The performance of the component shall be checked periodically as well as at the beginning and end of the exposure period.

4.2.4 Radio Frequency Interference

Test shall be performed to prove complicate with all requirements of paragraph 3.4.7.

4.3 Documentation

The basic philosophy of the data required hereunder is that it must be sufficient to summarize the program, document technical advances, fully describe developed equipment, and in particular, to permit fabrication, adjustment, test and operation of developed equipment by any qualified source.

4.3.1 Monthly Letter Reports

Informal monthly letter reports shall be provided in 3 copies by the 10th of the month following the report (calendar) month. They shall comply with requirements of TID-S-100 for type I reports, and shall include:

- (1) summary of work performed during the month;
- (2) special problems and their solutions;
- (3) whether project on schedule and whether funding adequate to complete the program;
- (4) work plan for the coming month;
- (5) recommendations, if any, for desirable specification changes;
- (6) a reliability analysis;
- (7) a complete failure analysis;
- (8) radiation effects analysis;
- (9) complete design information and drawings for any test jigs used in testing the unit;
- (10) test procedures.

5. Preparation for Delivery

5.1 General

Equipment prepared for delivery shall be packaged in manner to ensure delivery without damage. It shall be enclosed by a vapor barrier, which shall contain a desiccant. The humidity state shall be ascertainable without breakage of the vapor seal. The packaging shall also be suitable for storage. The package and equipment shall survive a temperature range (non-operating) of -30°C to $+70^{\circ}\text{C}$.

5.2 Shipping Instructions

The F. O. B. destination shall be Goddard Space Flight Center, Greenbelt, Maryland. Packages shall be marked as follows:

National Aeronautics and Space Administration

Goddard Space Flight Center

Code 615

Greenbelt, Maryland 20771

Attention: Dr. R. G. Stone

HANDLE WITH EXTREME CAUTION

Completed hardware shall be packaged as specified by level C of FED-STD-102.

NOTES: UNLESS OTHERWISE SPECIFIED:

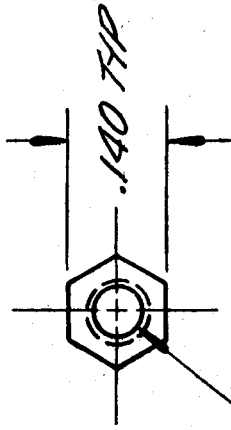
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#2-56 UNC-2A

1/16" THD RELIEF

.120

1/2

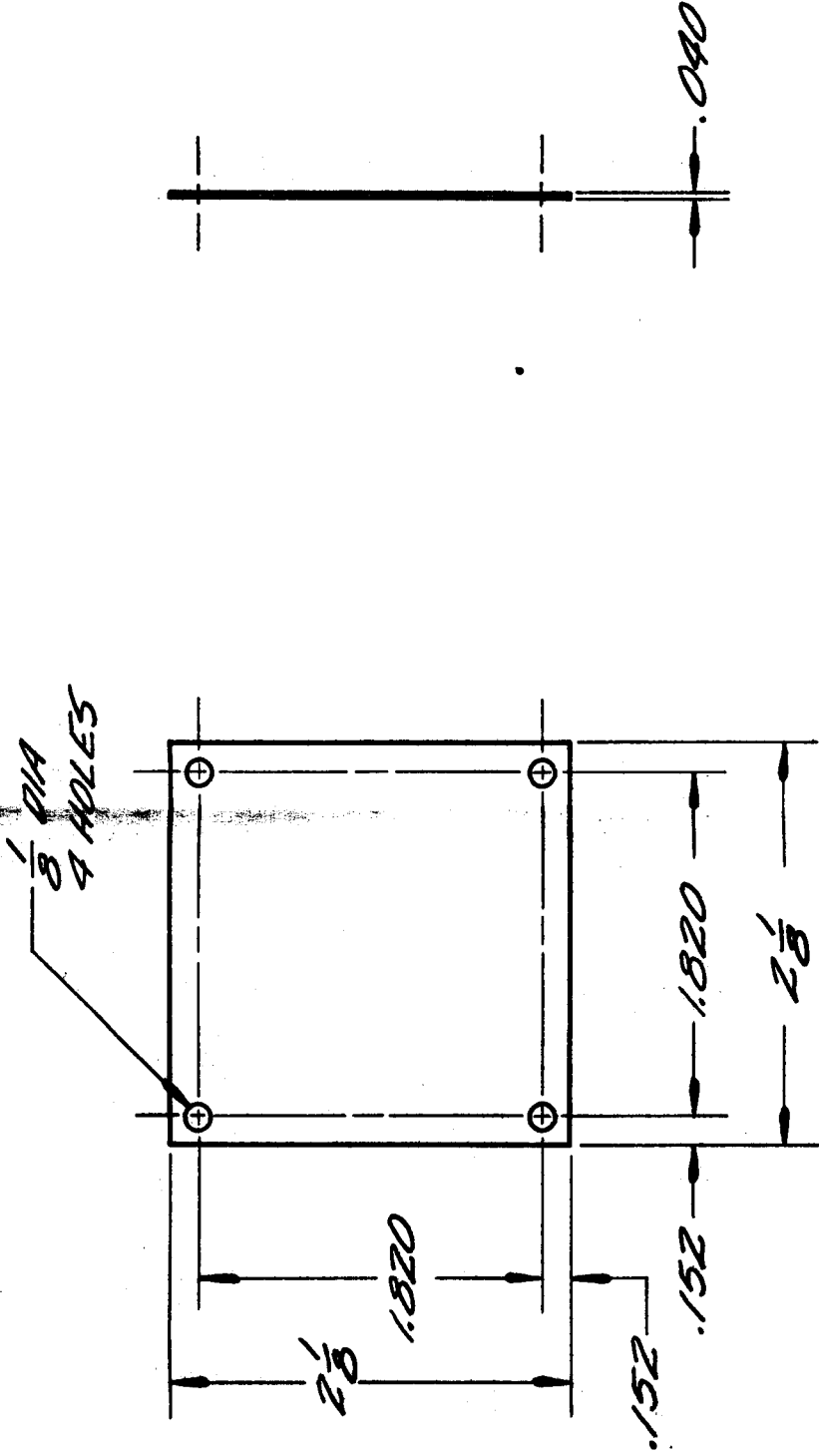


TAP DRILL .093 DEEP MAX
BOTTOM TAP #2-56 UNC-2A
X .080 MIN FULL THD DEPTH

WTA - 2158B-483

REV		DATE	BY	DESCRIPTION	APPROVED
LIST OF MATERIAL					
NAME		INIT	DATE	RAE STANDOFF	
DESIGNER				NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND	
DRAWN				GB-RAE-1155-001	
CHECKED				CODE	
APPROVED				SHEET 1 OF 1	
APPROVED				UNIT WT	
SCALE				UNIT WT	
MATERIAL		CRE5 303			
DO NOT SCALE THIS DRAWING					
REMOVE BURRS, BREAK SHARP EDGES					
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCE ON FRACTIONS ± 1/64 DECIMALS ± .005, ANGLES ± —		SCALE: 4/1 FINISH: NONE CLIENT:			
2158B-483		USED ON		APPLICATION	
NEXT ASSY.					

REV	DATE	BY	DESCRIPTION	APPVD



W.T.A. 2158-B-485

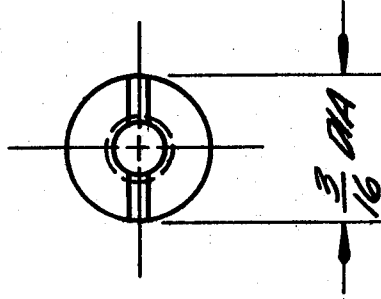
LIST OF MATERIAL

MATERIAL <i>TEFLON</i>		NAME	INIT	DATE	<div>RAE</div> <div>BOARD, INSULATOR</div>	<div>NATIONAL AERONAUTICS AND SPACE ADMINISTRATION</div> <div>GODDARD SPACE FLIGHT CENTER</div> <div>GREENBELT, MARYLAND</div>
		DESIGNER				
		DRAWN				
		W T A				
		CHECKED				
DO NOT SCALE THIS DRAWING		APPROVED			<div>SCALE</div> <div>UNIT WT</div> <div>CODE</div> <div>SHEET 1 OF 1</div>	
REMOVE BURRS, BREAK SHARP EDGES		APPROVED				
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES		SCALE: 1/1				
TOLERANCE ON FRACTIONS ± 1/64		FINISH:				
DECIMALS ± .005 ANGLES ± —		CLIENT:				
NEXT ASSY.		APPLICATION		USED ON		

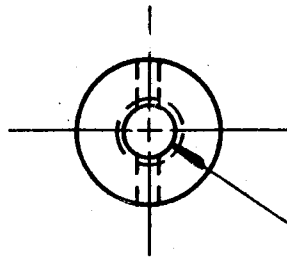
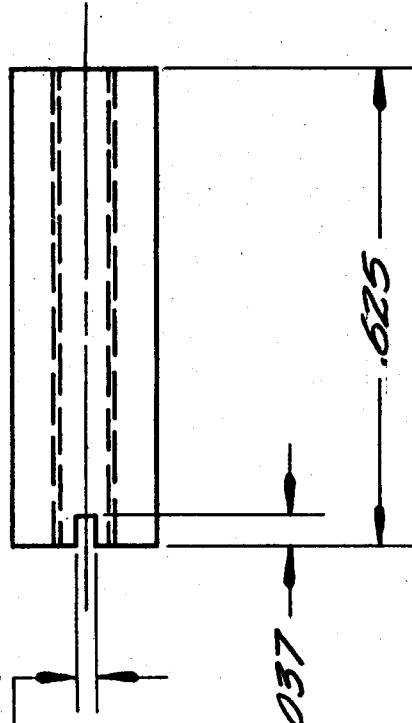
2158-B-485

NOTES: UNLESS OTHERWISE SPECIFIED:

1. $\sqrt{.63}$ FINISH ALL OVER



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#2-56 UNC-2B THRU

W.T.A. 2158-B-484

LIST OF MATERIAL

NAME	INIT	DATE
DESIGNER		
DRAWN W.T.A.		
CHECKED		
APPROVED		
APPROVED		

MATERIAL	BRASS
DO NOT SCALE THIS DRAWING	
REMOVE BURRS, BREAK SHARP EDGES	
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCE ON FRACTIONS $\pm \frac{1}{64}$ DECIMALS $\pm .005$, ANGLES \pm	SCALE: 4/1
	FINISH: <i>MILLED</i>
	CLIENT:

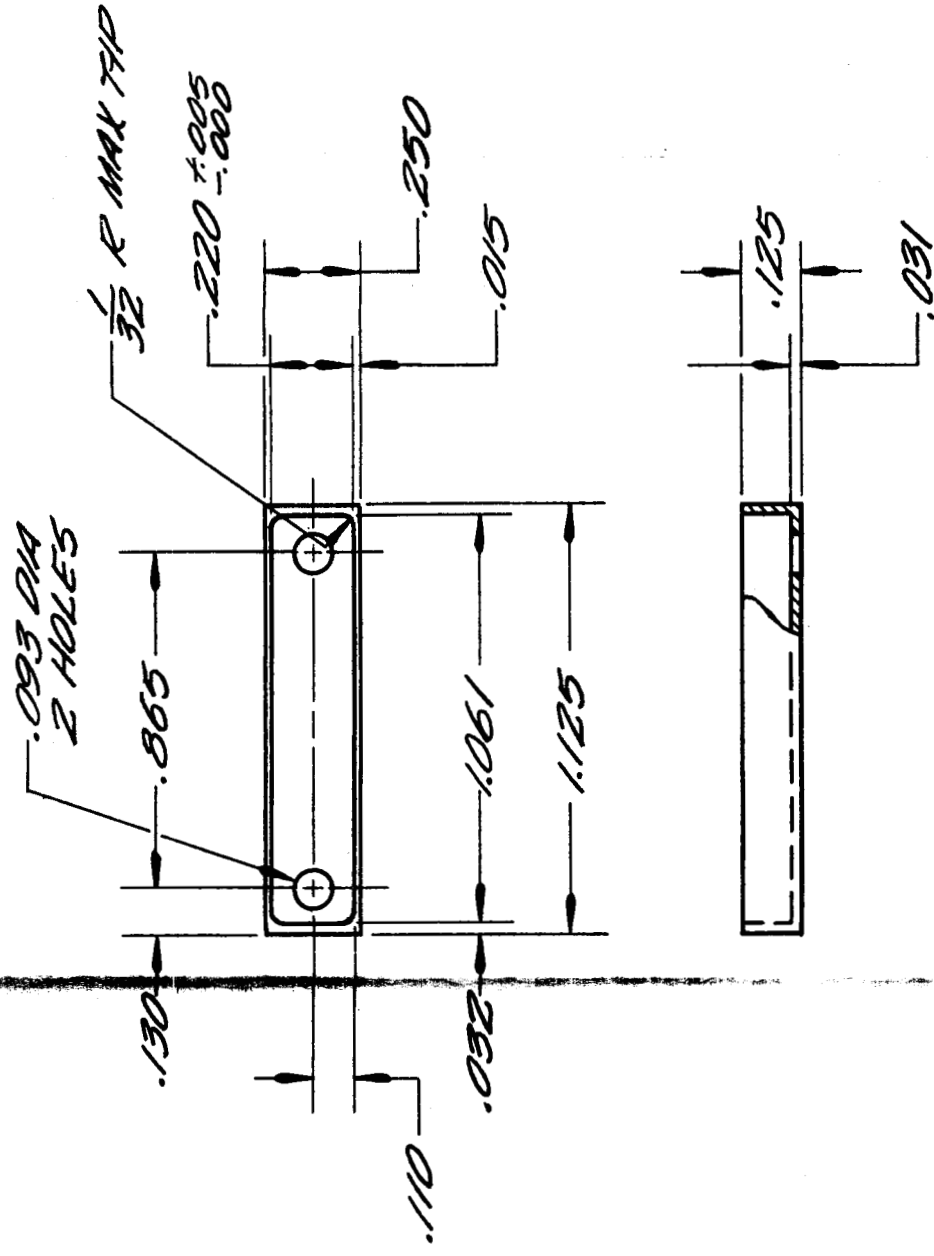
2158-B-484	USED ON	APPLICATION
NEXT ASSY.		

NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND	RAE SPACER	SCALE	UNIT WT	CODE	SHEET 1 OF 1
GB-RAE-1155-003					

REV	DATE	BY	DESCRIPTION	APPROVED

✓ 63/ FINISH ALL OVER

2. FURNISH: IDENTITE #15A MIK
ALLIED RESEARCH
BALTIMORE, MARYLAND



WTA-2158-B-481

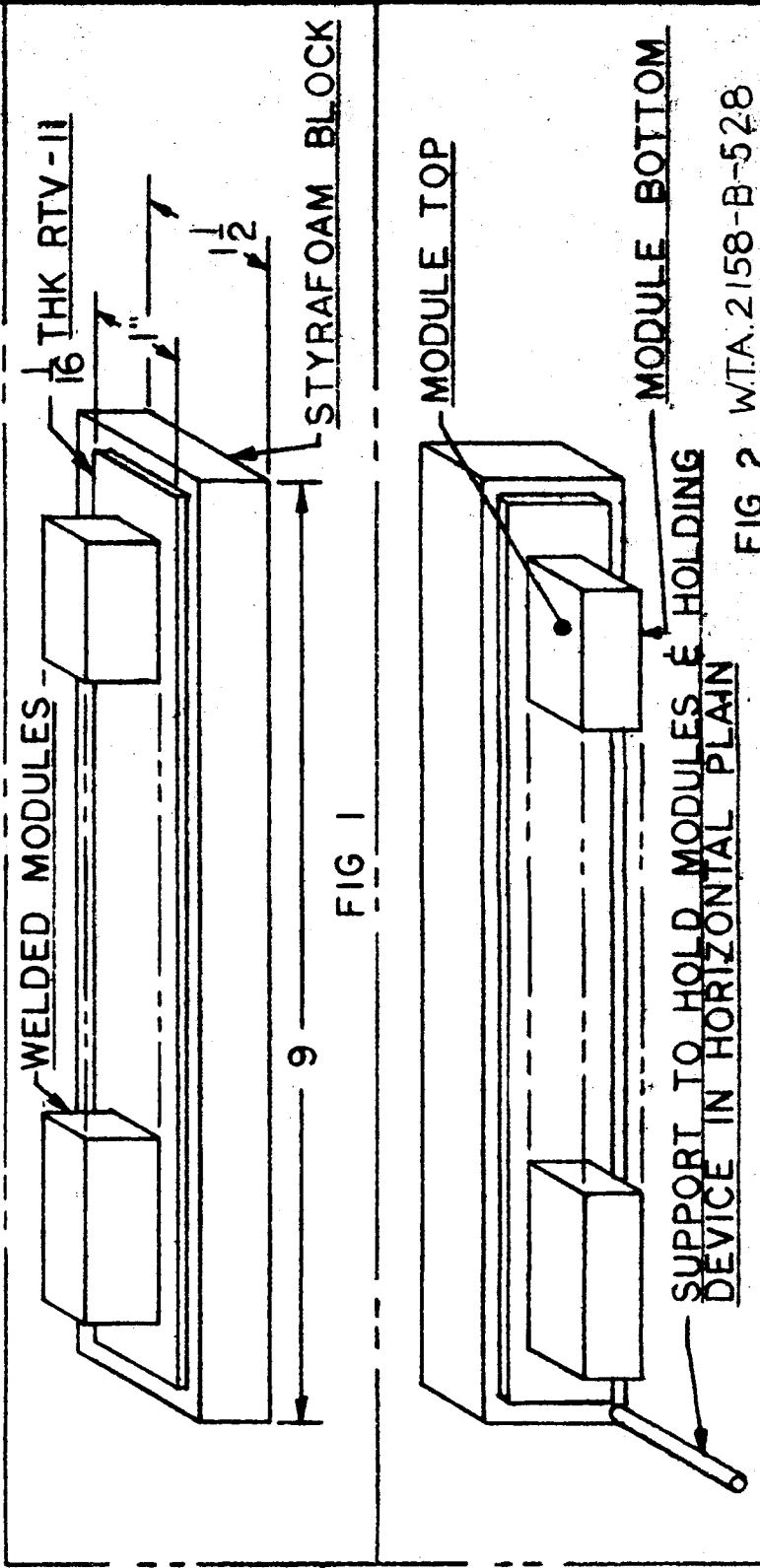
LIST OF MATERIAL									
MATERIAL		NAME		INIT	DATE	RAE CAN		NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND	
		DESIGNER							
		DRAWN							
		W T A							
		CHECKED							
		DO NOT SCALE THIS DRAWING							
		REMOVE BURRS, BREAK SHARP EDGES							
		UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES							
		TOLERANCE ON FRACTIONS $\pm \frac{1}{64}$							
		DECIMALS $\pm .005$, ANGLES \pm							
		SCALE: 2/1							
		FINISH: <i>ANNEAL</i>							
		CLIENT:							
		USED ON							
		NEXT ASSY.							
		APPLICATION							
		GB-RAE-1155-004							
		CODE							
		UNIT WT							
		SHEET							
		OF							

A. Encapsulation Procedure

- Trim all modules to dimensions shown on individual welded module drawings.
- Load Header, (top A layer film with module output lead locations triangles - punched), leaving headers at least 1/4 inch larger on all sides than the welded module.
- Q. C. for proper loading and header punching.
- Insert welded module leads through a 1/16" thick strip of RTV-11, or equivalent, using a strip of styrafoam for backing (See Figure 1).
- Preheat welded modules in a 70° - 80°C oven for at least 15 minutes.
- Conformal coat welded modules with sufficient epoxy (Armstrong C-7--see Preparation of Epoxy) to bridge all components and to run between all layers of film. (Note: The amount of epoxy to be used is proportional to the component density of the module.) The epoxy should be applied with Walsco Electronics No. 987 Chem-O-Jector, or equivalent. It should be noted that the epoxy should be applied as quickly as possible since the modules cool rapidly upon exposure to room temperature.
- Return modules to oven and heat at 70° - 80°C for 15 minutes (Caution: Make sure modules are in a horizontal position as shown in Figure 2. This will allow excess epoxy to run off modules.) Then turn over the module holding device, such that the module top in Figure 2 is now in the bottom position. (Note: At this point any excess epoxy may be wiped from the surface of the modules with a small mixing splint). Leave the modules in the oven for an additional 15 minutes.
- Remove the modules from the oven and let cool at room temperature for 4 or more hours before removing the modules from the RTV-11 strip and styrafoam backing material. After removal from the RTV-11 strip, trim the headers to size as indicated on the individual welded module drawings.

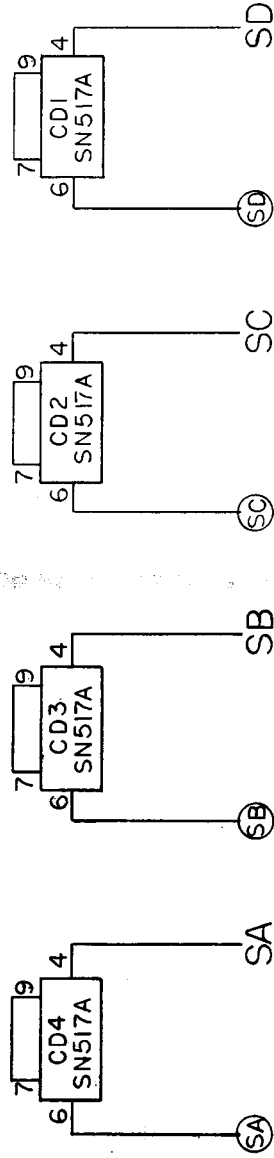
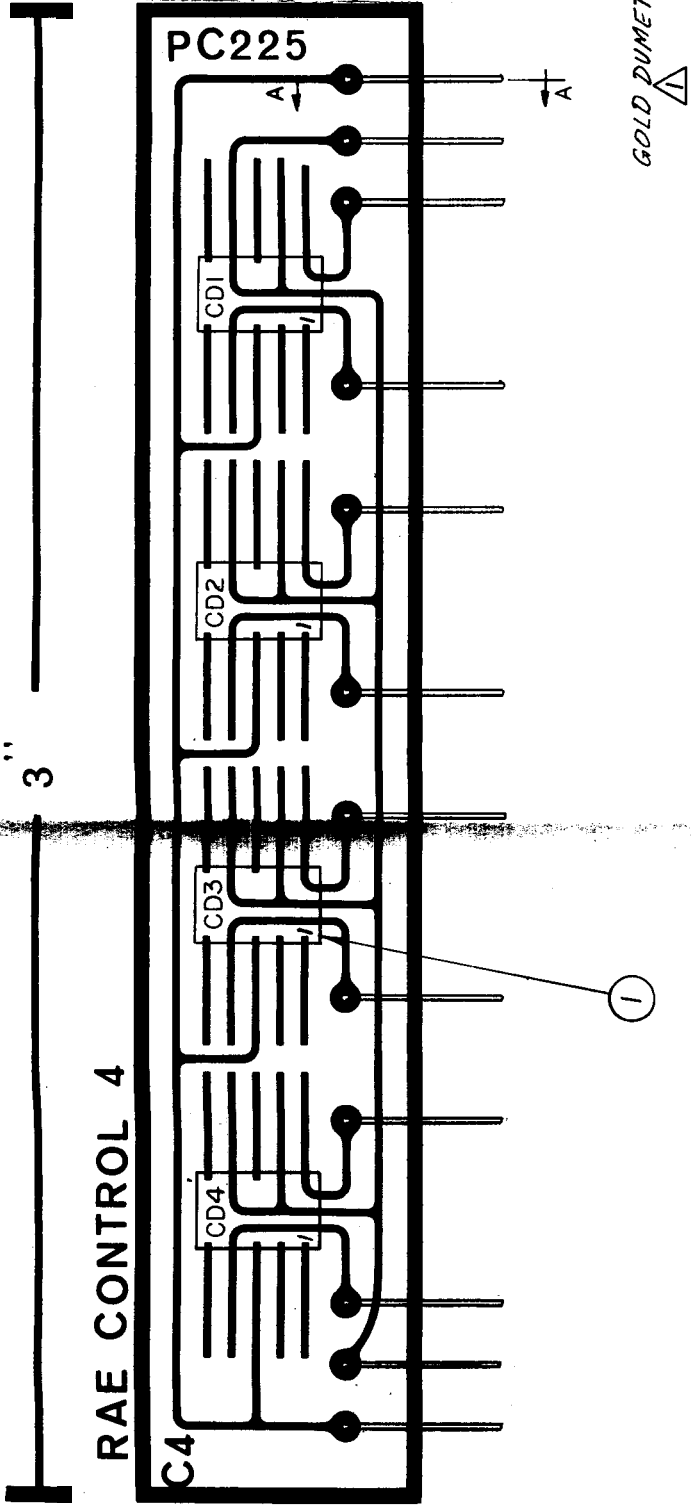
B. Preparation of Epoxy

- Materials
 - Armstrong Epoxy Resin Products (C-7) resin
 - Armstrong Epoxy Resin Products activator (W)
- Mixture
 - C-7 - - - - - 35 grams
 - Activator W - - - - - 15 grams(i. e., 7 to 3 parts by weight (Pbw) Resin to Activator respectively to allow for minimum outgassing requirements of the epoxy).
- Mix the resin and activator thoroughly for at least 5 minutes, then set aside for 15 minutes at room temperature to allow the trapped air bubbles to rise to the surface. Stir lightly at 5 minute Intervals. Then fill the syringe. The epoxy is now ready for Step (6) of Encapsulation Procedure.



MATERIAL		NAME	INIT	DATE
		DESIGNER		
		DRAWN	W T A	
		CHECKED		
		APPROVED		
		APPROVED		
DO NOT SCALE THIS DRAWING				
REMOVE BURRS, BREAK SHARP EDGES				
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES		SCALE: ~		
TOLERANCE ON FRACTIONS ± ~		FINISH: ~		
DECIMALS ± ~, ANGLES ± ~		CLIENT: N R L		
NEXT ASSY.	USED ON			
APPLICATION				

RAE		ENCAPSULATION		PROCEDURE	
NATIONAL AERONAUTICS AND SPACE ADMINISTRATION		GODDARD SPACE FLIGHT CENTER		GREENBELT, MARYLAND	
GB-RAE-1155-005		CODE	SHEET	OF	



CIRCUIT COMPONENT LIST			REMARKS
PART NO.	QTY	DESCRIPTION	
1	4	SN517A CLOCK DRIVER	T.1

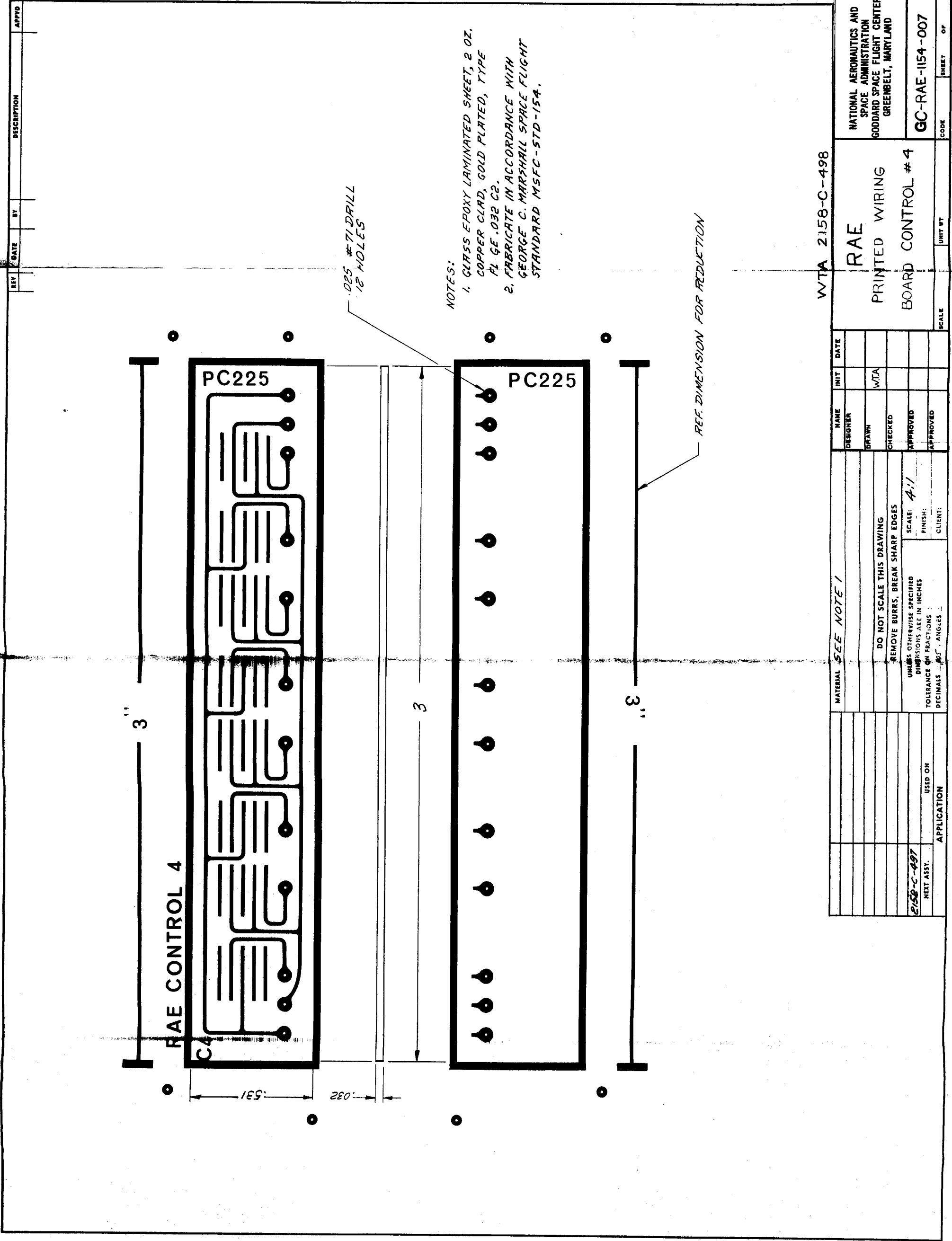
NOTES:

△ PRIOR TO COMPONENTS, INSTALL OUTPUT LEADS .020 GOLD DUMET.

2. SOLDER PER NASA QUALITY PUBLICATION NPC 200-4.

WTA 2158-C-497

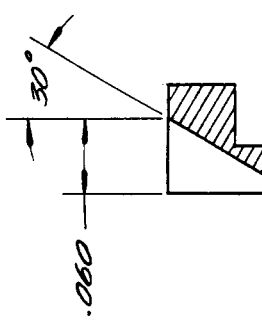
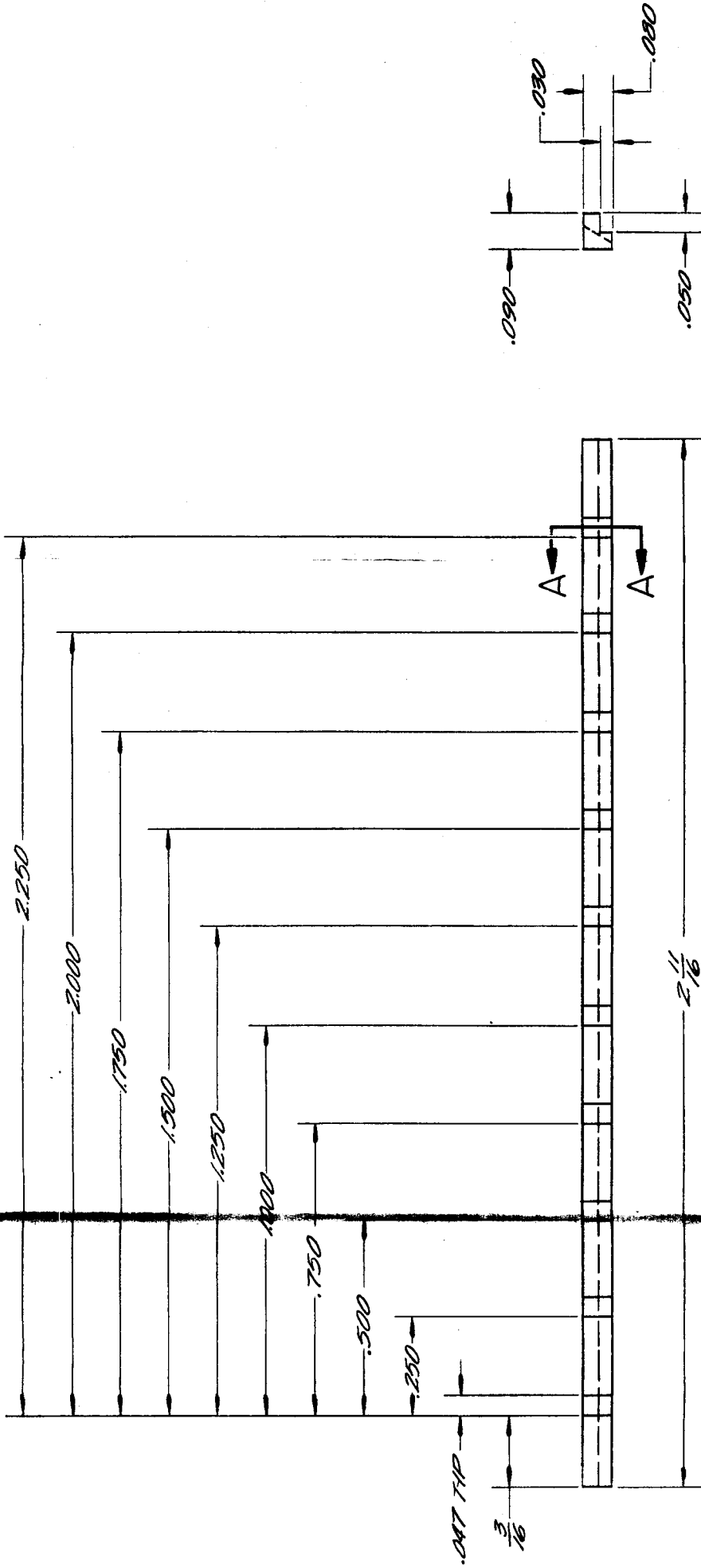
MATERIAL		NAME		INIT	DATE
DESIGNER		DRAWN		WTA	
DO NOT SCALE THIS DRAWING		CHECKED			
REMOVE BURRS, BREAK SHARP EDGES		APPROVED			
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES		SCALE: 2/1			
TOLERANCE ON FRACTIONS DECIMALS		FINISH:			
ANGLES		CLIENT:			
USED ON		APPLICATION			
NEXT ASSY.					
2158-C-501					
RAE PRINTED WIRING ASSY. CONTROL #4					
NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND					
GC-RAE-1155-006					
CODE		SHEET		OF	



WTA 2158-C-498

MATERIAL		SEE NOTE 1		NAME	INIT	DATE	RAE		NATIONAL AERONAUTICS AND SPACE ADMINISTRATION		
				DESIGNER			PRINTED WIRING		SPACE ADMINISTRATION		
				DRAWN	WTA		BOARD CONTROL #4		GODDARD SPACE FLIGHT CENTER		
				CHECKED					GREENBELT, MARYLAND		
				APPROVED					GC-RAE-1154-007		
				APPROVED					CODE		
										SHEET	
										OF	

2158-C-497	USED ON	APPLICATION	DECIMALS	ANGLES	FINISH	SCALE	CLIENT
NEXT ASSY.							



SECTION A-A
SCALE: 10
TYP 10 PLACES

WTA 2158-C-482

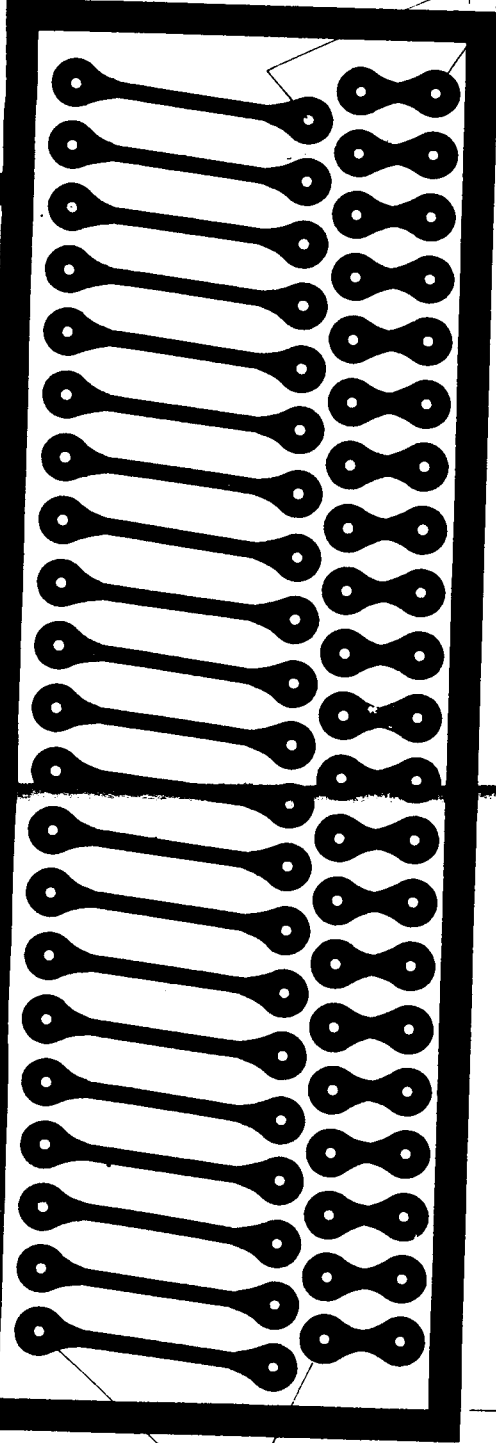
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		DESIGNER			
		DRAWN		WTA	
		CHECKED			
		SCALE: 10			
		FINISH:			
		CLIENT:			
		DO NOT SCALE THIS DRAWING			
		REMOVE BURRS, BREAK SHARP EDGES			
		UNLESS OTHERWISE SPECIFIED			
		DIMENSIONS ARE IN INCHES			
		TOLERANCE ON FRACTIONS ± 1/64			
		DECIMALS ± .005, ANGLES ± 2°			
		USED ON			
		APPLICATION			
		NEXT ASST.			
		APPROVED			
		SCALE		UNIT WT	SHEET OF
		RAE			
		RETAINING STRIP			
		NATIONAL AERONAUTICS AND			
		SPACE ADMINISTRATION			
		GODDARD SPACE FLIGHT CENTER			
		GREENBELT, MARYLAND			
		GC-RAE-1155-009			

REV	DATE	BY	DESCRIPTION	APPROV

REF DIMENSIONS FOR REDUCTION

2"

RAE TEST LIMITER PC222



.025 #71 DRILL
42 HOLES

2.737

.032

.021 #75 DRILL 42 HOLES

NOTES:

1. GLASS EPOXY LAMINATED SHEET, 2 OZ. COPPER CLAD, GOLD PLATED. TYPE FL GE.032 C2.
2. FABRICATE IN ACCORDANCE WITH GEORGE C. MARSHALL SPACE FLIGHT STANDARD MSFC-STD-154.

WTA 2158-C-496

MATERIAL SEE NOTE 1		NAME DESIGNER	INIT	DATE
		DRAWN	WTA	
DO NOT SCALE THIS DRAWING		CHECKED		
REMOVE BURRS, BREAK SHARP EDGES		APPROVED		
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES		SCALE: 4:1		
TOLERANCE ON FRACTIONS ± DECIMALS ±.005°, ANGLES ±		FINISH:		
		CLIENT:		
2158-D-495 NEXT ASSY.	USED ON	APPLICATION		

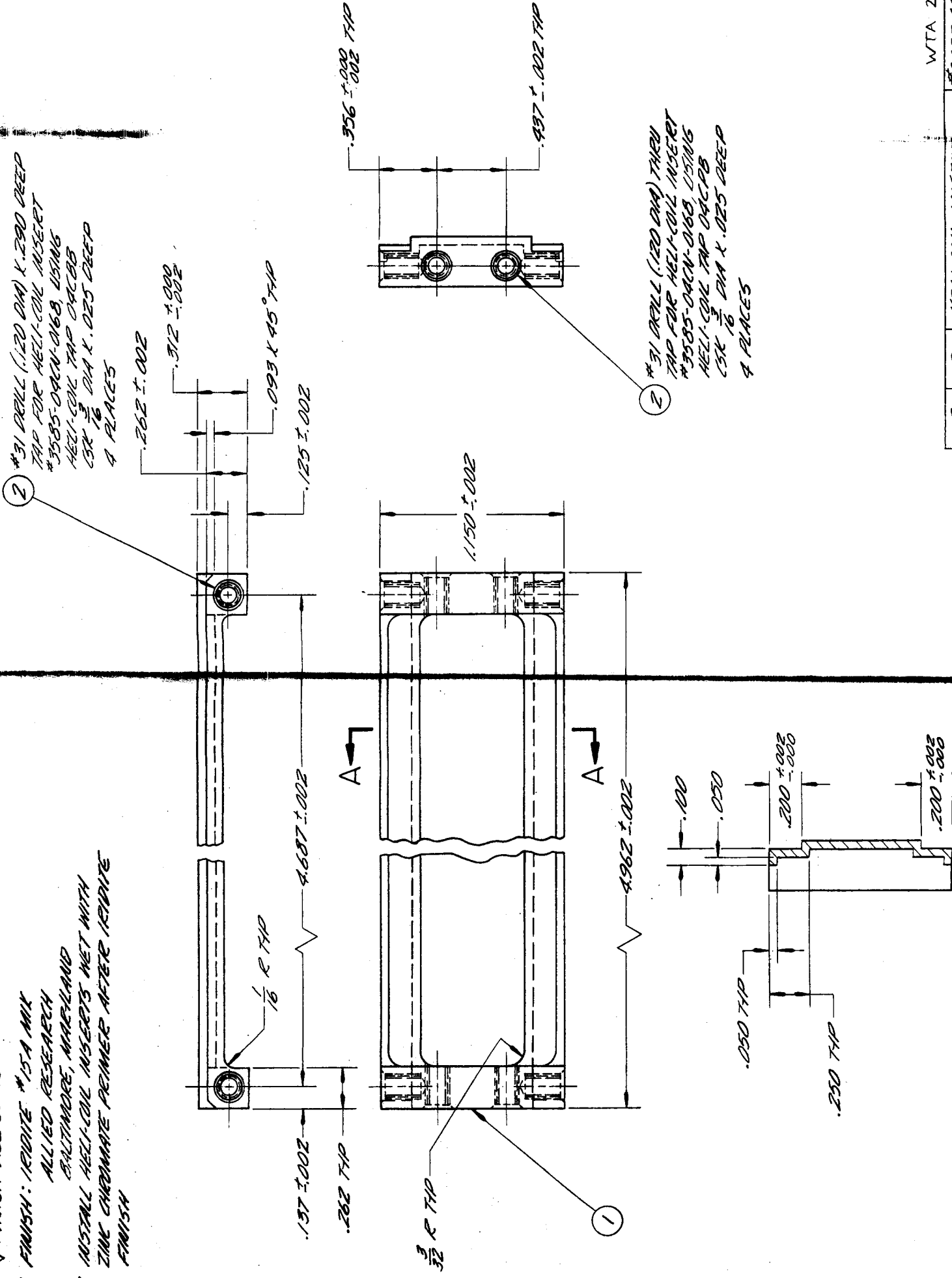
RAE PRINTED WIRING BOARD TEST LIMITER		SCALE	UNIT WT	SHEET	OF
NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		CODE	GC-RAE-1155-010		

NATIONAL AERONAUTICS AND
SPACE ADMINISTRATION
GODDARD SPACE FLIGHT CENTER
GREENBELT, MARYLAND

GC-RAE-1155-010

NOTES: UNLESS OTHERWISE SPECIFIED:

1. FINISH: IRIDITE #15A MIX
ALLIED RESEARCH
BALTIMORE, MARYLAND
2. FINISH: IRIDITE #15A MIX
ALLIED RESEARCH
BALTIMORE, MARYLAND
3. INSTALL HELI-COL INSERTS NET WITH
ZINC CHROMATE PRIMER AFTER IRIDITE
FINISH



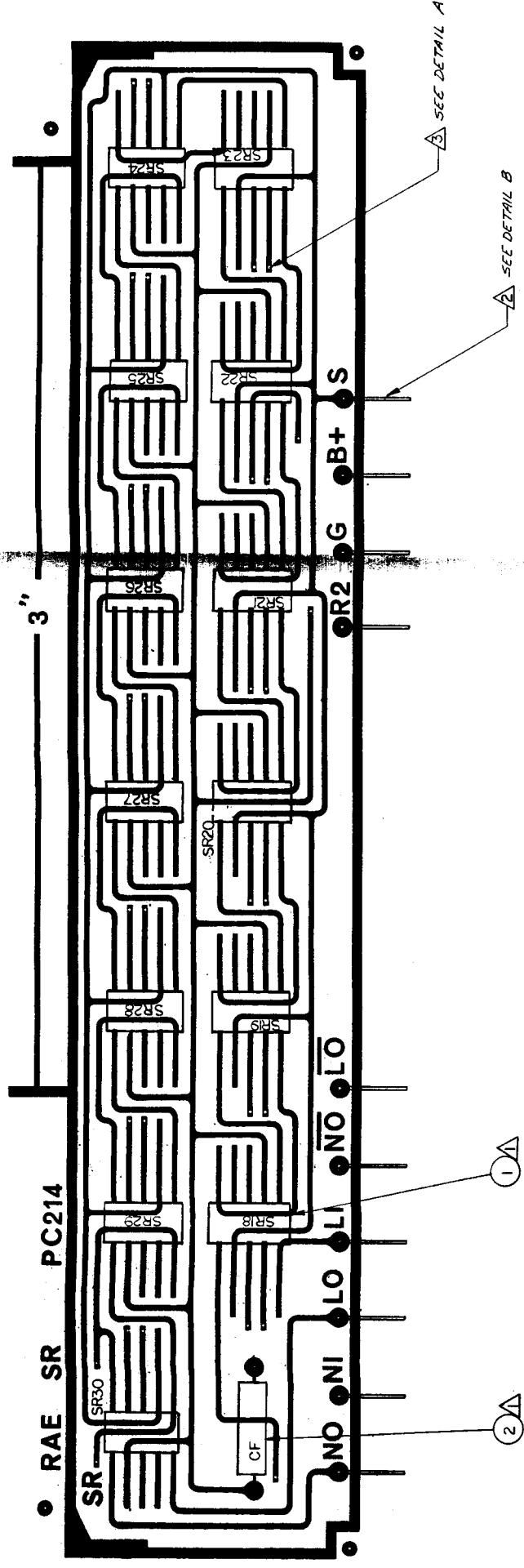
WTA 2158-C-477

ITEM NO.	DESCRIPTION	QUANTITY	UNIT	REMARKS
1	HELI-COL INSERT	4	PCS	HELI-COL INSERT
2	REAR PANEL	1	PC	MAGNESIUM 5" X 1 1/2" X 1/8" THK
3	NO. 100	1	PC	NO. 100
4	NO. 100	1	PC	NO. 100

LIST OF MATERIAL

MATERIAL	SEE LIST OF MATERIAL	NAME	INIT	DATE
		DESIGNER		
		DRAWN	WTA	
		CHECKED		
		APPROVED		
		APPROVED		
		CLIENT:		
		FINISH: IRIDITE		
		SCALE: 1/16"		
		REMOVE BURRS, BREAK SHARP EDGES		
		DO NOT SCALE THIS DRAWING		
		UNLESS OTHERWISE SPECIFIED		
		DIMENSIONS ARE IN INCHES		
		TOLERANCE ON FRACTIONS ± 1/64		
		DECIMALS ± .005, ANGLES ± 2°		
		USED ON		
		NET ASSY.		
		APPLICATION		

NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND	RAE PANEL, REAR	SCALE	UNIT WT	CODE	SHEET	OF
				G-C-RAE-1155-011		



NOTES:

△ COMPONENT DESIGNATION SHOWN FOR -1 ASSY ONLY.

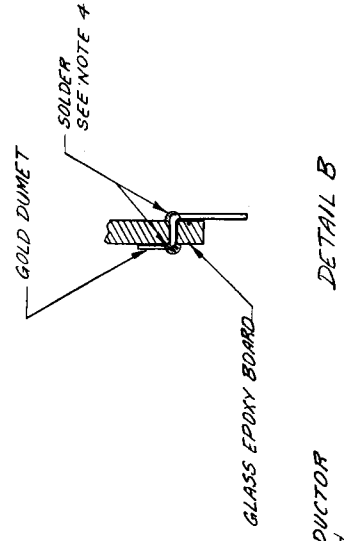
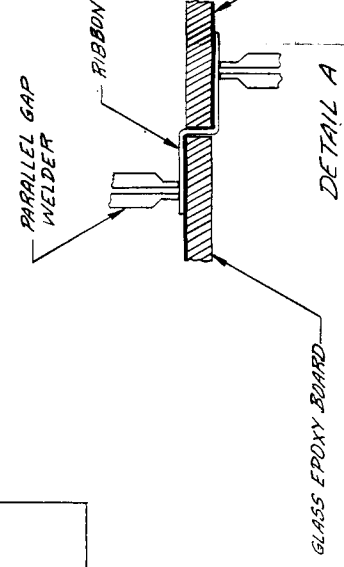
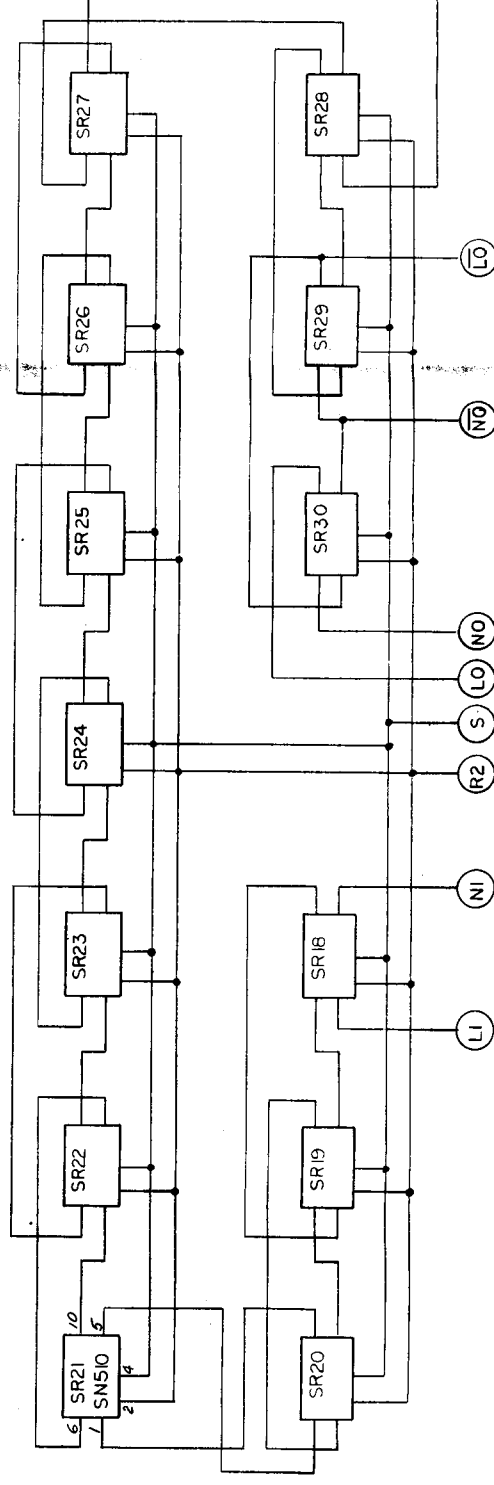
12 STEP 1. INSTALL OUTPUT LEADS .020 GOLD DUMET

3. STEP 2. WELDED FEEDTHRU TO BE INSTALLED PRIOR TO COMPONENTS, 31 PLACES. FEEDTHRU MATERIAL 0.01X0.004 FODAR FIBBON.

4. SOLDER³ PER NASA QUALITY PUBLICATION NPC 200-4.

TABLE OF ASSEMBLIES					
FIND NO.	REF. DESIGNATIONS				
	-1	-2	-3	-4	
1	SR18 THRU SR20	SR31 THRU SR43	SR44 THRU SR56	SR57 THRU SR68	
2	CF	CF	CF	CF	

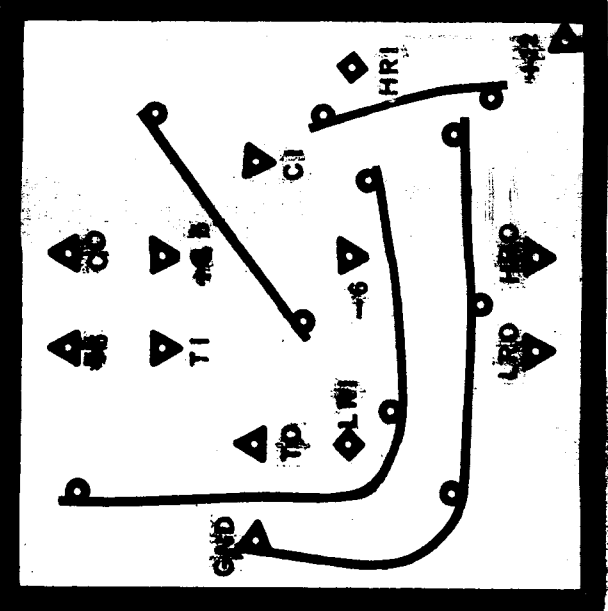
CIRCUIT COMPONENT LIST			REMARKS
FIND NO.	NO. REQ'D	DESCRIPTION	
1	13	SINCO R-5 FLIP-FLOP COUNTER	T.I.
2	1	CAPACITOR 1MFD 350V 15V	SPRAGUE



WTA 2158-D-491

[illegible]

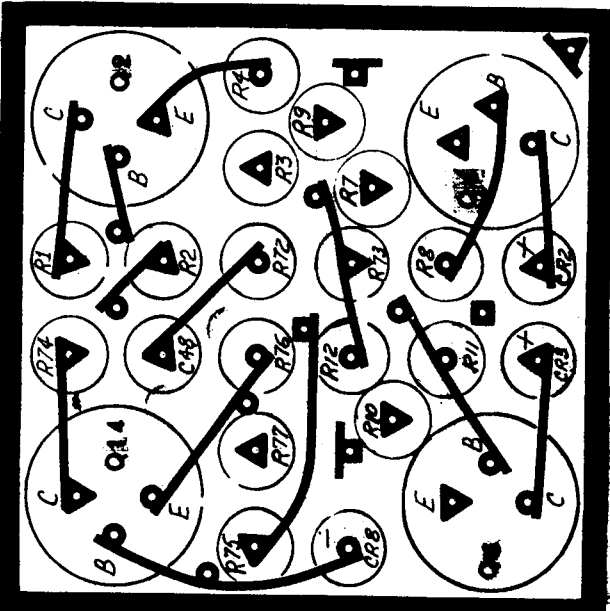
T&R DRIVER



WM95A

A2

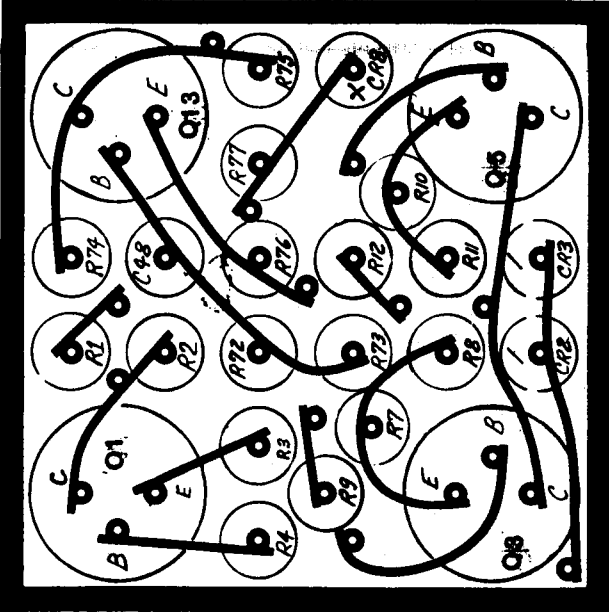
T&R DRIVER



WM95A

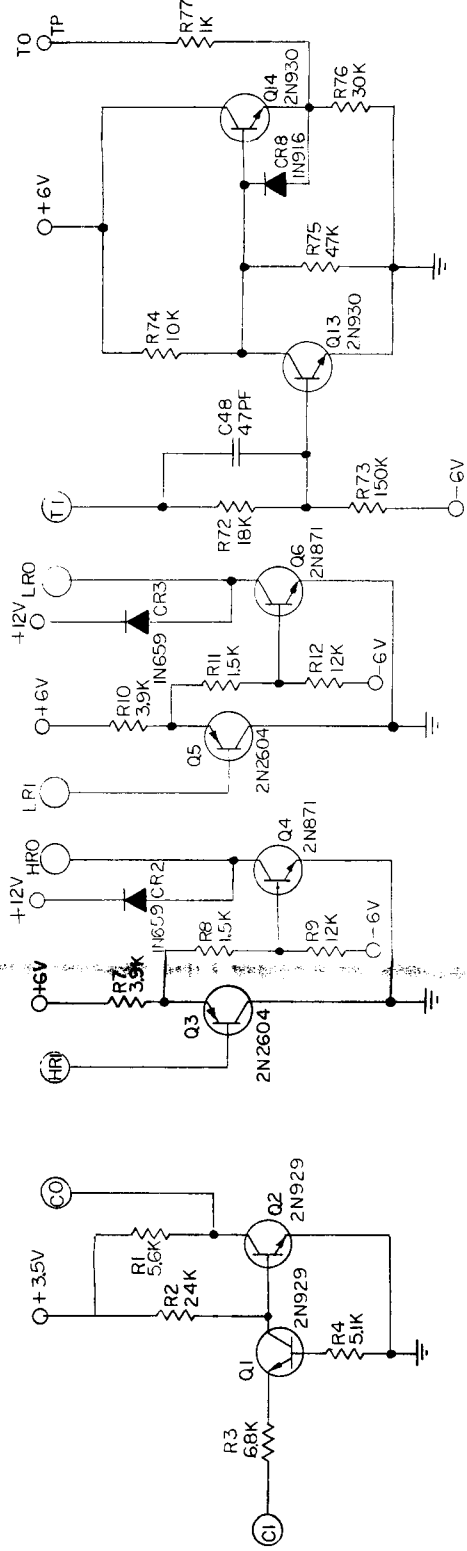
A1

T&R DRIVER

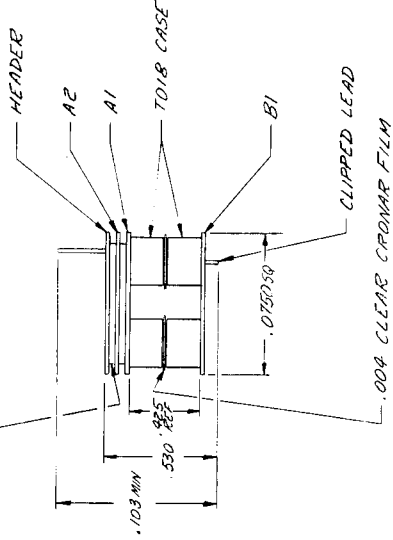


WM95A

B1



INTERCONNECTION WIRES, .010 X .020 GRADE A
SOFT NICKEL



UNENCAPSULATED MODULE CONFIGURATION

SCALE: NONE

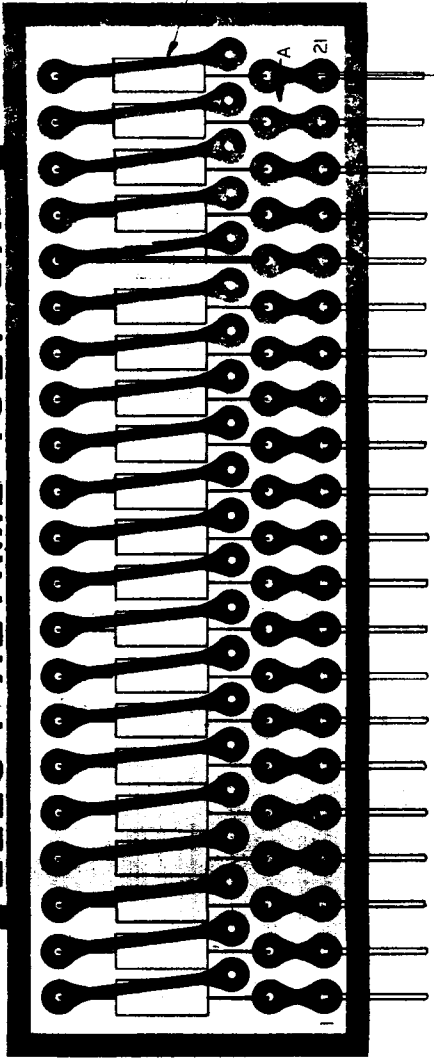
FOR ENCAPSULATION PROCEDURE SEE DWG NO. 2158-B-528

Q14	TRANSISTOR	2N930	T.1
Q13	↑	2N930	T.1
Q6		2N871	FAIRCHILD
Q5		2N2604	T.1
Q4		2N871	FAIRCHILD
Q3	↑	2N2604	T.1
Q2		2N929	T.1
Q1	TRANSISTOR	2N929	T.1
CR8	DIODE	1N916A	T.1
CR3	DIODE	1N916A	T.1
CR2	DIODE	1N916A	T.1
C48	CAPACITOR	47PF MC80	AERONOX
R77	RESISTOR	1K 1/4W ± 5%	WELWYN
R76	↑	30K	↑
R75		47K	↑
R74		10K	↑
R73		150K	↑
R72		18K	↑
R71		12K	↑
R11		15K	↑
R10		3.9K	↑
R9		12K	↑
R8		15K	↑
R7		3.9K	↑
R4		51K	↑
R3		6.8K	↑
R2	↑	24K	↑
R1	RESISTOR	5.6K 1/4W ± 5%	WELWYN
DESIGN	DESCRIPTION	REMARKS	
LIST OF MATERIAL WTA 2158-D-577			

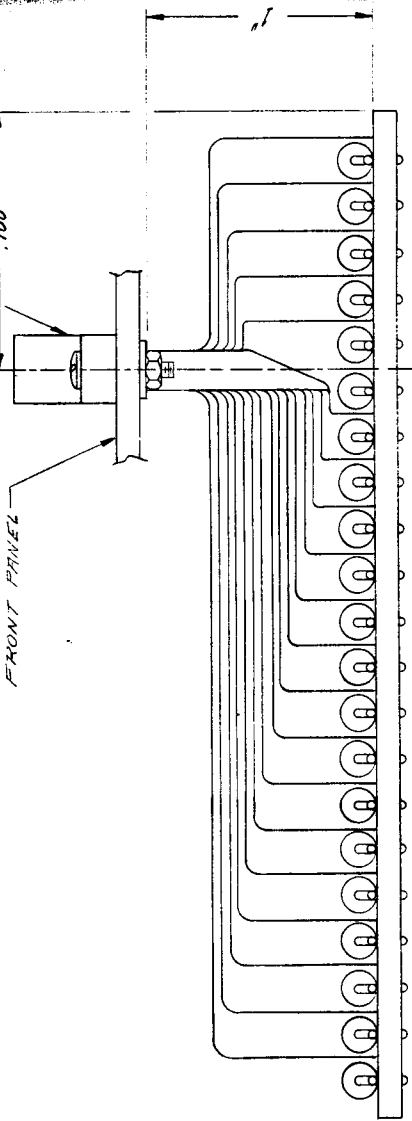
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5"

RAE TEST LIMITER BCSS5



SECTION A-A



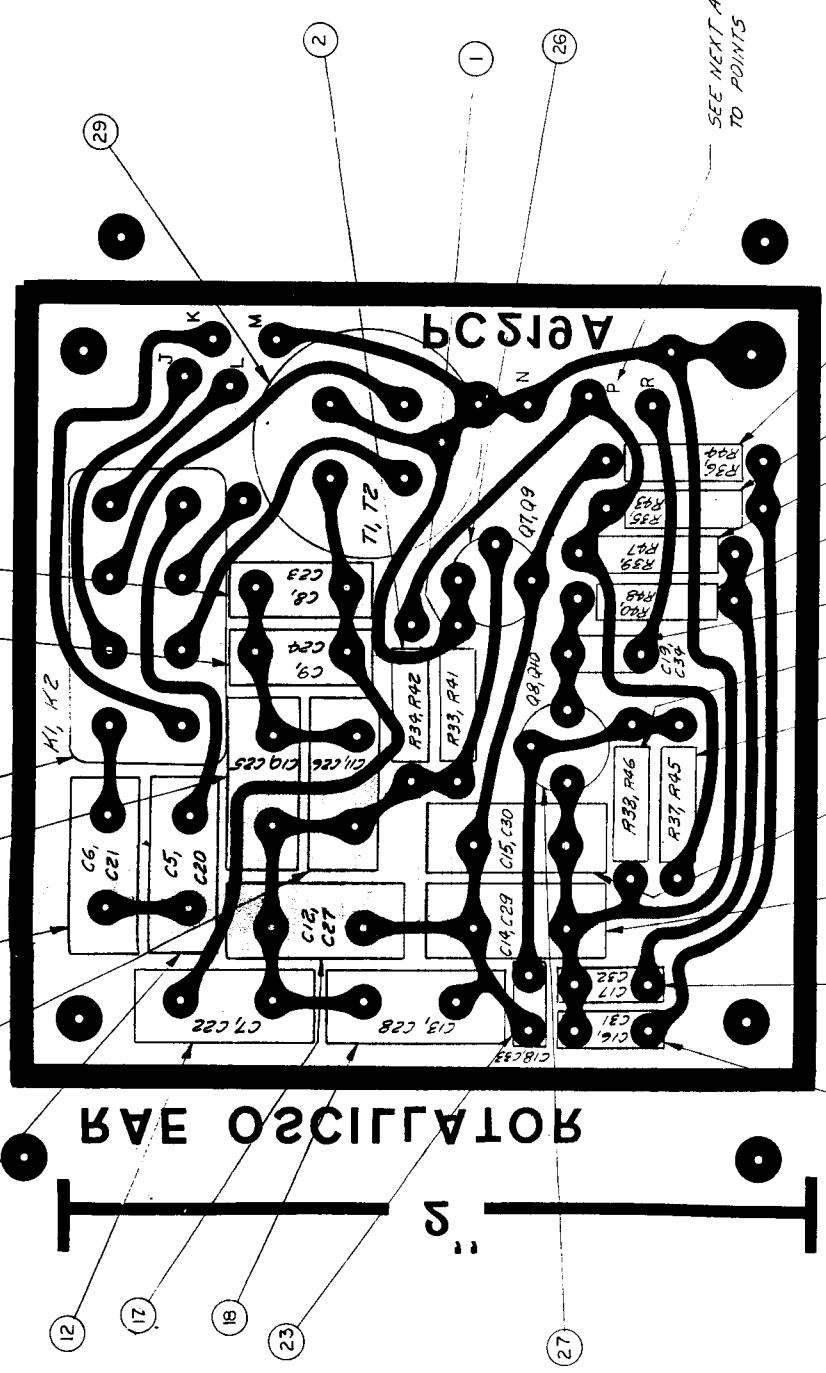
CIRCUIT COMPONENT LIST			REMARKS
FIND NO.	REQD	DESCRIPTION	
1	21	RESISTOR 43K 1/4W ±5%	WELNXX
2	1	CONNECTOR WIREDS MDS1-21SHDD1	CANNON

NOTES:
1. SOLDER PER MIL-STD-883C QUALITY PUBLICATION
NPC 200-4.

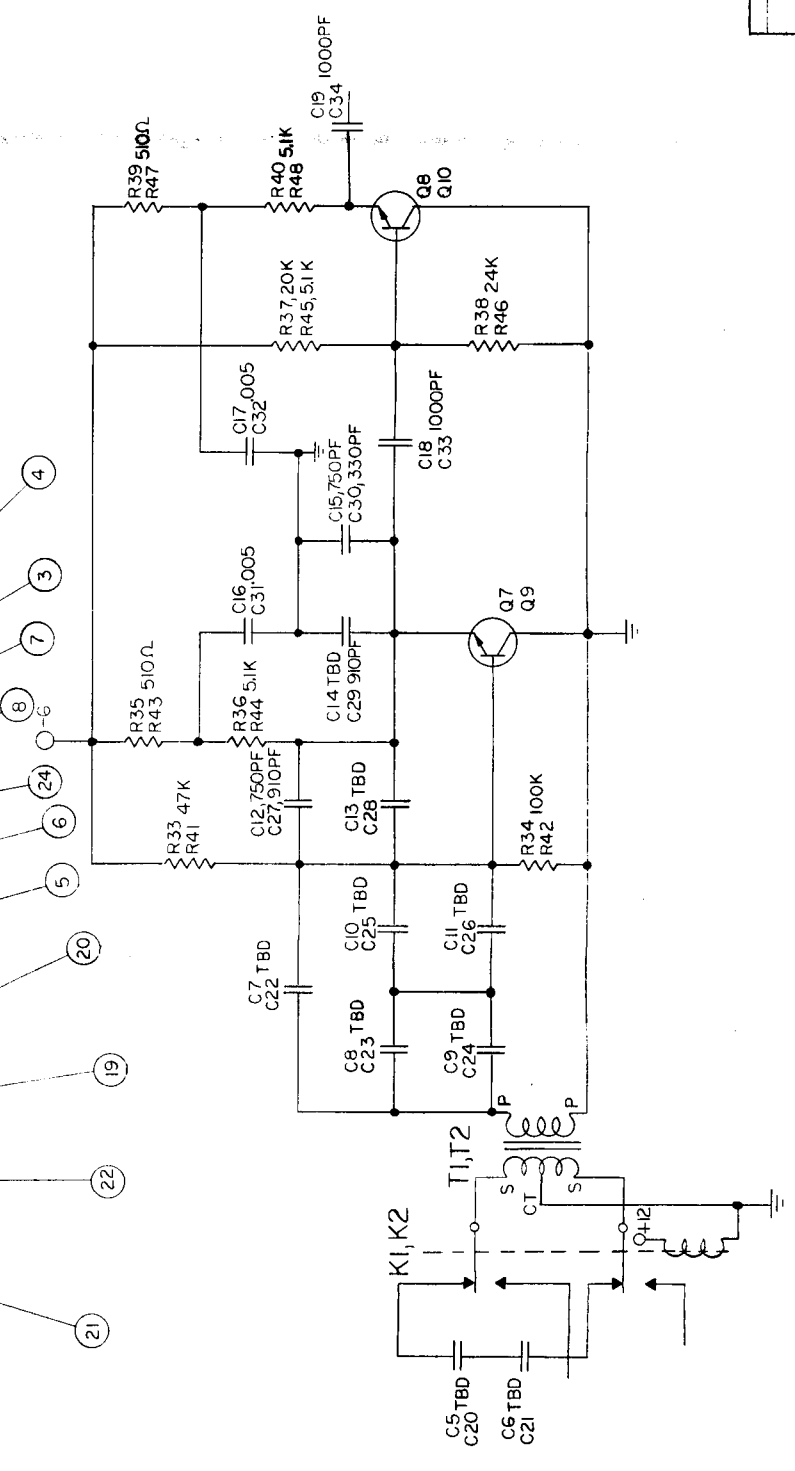
WTA 2158-D-495

NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		RAE PRINTED WIRING ASSY. TEST LIMITER		WTA 2158-D-495		CODE UNIT WT SCALE		SHEET 1 OF 1	
MATERIAL		NAME DESIGNER WTA CHECKED APPROVED		INIT DATE		DATE		DATE	
DO NOT SCALE THIS DRAWING REMOVE BURRS, BREAK SHARP EDGES UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCE ON FRACTIONS ± DECIMALS ±.005, ANGLES ±		SCALE: 4:1 FINISH: CLIENT:							
USED ON NEXT ASSY.		APPLICATION							

NOTES:
1. THIS DRAWING APPLIES FOR BOTH OSCILLATOR I & 2.
2. CHECK LIST OF COMPONENTS FOR VALUE DIFFERENCES.
3. SEE NEXT ASSEMBLY FOR WIRE-BOARD WIRING.
3. SOLDER PER NASA QUALITY PUBLICATION NPC-200-4



SEE NEXT ASSY FOR CONNECTION TO POINTS J, K, L, M, N, P, R.



-I-				-II-			
LIST OF COMPONENTS				LIST OF COMPONENTS			
FOR BOTH REQD	USED IN HIGH ASSC	DESCRIPTION	USED IN LOW ASSC	DESCRIPTION	REMARKS FOR BOTH		
1	R33	RESISTOR, 47K 1/4W 15%	R41	RESISTOR, 47K 1/4W 15%	WELNIN		
2	R34	RESISTOR, 100K 1/4W 15%	R42	RESISTOR, 100K 1/4W 15%	WELNIN		
3	R35	RESISTOR, 51K 1/4W 15%	R43	RESISTOR, 51K 1/4W 15%	WELNIN		
4	R36	RESISTOR, 20K 1/4W 15%	R44	RESISTOR, 20K 1/4W 15%	WELNIN		
5	R37	RESISTOR, 24K 1/4W 15%	R45	RESISTOR, 24K 1/4W 15%	WELNIN		
6	R38	RESISTOR, 51K 1/4W 15%	R46	RESISTOR, 51K 1/4W 15%	WELNIN		
7	R39	RESISTOR, 51K 1/4W 15%	R47	RESISTOR, 51K 1/4W 15%	WELNIN		
8	R40	RESISTOR, 51K 1/4W 15%	R48	RESISTOR, 51K 1/4W 15%	WELNIN		
9	C5	CAPACITOR, TO BE DETERM.	C20	CAPACITOR, TO BE DETERM.			
10	C6	CAPACITOR, TO BE DETERM.	C21	CAPACITOR, TO BE DETERM.			
11	C7	CAPACITOR, TO BE DETERM.	C22	CAPACITOR, TO BE DETERM.			
12	C8	CAPACITOR, TO BE DETERM.	C23	CAPACITOR, TO BE DETERM.			
13	C9	CAPACITOR, TO BE DETERM.	C24	CAPACITOR, TO BE DETERM.			
14	C10	CAPACITOR, TO BE DETERM.	C25	CAPACITOR, TO BE DETERM.			
15	C11	CAPACITOR, TO BE DETERM.	C26	CAPACITOR, TO BE DETERM.			
16	C12	CAPACITOR, TO BE DETERM.	C27	CAPACITOR, TO BE DETERM.			
17	C13	CAPACITOR, TO BE DETERM.	C28	CAPACITOR, TO BE DETERM.			
18	C14	CAPACITOR, TO BE DETERM.	C29	CAPACITOR, TO BE DETERM.			
19	C15	CAPACITOR, TO BE DETERM.	C30	CAPACITOR, TO BE DETERM.			
20	C16	CAPACITOR, TO BE DETERM.	C31	CAPACITOR, TO BE DETERM.			
21	C17	CAPACITOR, TO BE DETERM.	C32	CAPACITOR, TO BE DETERM.			
22	C18	CAPACITOR, TO BE DETERM.	C33	CAPACITOR, TO BE DETERM.			
23	C19	CAPACITOR, TO BE DETERM.	C34	CAPACITOR, TO BE DETERM.			
24	C20	CAPACITOR, TO BE DETERM.	C35	CAPACITOR, TO BE DETERM.			
25	C21	CAPACITOR, TO BE DETERM.	C36	CAPACITOR, TO BE DETERM.			
26	C22	CAPACITOR, TO BE DETERM.	C37	CAPACITOR, TO BE DETERM.			
27	C23	CAPACITOR, TO BE DETERM.	C38	CAPACITOR, TO BE DETERM.			
28	C24	CAPACITOR, TO BE DETERM.	C39	CAPACITOR, TO BE DETERM.			
29	C25	CAPACITOR, TO BE DETERM.	C40	CAPACITOR, TO BE DETERM.			
30	C26	CAPACITOR, TO BE DETERM.	C41	CAPACITOR, TO BE DETERM.			
31	C27	CAPACITOR, TO BE DETERM.	C42	CAPACITOR, TO BE DETERM.			
32	C28	CAPACITOR, TO BE DETERM.	C43	CAPACITOR, TO BE DETERM.			
33	C29	CAPACITOR, TO BE DETERM.	C44	CAPACITOR, TO BE DETERM.			
34	C30	CAPACITOR, TO BE DETERM.	C45	CAPACITOR, TO BE DETERM.			
35	C31	CAPACITOR, TO BE DETERM.	C46	CAPACITOR, TO BE DETERM.			
36	C32	CAPACITOR, TO BE DETERM.	C47	CAPACITOR, TO BE DETERM.			
37	C33	CAPACITOR, TO BE DETERM.	C48	CAPACITOR, TO BE DETERM.			
38	C34	CAPACITOR, TO BE DETERM.	C49	CAPACITOR, TO BE DETERM.			
39	C35	CAPACITOR, TO BE DETERM.	C50	CAPACITOR, TO BE DETERM.			
40	C36	CAPACITOR, TO BE DETERM.	C51	CAPACITOR, TO BE DETERM.			
41	C37	CAPACITOR, TO BE DETERM.	C52	CAPACITOR, TO BE DETERM.			
42	C38	CAPACITOR, TO BE DETERM.	C53	CAPACITOR, TO BE DETERM.			
43	C39	CAPACITOR, TO BE DETERM.	C54	CAPACITOR, TO BE DETERM.			
44	C40	CAPACITOR, TO BE DETERM.	C55	CAPACITOR, TO BE DETERM.			
45	C41	CAPACITOR, TO BE DETERM.	C56	CAPACITOR, TO BE DETERM.			
46	C42	CAPACITOR, TO BE DETERM.	C57	CAPACITOR, TO BE DETERM.			
47	C43	CAPACITOR, TO BE DETERM.	C58	CAPACITOR, TO BE DETERM.			
48	C44	CAPACITOR, TO BE DETERM.	C59	CAPACITOR, TO BE DETERM.			
49	C45	CAPACITOR, TO BE DETERM.	C60	CAPACITOR, TO BE DETERM.			
50	C46	CAPACITOR, TO BE DETERM.	C61	CAPACITOR, TO BE DETERM.			
51	C47	CAPACITOR, TO BE DETERM.	C62	CAPACITOR, TO BE DETERM.			
52	C48	CAPACITOR, TO BE DETERM.	C63	CAPACITOR, TO BE DETERM.			
53	C49	CAPACITOR, TO BE DETERM.	C64	CAPACITOR, TO BE DETERM.			
54	C50	CAPACITOR, TO BE DETERM.	C65	CAPACITOR, TO BE DETERM.			
55	C51	CAPACITOR, TO BE DETERM.	C66	CAPACITOR, TO BE DETERM.			
56	C52	CAPACITOR, TO BE DETERM.	C67	CAPACITOR, TO BE DETERM.			
57	C53	CAPACITOR, TO BE DETERM.	C68	CAPACITOR, TO BE DETERM.			
58	C54	CAPACITOR, TO BE DETERM.	C69	CAPACITOR, TO BE DETERM.			
59	C55	CAPACITOR, TO BE DETERM.	C70	CAPACITOR, TO BE DETERM.			
60	C56	CAPACITOR, TO BE DETERM.	C71	CAPACITOR, TO BE DETERM.			
61	C57	CAPACITOR, TO BE DETERM.	C72	CAPACITOR, TO BE DETERM.			
62	C58	CAPACITOR, TO BE DETERM.	C73	CAPACITOR, TO BE DETERM.			
63	C59	CAPACITOR, TO BE DETERM.	C74	CAPACITOR, TO BE DETERM.			
64	C60	CAPACITOR, TO BE DETERM.	C75	CAPACITOR, TO BE DETERM.			
65	C61	CAPACITOR, TO BE DETERM.	C76	CAPACITOR, TO BE DETERM.			
66	C62	CAPACITOR, TO BE DETERM.	C77	CAPACITOR, TO BE DETERM.			
67	C63	CAPACITOR, TO BE DETERM.	C78	CAPACITOR, TO BE DETERM.			
68	C64	CAPACITOR, TO BE DETERM.	C79	CAPACITOR, TO BE DETERM.			
69	C65	CAPACITOR, TO BE DETERM.	C80	CAPACITOR, TO BE DETERM.			
70	C66	CAPACITOR, TO BE DETERM.	C81	CAPACITOR, TO BE DETERM.			
71	C67	CAPACITOR, TO BE DETERM.	C82	CAPACITOR, TO BE DETERM.			
72	C68	CAPACITOR, TO BE DETERM.	C83	CAPACITOR, TO BE DETERM.			
73	C69	CAPACITOR, TO BE DETERM.	C84	CAPACITOR, TO BE DETERM.			
74	C70	CAPACITOR, TO BE DETERM.	C85	CAPACITOR, TO BE DETERM.			
75	C71	CAPACITOR, TO BE DETERM.	C86	CAPACITOR, TO BE DETERM.			
76	C72	CAPACITOR, TO BE DETERM.	C87	CAPACITOR, TO BE DETERM.			
77	C73	CAPACITOR, TO BE DETERM.	C88	CAPACITOR, TO BE DETERM.			
78	C74	CAPACITOR, TO BE DETERM.	C89	CAPACITOR, TO BE DETERM.			
79	C75	CAPACITOR, TO BE DETERM.	C90	CAPACITOR, TO BE DETERM.			
80	C76	CAPACITOR, TO BE DETERM.	C91	CAPACITOR, TO BE DETERM.			
81	C77	CAPACITOR, TO BE DETERM.	C92	CAPACITOR, TO BE DETERM.			
82	C78	CAPACITOR, TO BE DETERM.	C93	CAPACITOR, TO BE DETERM.			
83	C79	CAPACITOR, TO BE DETERM.	C94	CAPACITOR, TO BE DETERM.			
84	C80	CAPACITOR, TO BE DETERM.	C95	CAPACITOR, TO BE DETERM.			
85	C81	CAPACITOR, TO BE DETERM.	C96	CAPACITOR, TO BE DETERM.			
86	C82	CAPACITOR, TO BE DETERM.	C97	CAPACITOR, TO BE DETERM.			
87	C83	CAPACITOR, TO BE DETERM.	C98	CAPACITOR, TO BE DETERM.			
88	C84	CAPACITOR, TO BE DETERM.	C99	CAPACITOR, TO BE DETERM.			
89	C85	CAPACITOR, TO BE DETERM.	C100	CAPACITOR, TO BE DETERM.			

WT 2158-D-506

RAE
PRINTED WIRING ASSY
RAE OSCILLATOR

NATIONAL AERONAUTICS AND SPACE ADMINISTRATION
GODDARD SPACE FLIGHT CENTER
GREENBELT, MARYLAND

GD-RAE-1155-205

SCALE: 4:1

UNLESS OTHERWISE SPECIFIED
DIMENSIONS ARE IN INCHES
TOLERANCE ON FRACTIONS:
DECIMALS ANGLES

DO NOT SCALE THIS DRAWING
REMOVE BURRS, BREAK SHARP EDGES

SCALE: 4:1
FINISH:
TOLERANCE ON FRACTIONS:
DECIMALS ANGLES

CLIENT:

APPROVED

DATE

NAME

INIT

DATE

SCALE

UNIT

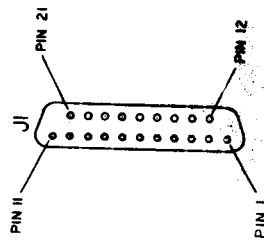
BY

DATE

DESCRIPTION

APPD

FRONT PANEL TEST CONNECTOR



J2

DATA OUT

400 CPS CLOCK

TEMP MONITOR

CHASSIS GND

POWER GND

SHIFT COMMAND

SHIELD GND

-6V

+3.5V TP

+12V

+6V

READ COMMAND

J3

A1

A2

J4

RELAY READ COMMAND

+3.5V

TEMP MONITOR

DATA OUT

END

-6V

SHIFT PULSE

SHIFT COMMAND

(H) HIGH RELAY

RL

LOW RELAY

CO-CS3

CO-CS2

CO-CS1

74

CE

+12V

-6V

NO-CS3

NO-CS2

NO-CS1

2K

2K

2K

2K

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2K

2K

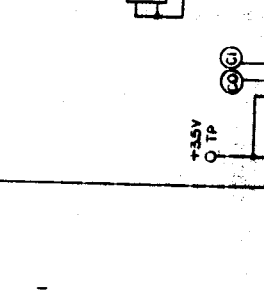
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J2

DATA OUT

400 CPS CLOCK

TEMP MONITOR

CHASSIS GND

POWER GND

SHIFT COMMAND

SHIELD GND

-6V

+3.5V TP

+12V

+6V

READ COMMAND

J3

A1

A2

J4

RELAY READ COMMAND

+3.5V

TEMP MONITOR

DATA OUT

END

-6V

SHIFT PULSE

SHIFT COMMAND

(H) HIGH RELAY

RL

LOW RELAY

CO-CS3

CO-CS2

CO-CS1

74

CE

+12V

-6V

NO-CS3

NO-CS2

NO-CS1

2K

2K

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2K

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2K

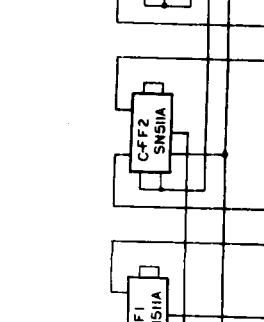
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J2

DATA OUT

400 CPS CLOCK

TEMP MONITOR

CHASSIS GND

POWER GND

SHIFT COMMAND

SHIELD GND

-6V

+3.5V TP

+12V

+6V

READ COMMAND

J3

A1

A2

J4

RELAY READ COMMAND

+3.5V

TEMP MONITOR

DATA OUT

END

-6V

SHIFT PULSE

SHIFT COMMAND

(H) HIGH RELAY

RL

LOW RELAY

CO-CS3

CO-CS2

CO-CS1

74

CE

+12V

-6V

NO-CS3

NO-CS2

NO-CS1

2K

2K

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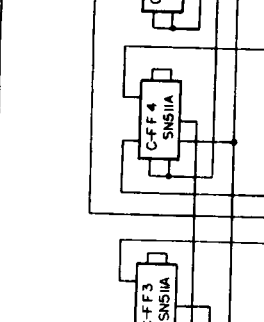
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J2

DATA OUT

400 CPS CLOCK

TEMP MONITOR

CHASSIS GND

POWER GND

SHIFT COMMAND

SHIELD GND

-6V

+3.5V TP

+12V

+6V

READ COMMAND

J3

A1

A2

J4

RELAY READ COMMAND

+3.5V

TEMP MONITOR

DATA OUT

END

-6V

SHIFT PULSE

SHIFT COMMAND

(H) HIGH RELAY

RL

LOW RELAY

CO-CS3

CO-CS2

CO-CS1

74

CE

+12V

-6V

NO-CS3

NO-CS2

NO-CS1

2K

2K

2K

2K

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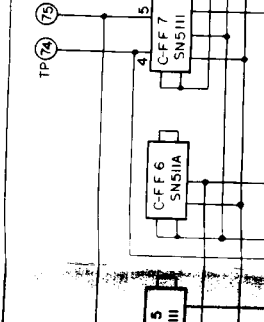
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2K



J2

DATA OUT

400 CPS CLOCK

TEMP MONITOR

CHASSIS GND

POWER GND

SHIFT COMMAND

SHIELD GND

-6V

+3.5V TP

+12V

+6V

READ COMMAND

J3

A1

A2

J4

RELAY READ COMMAND

+3.5V

TEMP MONITOR

DATA OUT

END

-6V

SHIFT PULSE

SHIFT COMMAND

(H) HIGH RELAY

RL

LOW RELAY

CO-CS3

CO-CS2

CO-CS1

74

CE

+12V

-6V

NO-CS3

NO-CS2

NO-CS1

2K

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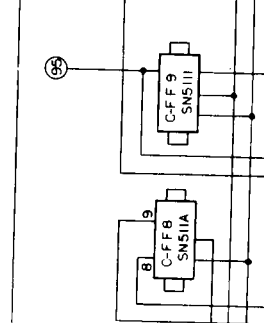
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J2

DATA OUT

400 CPS CLOCK

TEMP MONITOR

CHASSIS GND

POWER GND

SHIFT COMMAND

SHIELD GND

-6V

+3.5V TP

+12V

+6V

READ COMMAND

J3

A1

A2

J4

RELAY READ COMMAND

+3.5V

TEMP MONITOR

DATA OUT

END

-6V

SHIFT PULSE

SHIFT COMMAND

(H) HIGH RELAY

RL

LOW RELAY

CO-CS3

CO-CS2

CO-CS1

74

CE

+12V

-6V

NO-CS3

NO-CS2

NO-CS1

2K

2K

2K

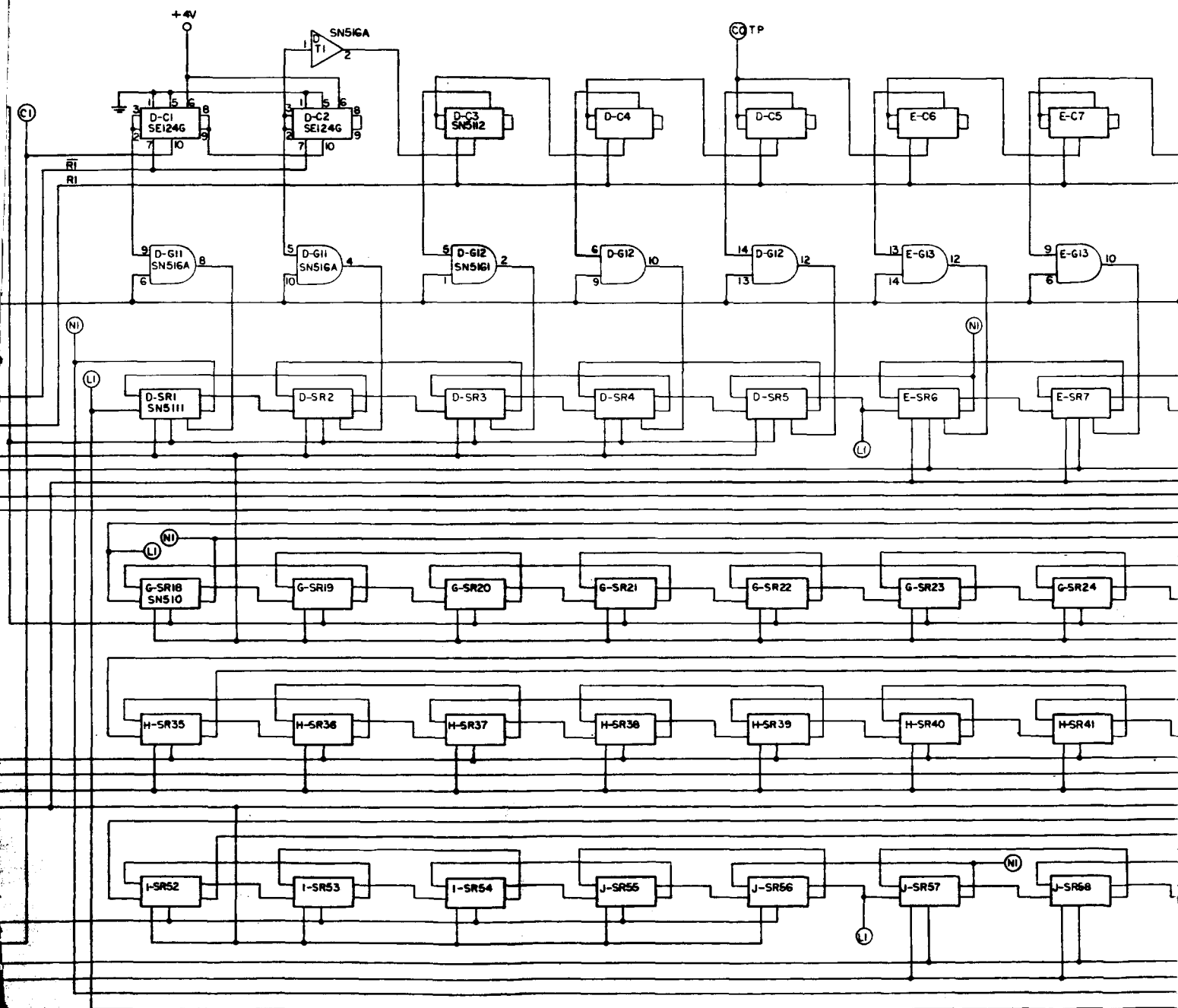
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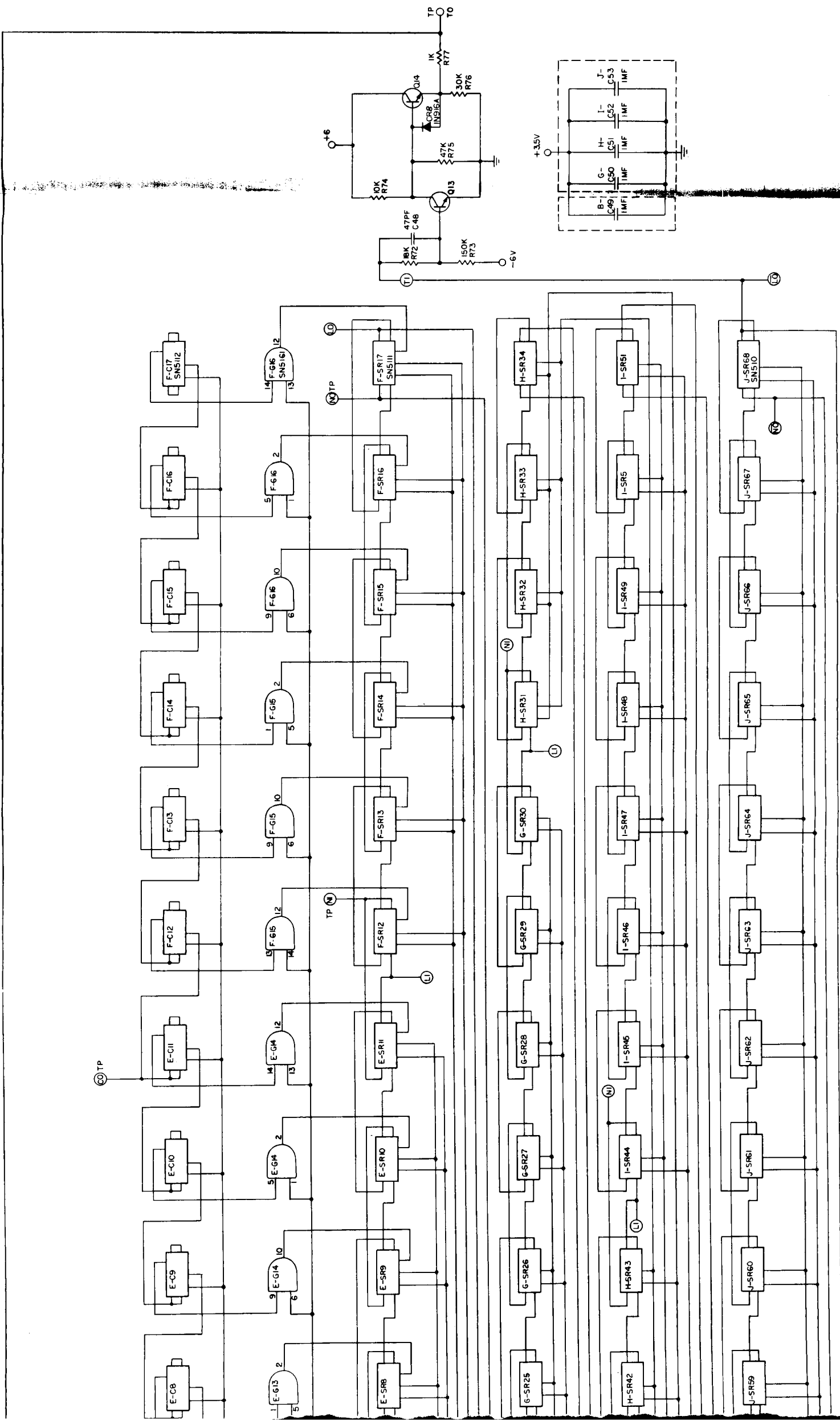
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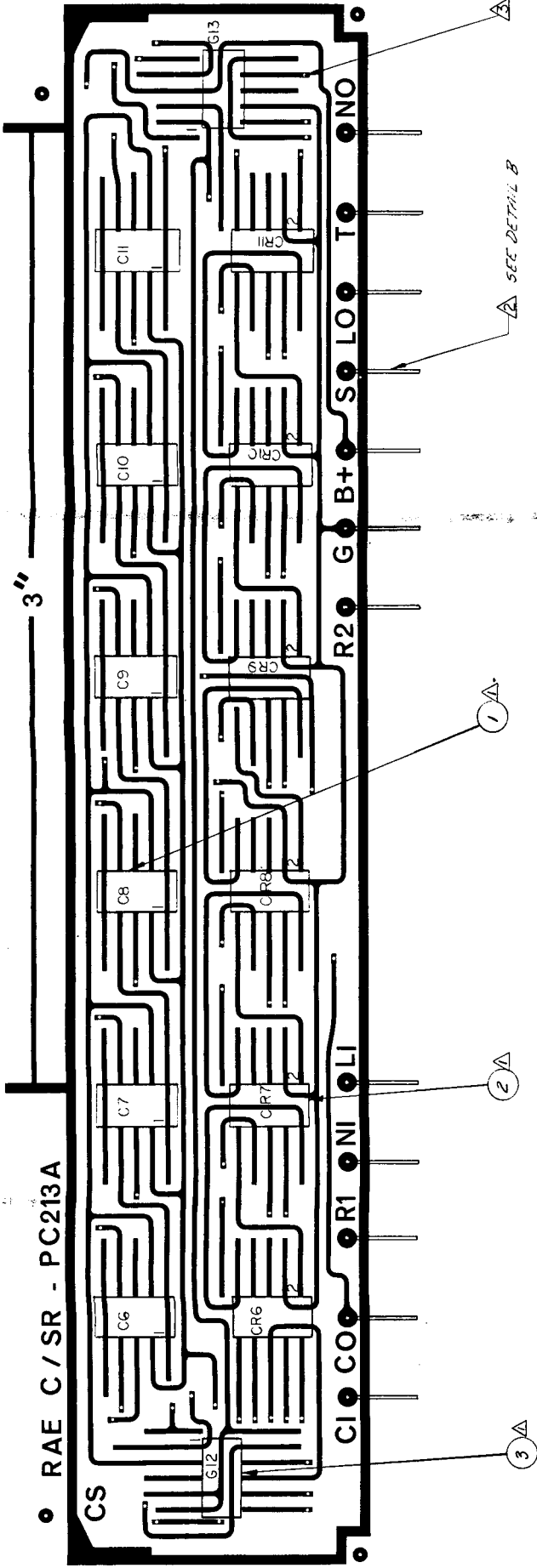




NOTES:
 1. ALL MEINW RESISTORS ARE THE RES. LEADS
 1. GOLD PLATED DUMET .025 DIA
 2. COMPONENTS LISTED FOR REFERENCE ONLY

QTY	REF	DESCRIPTION	VALUE	UNIT	REMARKS
70	2	TRANSFORMER	150-2-551	WTA	
69	2	RELAY	BRDTEC/PCAS	FILTRON	
68	2	RESISTOR	500K 1/2W 1% DM	IRC	
67	1	RES	100K 1/2W 1% DM	IRC	
66	2	RES	100K 1/2W 1% DM	MEINW	
65	3	RES	10K 1/2W 1% DM	MEINW	
64	20	RES	10K 1/2W 1% DM	MEINW	
63	12	RES	10K 1/2W 1% DM	MEINW	
62	3	RES	10K 1/2W 1% DM	MEINW	
61	2	RES	10K 1/2W 1% DM	MEINW	
60	3	RES	10K 1/2W 1% DM	MEINW	
59	2	RES	10K 1/2W 1% DM	MEINW	
58	7	RES	10K 1/2W 1% DM	MEINW	
57	1	RES	10K 1/2W 1% DM	MEINW	
56	2	RES	10K 1/2W 1% DM	MEINW	
55	2	RES	10K 1/2W 1% DM	MEINW	
54	2	RES	10K 1/2W 1% DM	MEINW	
53	2	RES	10K 1/2W 1% DM	MEINW	
52	1	RES	10K 1/2W 1% DM	MEINW	
51	4	RES	10K 1/2W 1% DM	MEINW	
50	2	RES	10K 1/2W 1% DM	MEINW	
49	2	RES	10K 1/2W 1% DM	MEINW	
48	2	RES	10K 1/2W 1% DM	MEINW	
47	1	CAP	1000PF 1/2W 5% DM	MEINW	
46	1	CAP	1000PF 1/2W 5% DM	MEINW	
45	1	CAP	1000PF 1/2W 5% DM	MEINW	
44	1	CAP	1000PF 1/2W 5% DM	MEINW	
43	1	CAP	1000PF 1/2W 5% DM	MEINW	
42	1	CAP	1000PF 1/2W 5% DM	MEINW	
41	1	CAP	1000PF 1/2W 5% DM	MEINW	
40	1	CAP	1000PF 1/2W 5% DM	MEINW	
39	1	CAP	1000PF 1/2W 5% DM	MEINW	
38	1	CAP	1000PF 1/2W 5% DM	MEINW	
37	1	CAP	1000PF 1/2W 5% DM	MEINW	
36	1	CAP	1000PF 1/2W 5% DM	MEINW	
35	1	CAP	1000PF 1/2W 5% DM	MEINW	
34	1	CAP	1000PF 1/2W 5% DM	MEINW	
33	1	CAP	1000PF 1/2W 5% DM	MEINW	
32	1	CAP	1000PF 1/2W 5% DM	MEINW	
31	1	CAP	1000PF 1/2W 5% DM	MEINW	
30	1	CAP	1000PF 1/2W 5% DM	MEINW	
29	1	CAP	1000PF 1/2W 5% DM	MEINW	
28	1	CAP	1000PF 1/2W 5% DM	MEINW	
27	1	CAP	1000PF 1/2W 5% DM	MEINW	
26	1	CAP	1000PF 1/2W 5% DM	MEINW	
25	1	CAP	1000PF 1/2W 5% DM	MEINW	
24	1	CAP	1000PF 1/2W 5% DM	MEINW	
23	1	CAP	1000PF 1/2W 5% DM	MEINW	
22	1	CAP	1000PF 1/2W 5% DM	MEINW	
21	1	CAP	1000PF 1/2W 5% DM	MEINW	
20	1	CAP	1000PF 1/2W 5% DM	MEINW	
19	1	CAP	1000PF 1/2W 5% DM	MEINW	
18	1	CAP	1000PF 1/2W 5% DM	MEINW	
17	1	CAP	1000PF 1/2W 5% DM	MEINW	
16	1	CAP	1000PF 1/2W 5% DM	MEINW	
15	1	CAP	1000PF 1/2W 5% DM	MEINW	
14	1	CAP	1000PF 1/2W 5% DM	MEINW	
13	1	CAP	1000PF 1/2W 5% DM	MEINW	
12	1	CAP	1000PF 1/2W 5% DM	MEINW	
11	1	CAP	1000PF 1/2W 5% DM	MEINW	
10	1	CAP	1000PF 1/2W 5% DM	MEINW	
9	1	CAP	1000PF 1/2W 5% DM	MEINW	
8	1	CAP	1000PF 1/2W 5% DM	MEINW	
7	1	CAP	1000PF 1/2W 5% DM	MEINW	
6	1	CAP	1000PF 1/2W 5% DM	MEINW	
5	1	CAP	1000PF 1/2W 5% DM	MEINW	
4	1	CAP	1000PF 1/2W 5% DM	MEINW	
3	1	CAP	1000PF 1/2W 5% DM	MEINW	
2	1	CAP	1000PF 1/2W 5% DM	MEINW	
1	1	CAP	1000PF 1/2W 5% DM	MEINW	

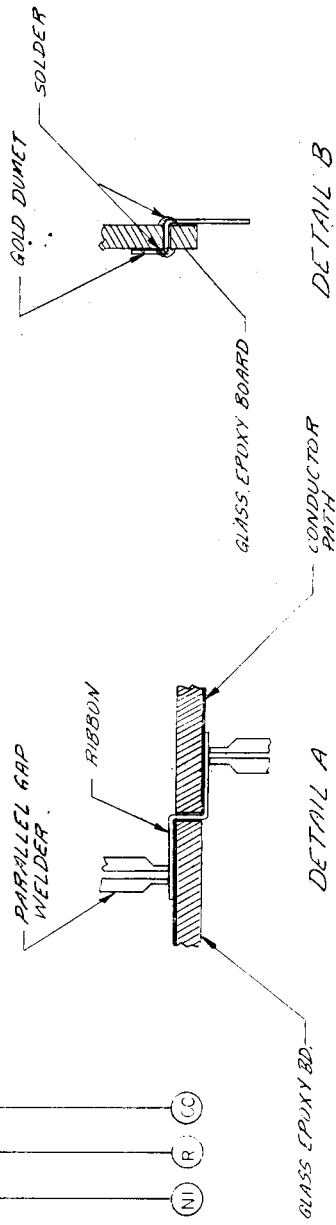




- NOTE:
- 1. COMPONENTS MUST BE SOLDERED TO BOARD ONLY 1 ONLY
 - 2. SEE DETAIL A FOR GOLD DUMMET
 - 3. STEP 1. INSTALL OUTPUT LEADS. 0.020 GOLD DUMMET
 - 4. STEP 2. WELDED FEEDTHRU TO BE INSTALLED PRIOR TO COMPONENTS 62 PLACES. FEEDTHRU MATERIAL 0.01 X 0.004
 - 5. IN GATES (G) CUT-OFF PINS 7 AND 8.
 - 6. IN SHIFT REGISTERS (SR) CUT OFF PINS 1, 8, AND 14.
 - 7. SOLDER PER NASA QUALITY PUBLICATION NPC 200-4.

CIRCUIT COMPONENT LIST			REMARKS
PIN NO	REF	DESCRIPTION	
1	6	SN512 RIPPLE-COUNTER F.F.	T.I.
2	6	SN111 A-S FLIP-FLOP	T.I.
3	2	SN5161 NAND/NOR GATE	T.I.

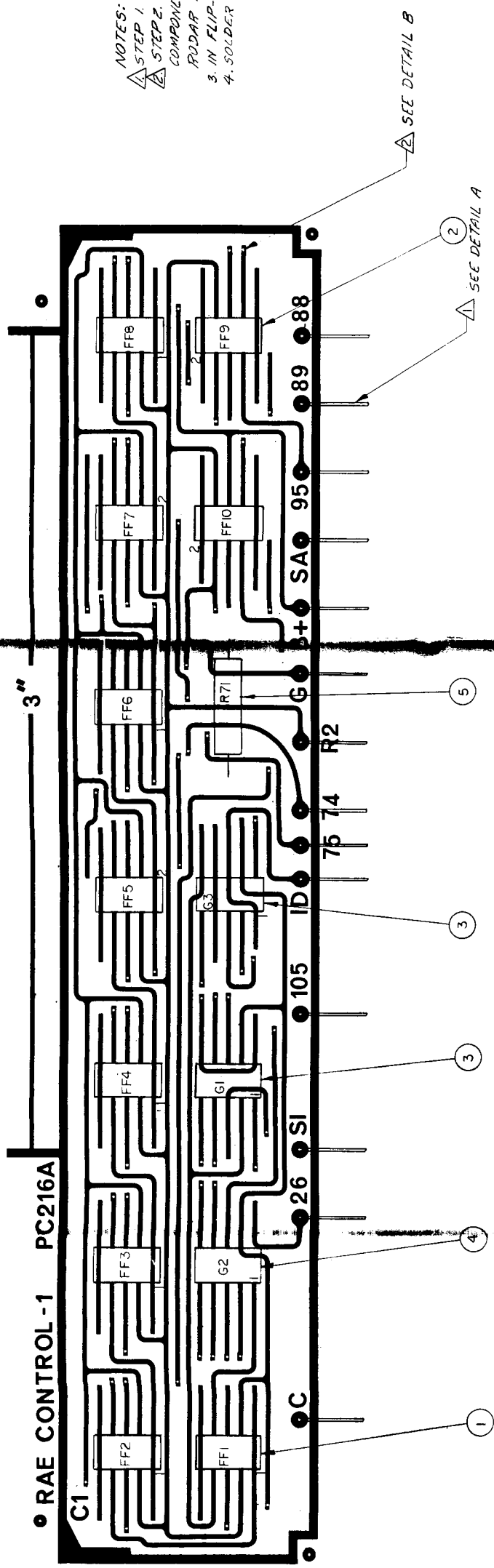
TABLE OF ASSEMBLIES		
PIN NO	REF	DESCRIPTION
1	-1	-2
1	C6 THRU C11	CR THRU C17
2	C16 THRU C17	C18 THRU C19
3	G12 X G13	G14 X G15



WTA 2158 D 487

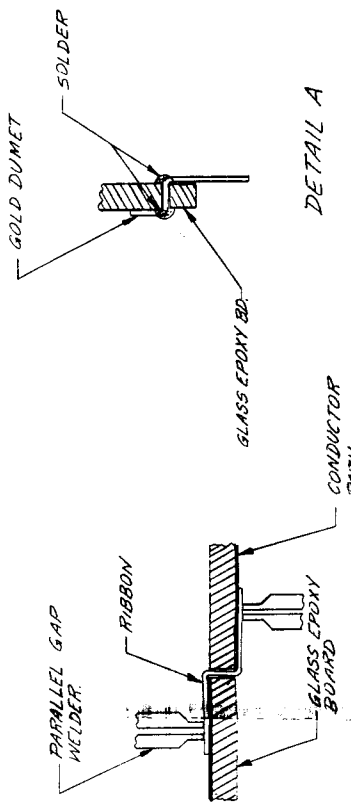
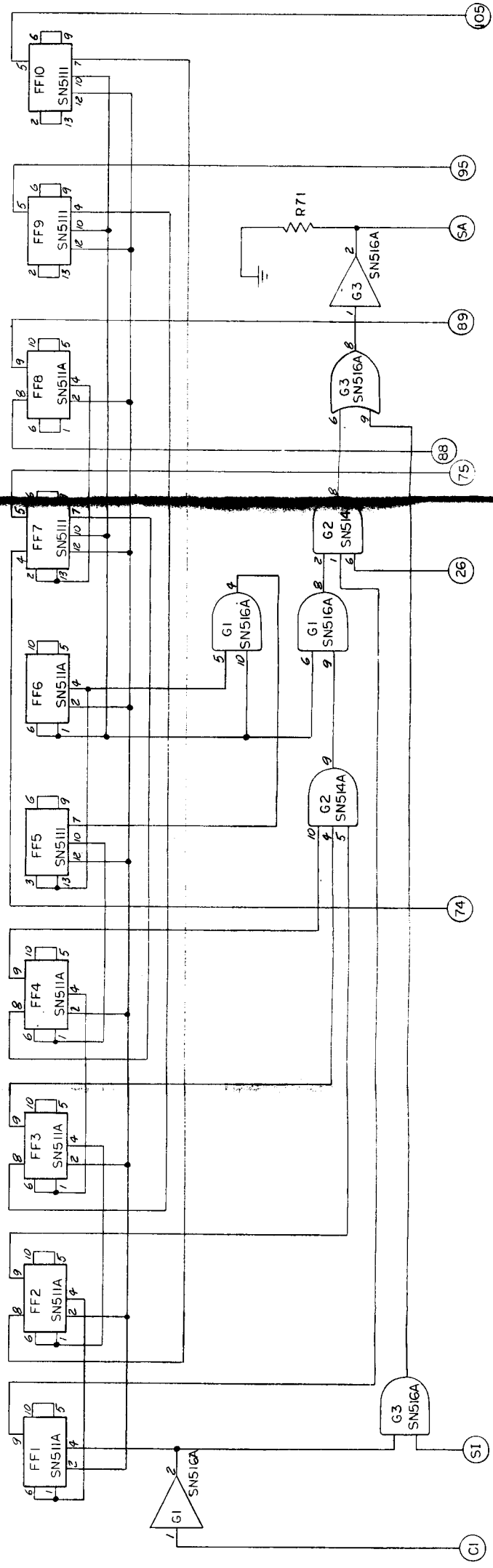
NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		RAE PRINTED WIRING ASSY COUNTER SHIFT REGISTER		SCALE	UNIT WT	CODE	SHEET 1 OF 1
MATERIAL		NAME		INIT	DATE	APPROVED	
DO NOT SCALE THIS DRAWING		DESIGNER		WTA		CHECKED	
REMOVE BURRS, BREAK SHARP EDGES		SCALE:		4:1		FINISH:	
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES		TOLERANCE ON FRACTIONS				ANGLES	
APPLICATION		USED ON				G.D. RAE-1155-102	

RAE CONTROL-1 PC216A



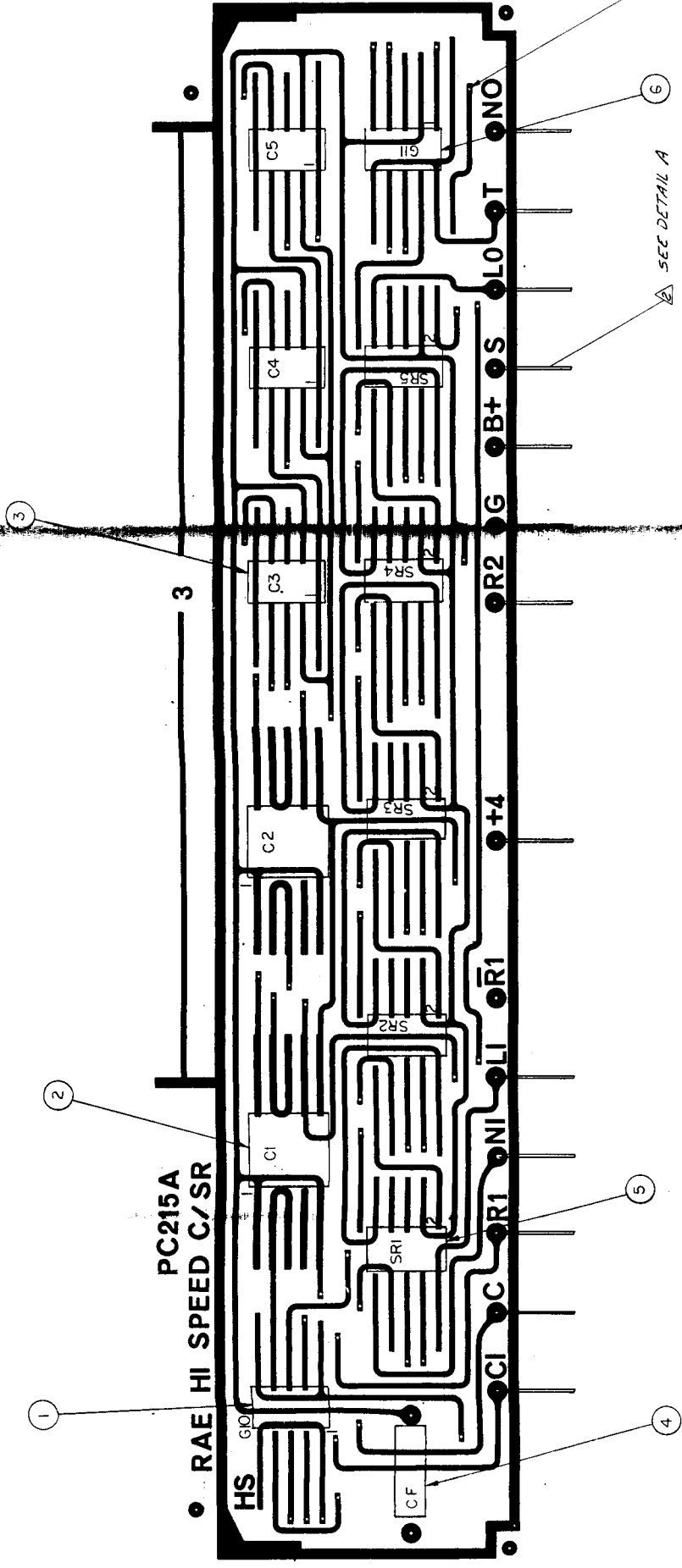
NOTES:
STEP 1. INSTALL OUTPUT LEADS 0.020 GOLD DUMET
STEP 2. WELDED FEEDTHRU TO BE INSTALLED PRIOR TO
COMPONENTS, 62 PLACES. FEEDTHRU MATERIAL 0.010x0.004
RADAR RIBBON.
3. IN FLIP-FLOPS (FF) 5, 7, 9, AND 10 CUTOFF 1, 8, AND 14 LEADS.
4. SOLDER PER NASA QUALITY PUBLICATION NPC 200-4.

CIRCUIT COMPONENT LIST			REMARKS
FIND NO.	QTY	DESCRIPTION	
1	6	SN511A R-S FLIP-FLOP COUNTER	T.I.
2	4	SN511A R-S FLIP-FLOP	T.I.
3	2	SN516A DUAL 2-INPUT NAND/NOR GATE	T.I.
4	1	SN516A DUAL 3-INPUT NAND/NOR GATE	T.I.
5	1	R71 RESISTOR	A-B



WTA 2158 D-495

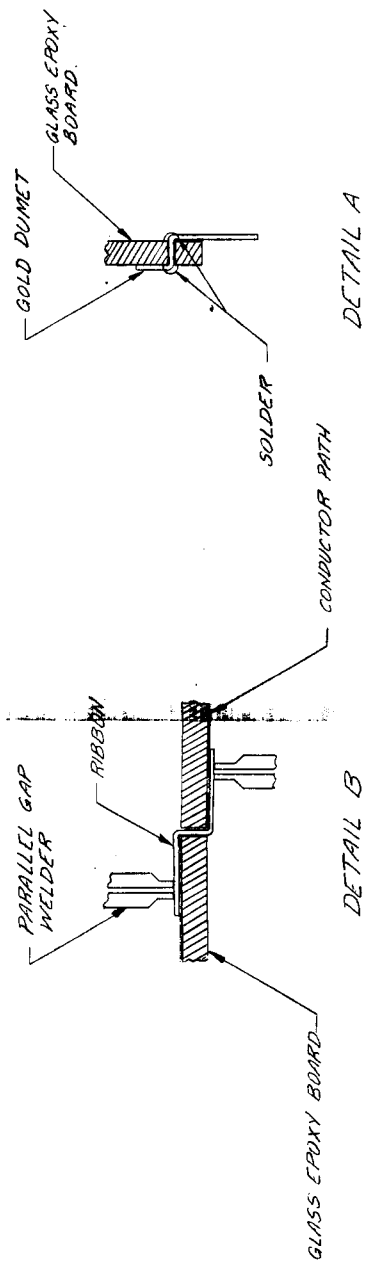
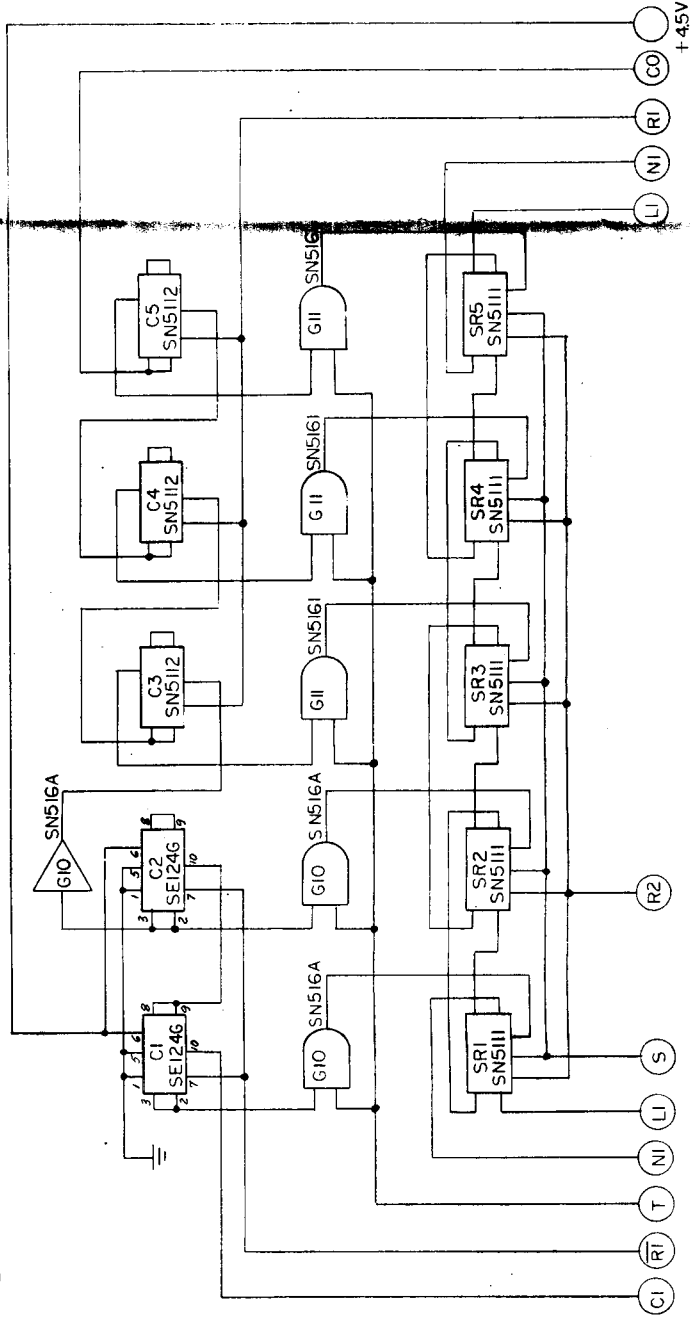
RAE PRINTED WIRING ASSY CONTROL CIRCUIT #1		NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		GD-RAE-1155-403	
NAME	INIT	DATE	SCALE	UNIT	BY
DESIGNED	WTA				
CHECKED					
APPROVED					
DO NOT SCALE THIS DRAWING. REMOVE BURRS, BREAK SHARP EDGES. UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCE ON FRACTIONS ± DECIMALS ± ° ANGLES ±			SCALE: 4:1 RWH: CLIENT:		
MATERIAL			USED ON		
2158-F-501 INT. ASSY.			APPLICATION		



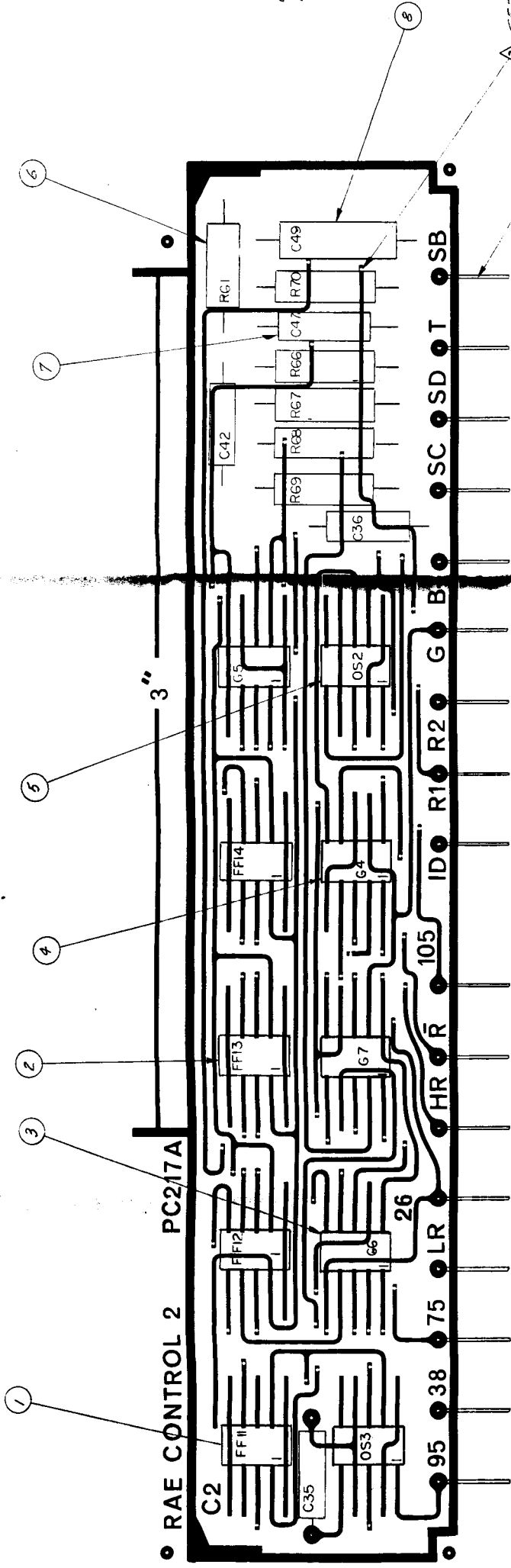
NOTES:

1. IN GATE 11 (6"U) CUT-OFF LEADS 7 & 8.
2. IN ALL SHIFT REGISTERS (5R) CUT-OFF LEADS 4, 8, & 14.
3. STEP 1. INSTALL OUTPUT LEADS 0202 GOLD DUMET.
STEP 2. WELDED FEEDTHRU TO B.C. INSTALLED PRIOR TO COMPONENTS, 5B PLACES. FEEDTHRU MATERIAL 901X1004 RODAR RIBBON.
4. SOLDER PER NASA QUALITY PUBLICATION NPC 200-4.

CIRCUIT COMPONENT LIST		REMARKS
FIND NO	NO. REV'D	DESCRIPTION
1	1	5N516A NAND/NOR GATE
2	2	SE124G BINARY ELEMENT
3	3	5N5112 RIPPLE-COUNTER F.F
4	1	CAPACITOR 150D 1MFD 150V
5	5	5N5111 R-S FLIP-FLOP
6	6	5N5161 NAND/NOR GATE
		T.I.
		SIGNETICS CORP.
		T.I.
		SPRAGUE
		T.I.
		T.I.



NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		RAE PRINTED WIRING ASSY HIS COUNTER SLIP REGISTER		CODE GD RAE-115-5-404	SHEET 1 OF 1
MATERIAL		NAME WTA	INIT	DATE	SCALE UNIT WT
DO NOT SCALE THIS DRAWING REMOVE BUBBLES, BREAK SHARP EDGES		CHECKED APPROVED	APPROVED	APPROVED	SCALE UNIT WT
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES		SCALE: 4:1	PHISH:	PHISH:	SCALE UNIT WT
TOLERANCE ON FRACTIONS ± DECIMALS ± . ANGLES ±		TOLERANCE ON FRACTIONS ± DECIMALS ± . ANGLES ±	TOLERANCE ON FRACTIONS ± DECIMALS ± . ANGLES ±	TOLERANCE ON FRACTIONS ± DECIMALS ± . ANGLES ±	SCALE UNIT WT
USED ON APPLICATION		USED ON APPLICATION	USED ON APPLICATION	USED ON APPLICATION	SCALE UNIT WT
2159 P-50 NEXT ASSY.		2159 P-50 NEXT ASSY.	2159 P-50 NEXT ASSY.	2159 P-50 NEXT ASSY.	SCALE UNIT WT

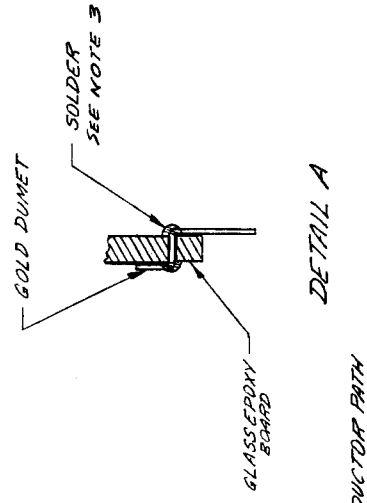
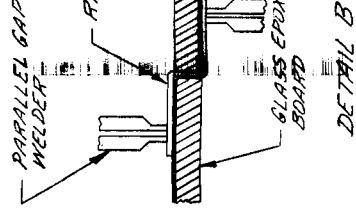
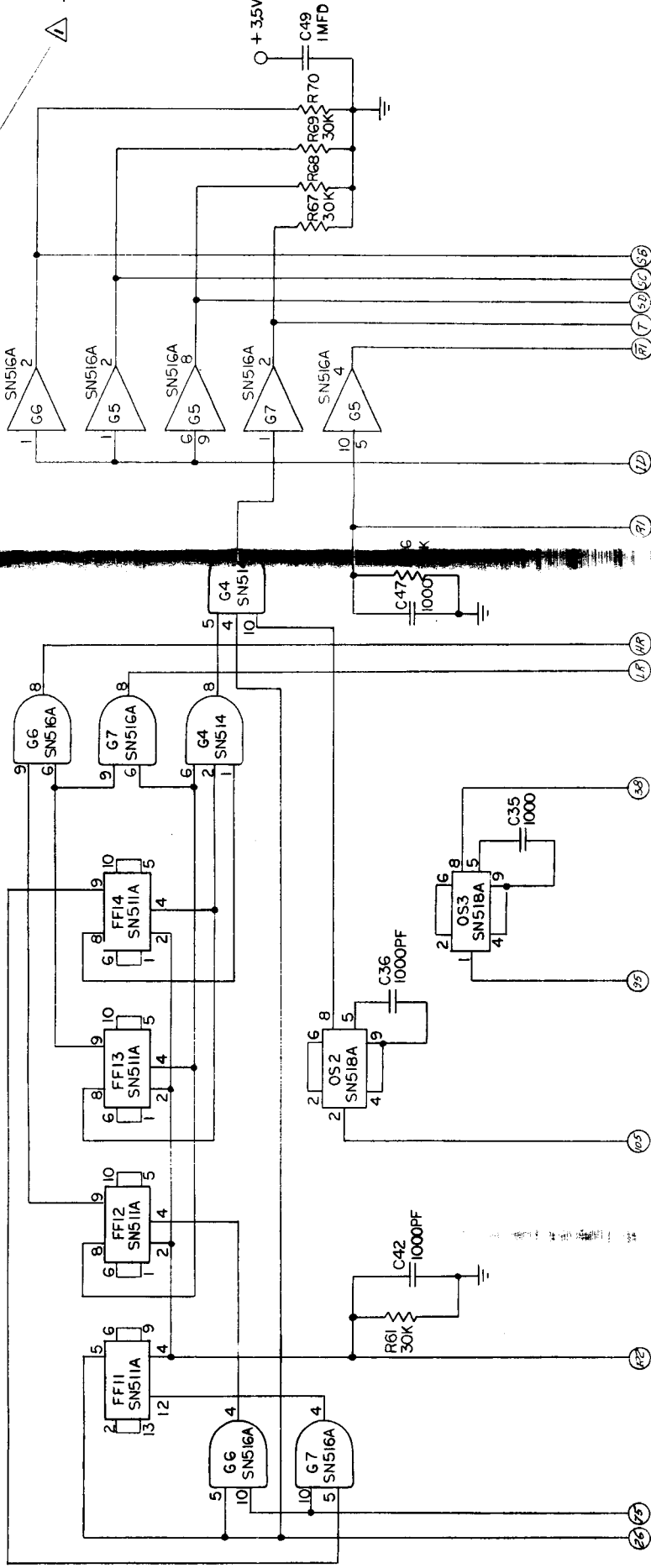


NOTES:
1. STEP 1. INSTALL OUTPUT LEADS 0.020 GOLD DUMMET
2. STEP 2. WELODED FEEDTHRU TO BE INSTALLED
PRIOR TO COMPONENTS, 76 PLACES. FEED THRU
MATERIAL 0.004X0.004 RODAR RIBBON.
3. SOLDER PER NASA QUALITY PUBLICATION NPC 200-4.

SEE DETAIL B

SEE DETAIL A

CIRCUIT COMPONENT LIST			REMARKS
FIND NO.	REF ID	DESCRIPTION	
1	1	SN5111 R/S FLIP-FLOP	T.I.
2	3	SN5114 R/S FLIP-FLOP COUNT	T.I.
3	3	SN5164 DUAL 2 INPUT NAND-NOR	T.I.
4	1	SN5114A DUAL 3 INPUT -N-	T.I.
5	2	SN5184 ONE SHOT	T.I.
6	5	RESISTOR 30K 1/4W ±5%	WELWYN
7	4	CAPACITOR 1000 PF	VITRAMON
8	1	CAPACITOR 1MFD 15V 150 D	SPRAGUE



WTB 2158-D-4-99

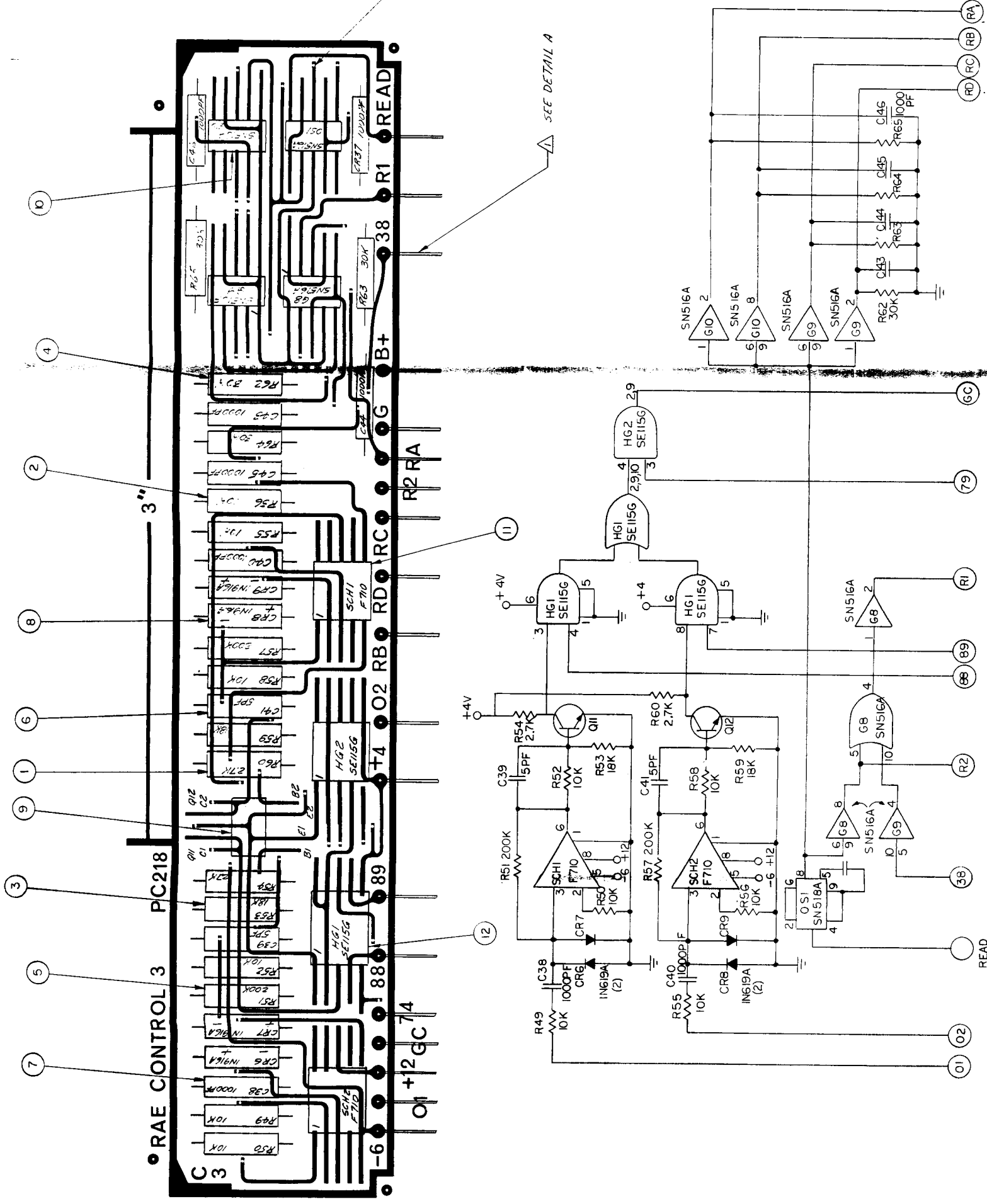
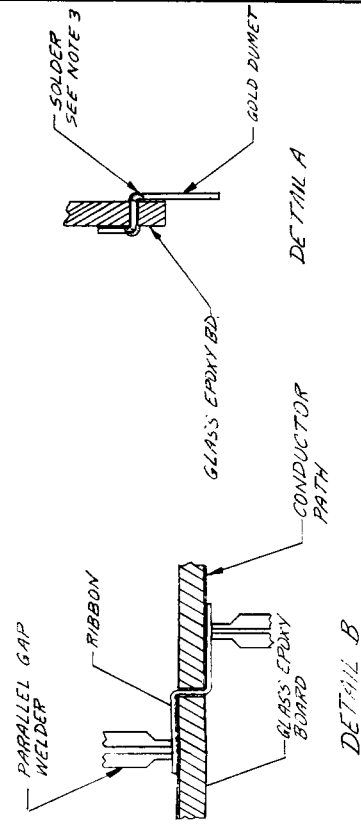
NAME		DATE
DESIGNER	INIT	
WTB		
DO NOT SCALE THIS DRAWING REMOVE BURS, BREAK SHARP EDGES UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCE ON FRACTIONS DECIMALS ANGLES		
SCALE: 4:1 FINISH: APPROVED CLIENT:		
APPLICATION		
NEXT ASST.		
MATERIAL		
RAE CONTROL 2 PRINTED WIRING ASSY CONTROL # 2		
NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		
GD-RAE-1155-405		

NOTES:

1. INSTALL OUTPUT LEADS 0.020 GOLD DUMET
2. STEP 1. WELDED FEED TAP TO BE INSTALLED PRIOR TO COMPONENTS, 50 PLACES. FEEDTHRU MATERIAL 0.001X0.004 ROD AND FINISH.
3. SOLDER PER NASA QUALITY PUBLICATION NPC 200-4.

2 SEE DETAIL B

CIRCUIT COMPONENT LIST		REMARKS
FOUND	DESCRIPTION	
1	RESISTOR 27K 1/4W $\pm 5\%$	NEELWYN
2	10K	NEELWYN
3	1K	NEELWYN
4	30K	NEELWYN
5	RESISTOR 200K 1/4W $\pm 5\%$	1RC
6	CAPACITOR 5PF	AEORNOX
7	CAPACITOR 1000PF	WITRAMON
8	DIODE 1N916A	
9	TRANSISTOR ENBO32	T1
10	5N156A DUAL INP/OUT	T1
11	ALTO AMPLIFIER SCH.1.2	FAIRCHILD
12	SE115G DTL DUAL NAND-NOR 4	SINGNETICS



WIA 2158-D-504

RAE
PRINTED WIRING ASSEMBLY

GD-RAE-1155-406

DO NOT SCALE THIS DRAWING

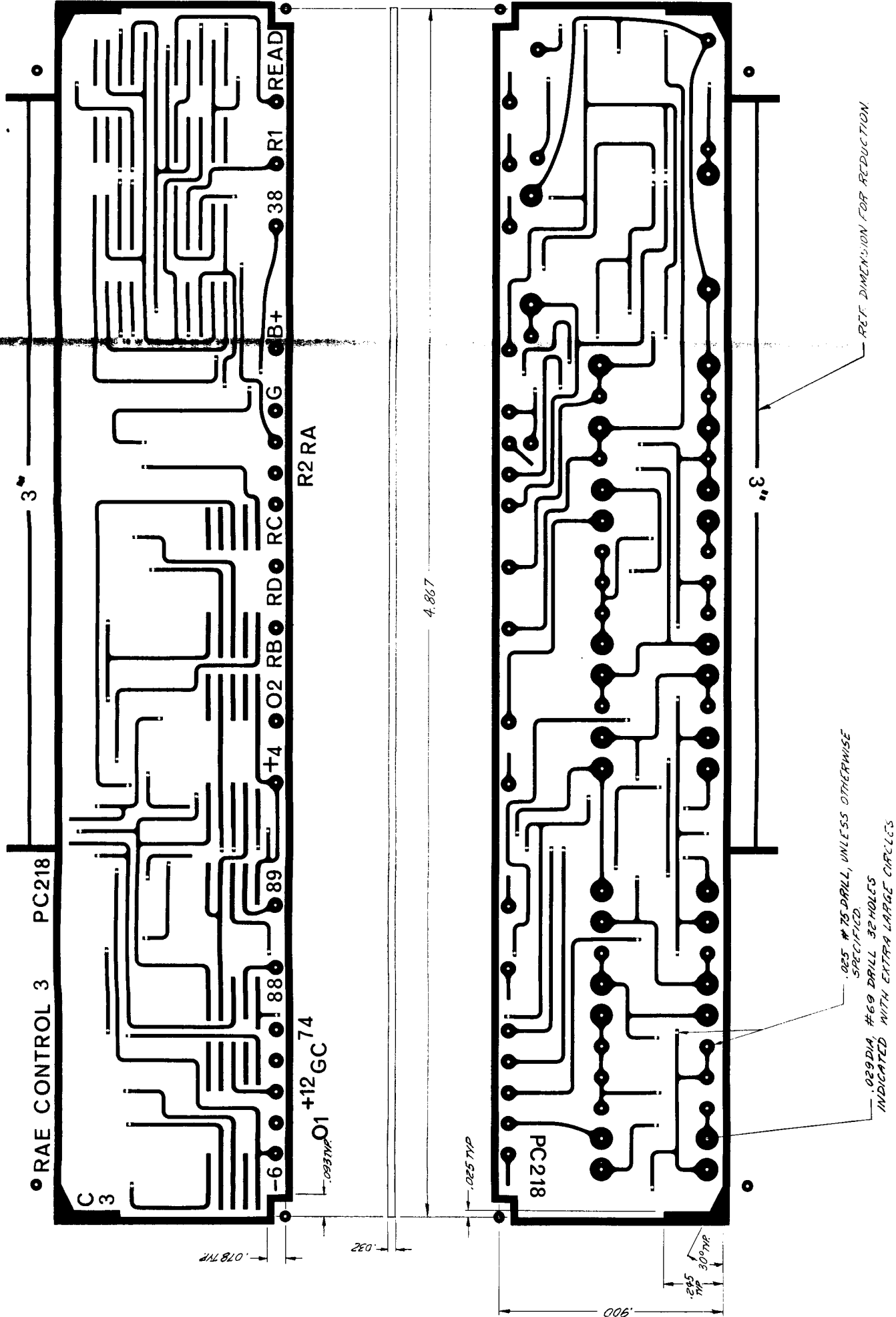
ALL DIMENSIONS ARE IN INCHES
AND FRACTIONS -

2158-P-501
NEXT ASSY.

APPLICATION

NOTES:

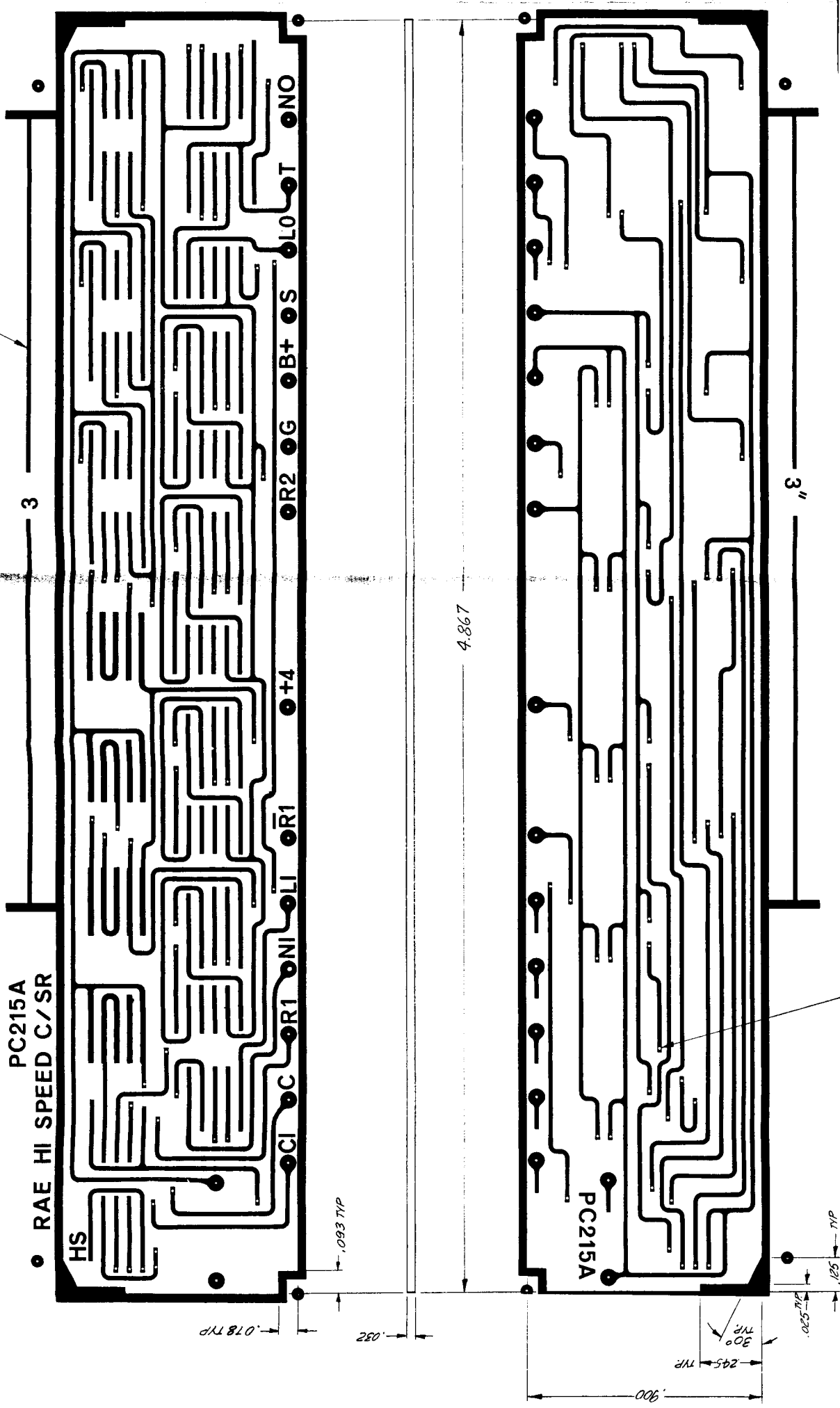
1. GLASS EPOXY LAMINATED SHEET, 2 OZ. COPPER (CLAD), GOLD PLATED, TYPE F1 GE.032 CR.
2. FABRICATE IN ACCORDANCE WITH GEORGE C. MARSHALL SPACE FLIGHT CENTER STANDARD MSFC-STD-154.



WTA 2158-B505

MATERIAL SEE NOTE 1.	NAME	INIT	DATE
	DESIGNER		
DO NOT SCALE THIS DRAWING	DRAWN		
	CHECKED		
REMOVE BURRS. BREAK SHARP EDGES	SCALE: 4:1	APPROVED	
UNLESS OTHERWISE SPECIFIED	FINISH:	APPROVED	
TOLERANCES ARE IN INCHES	CLIENT:		
DECIMALS .005" ANGLES .5°			
8/58 D-204	USED ON		
NEXT ASSY.	APPLICATION		
RAE PRINTED WIRING BOARD CONTROL # 3			
NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND			
GD-RAL-115-501			
CODE SHEET 1 OF 1			

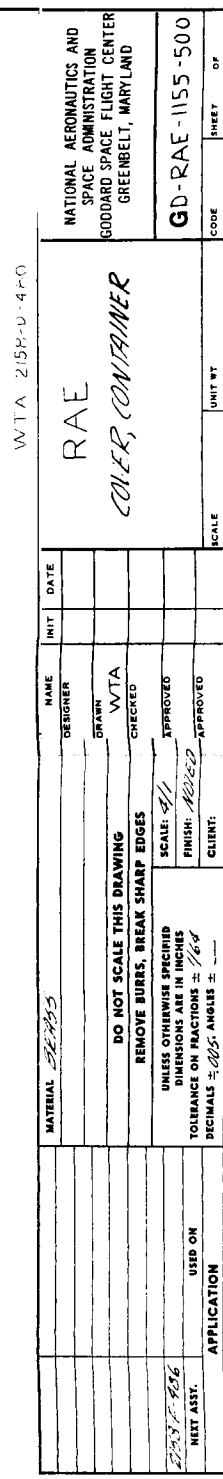
REF. DIMENSION FOR REDUCTION.



-0.021 #75 DRILL ALL HOLES.

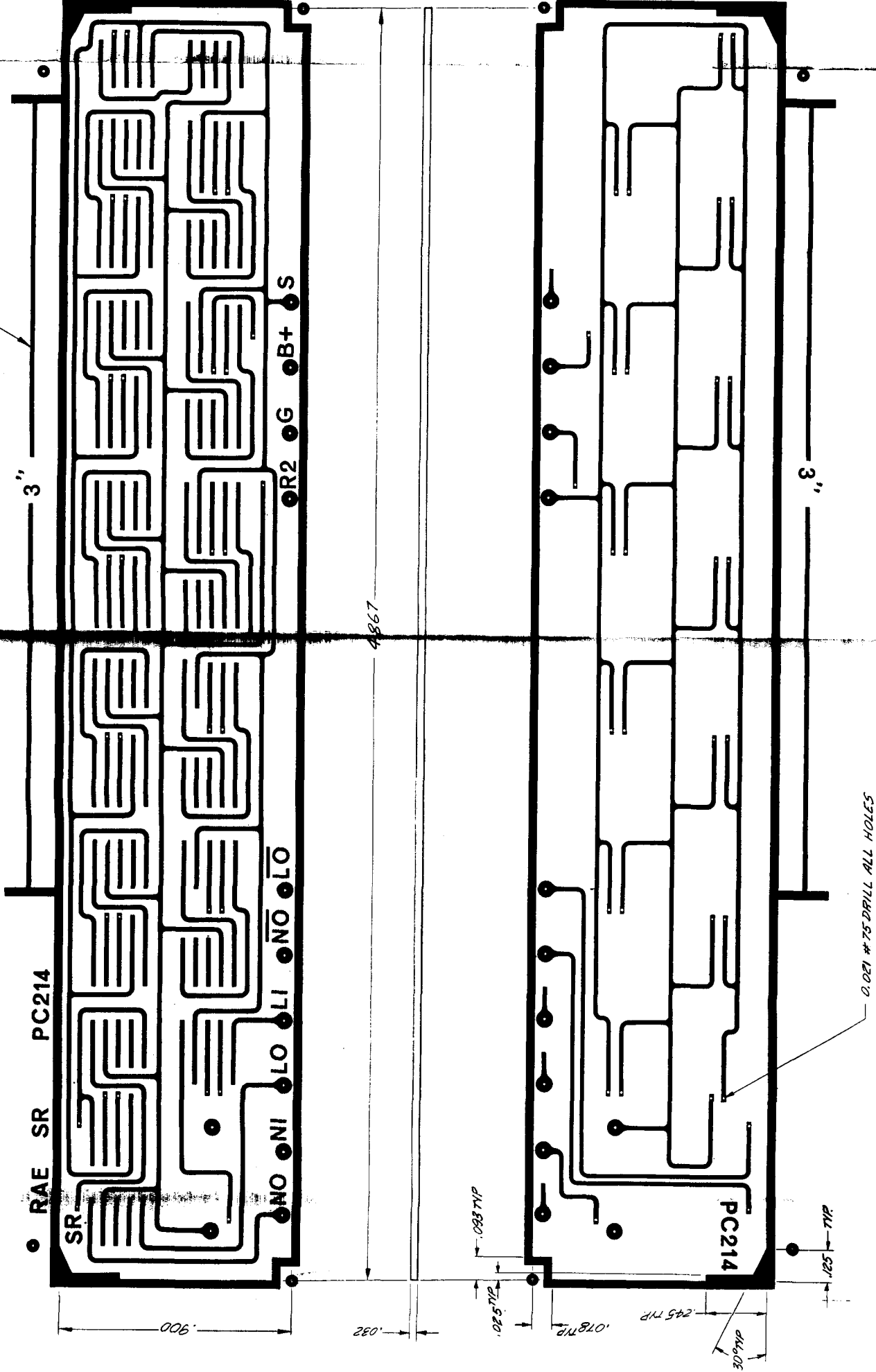
[illegible]

REV	DATE	BY	DESCRIPTION	APPROV



WTA 2158-V-420

REF DIMENSION FOR REDUCTION



- NOTES:
1. GLASS EPOXY LAMINATED SHEET, 2 OZ. COPPER CLAD, GOLD PLATED, TYPE FC GE, 0.032 CB.
 2. FABRICATE IN ACCORDANCE WITH GEORGE C. MARSHALL SPACE FLIGHT CENTER STANDARD MSFC-STD-154

WTA 2158-D-492

NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		RAE PRINTED WIRING BOARD SHIFT REGISTER		GD-RAE-1155-503	
NAME	INIT	DATE	SCALE	UNIT WT	SHEET 1 OF 1
DESIGNER					
DRAWN	WTA				
CHECKED					
APPROVED					
APPROVED					
MATERIAL SEE NOTE 1			DO NOT SCALE THIS DRAWING		
REMOVE BURS, BREAK SHARP EDGES			UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCES ON REACTIONS ± DECIMALS ±.005, ANGLES ±.7°		
2158-D-492 NEXT ASST.			USED ON APPLICATION		

Technical drawing of a mechanical part, likely a propeller hub or similar component, showing a cross-section with various dimensions and radii. The drawing includes the following specifications:

- Overall length: 4750
- Radius: $.250 \pm .003$
- Radius: $.187$
- Radius: $.093$
- Radius: $.093$ R TYP
- Radius: $.060$ R TYP
- Radius: $.020$ R TYP
- Dimension: 500
- Angle: $22^{\circ}30'$ TYP

Technical drawing of a capacitor probe assembly, showing a side view with dimensions and a top view with a central rectangular feature. The drawing includes various tolerances and a manufacturer's stamp.

Dimensions and Tolerances:

- Overall length: $7.095 \pm .002$
- Overall width: $1.87 \pm .002$
- Top view width: $.339$
- Top view height: $.039 \text{ THP}$
- Top view width (inner): $.156 \pm .002$
- Top view height (inner): $.437 \pm .002$
- Top view width (outer): $.317 \pm .002 \text{ THP}$
- Top view height (outer): $.187 \pm .002$

Manufacturer's Stamp:

SPECIFICATION NO. 2158-1
 PART NO. 2158-R-434
 CAPACITANCE PROBE
 SERIAL NO. 1
 MANUFACTURED JULY 1965 BY
 WASHINGTON TECHNOLOGICAL ASSOCIATES, INC.
 CONTRACT NO. NAS 5-9175

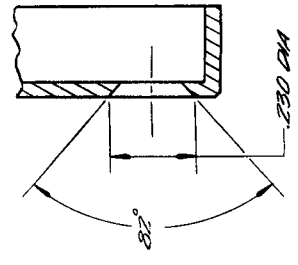
NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		RAE SIDE PANEL, RIGHT		SCALE		UNIT WT		CODE		SHEET 1 OF 1	
MATERIAL <i>ALUMINUM AL-30.15</i>		NAME		INIT		DATE					
		DESIGNER									
		DRAWN		WTA							
		CHECKED									
		DO NOT SCALE THIS DRAWING									
		REMOVE BURS, BREAK SHARP EDGES									
		UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES		SCALE: $\frac{1}{2} = 1$		APPROVED					
		TOLERANCE ON REACTIONS $\pm .005$		FINISH: <i>ANOD</i>		APPROVED					
		DIMENSIONS $\pm .005$, ANGLES $\pm .2^\circ$		CLIENT:							
		NEXT ASST.		USED ON							
		APPLICATION									

NATIONAL AERONAUTICS AND
SPACE ADMINISTRATION
GODDARD SPACE FLIGHT CENTER
GREENBELT, MARYLAND

GD-RAE-1155-504

CODE	SHEET 1 OF 1
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REV	DATE	BY	DESCRIPTION	APPVD



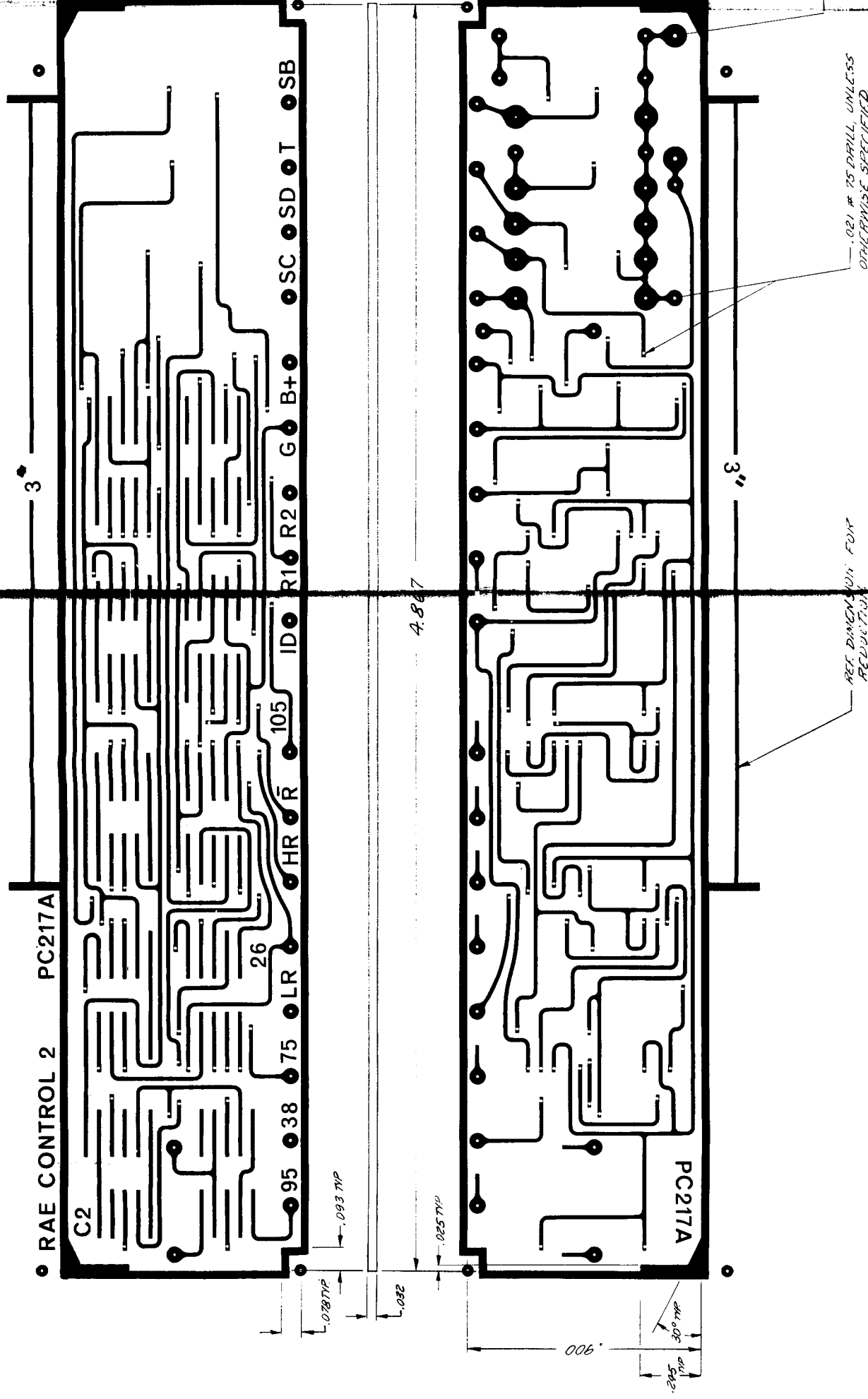
SECTION A-A
SCALE: $\frac{4}{1}$
TYP 4 PLACES

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REV	DATE	BY	DESCRIPTION	APYD
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NOTES:

1. GLASS EPOXY SHEET, LAMINATED, 2 OZ. COPPER CLAD, GOLD PLATED, TYPE FL GE .032 CZ.
2. FABRICATE IN ACCORDANCE WITH GEORGE C. MARSHALL SPACE FLIGHT CENTER STANDARD MSFC-STD-154.



WTA 2158-D-500

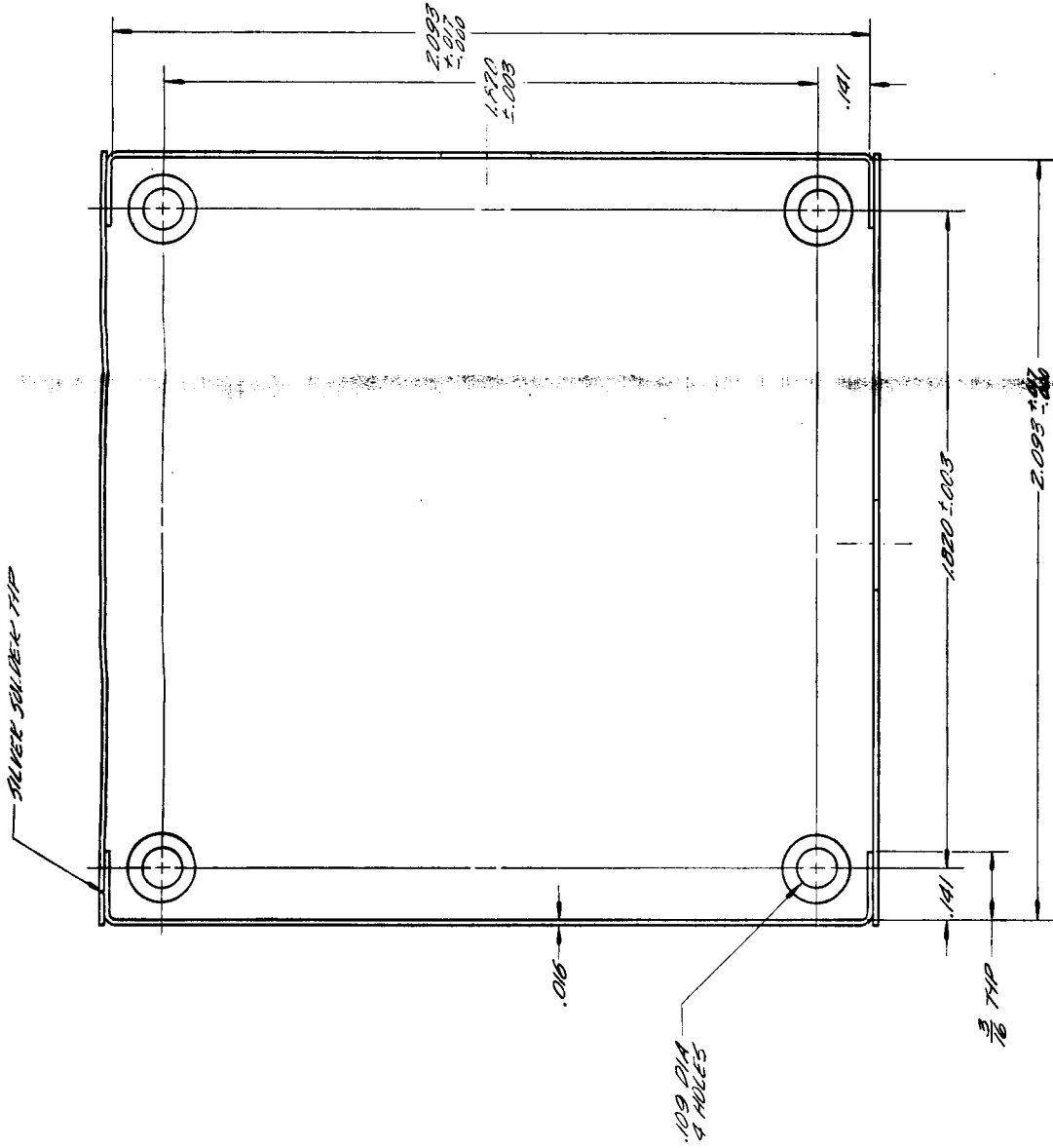
RAE
PRINTED WIRING
BOARD CONTROL #2

NATIONAL AERONAUTICS AND
SPACE ADMINISTRATION
GODDARD SPACE FLIGHT CENTER
GREENBELT, MARYLAND

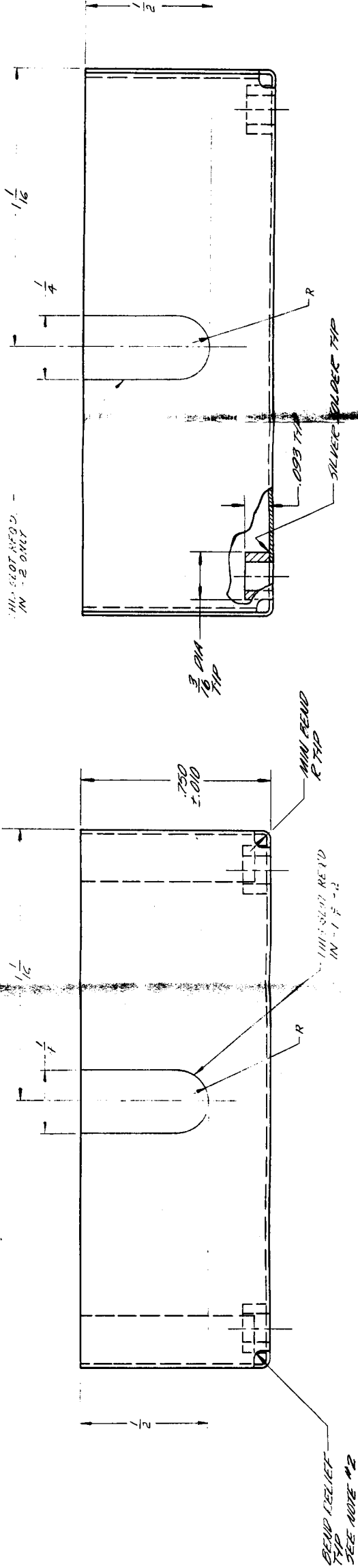
GD-RAE-1155-506

MATERIAL SEE NOTE 1.		DO NOT SCALE THIS DRAWING		SCALE: 4:1	
NAME		REMOVE BURRS, BREAK SHARP EDGES		UNITS: DIMENSIONS SPECIFIED	
SIGNATURE		WTA		DIMENSIONS ARE IN INCHES	
REVISION		CHECKED		TOLERANCE ON FRACTIONS .	
DATE		APPROVED		DECIMALS .005 ANGLES .1°	
INITIAL		APPROVED		APPLICATION	
DATE		APPROVED		NEXT ASSY.	

NOTES: UNLESS OTHERWISE SPECIFIED:
1. FINISH ALL OVER
2. FILL BEND RELIEFS WITH SOLDER
AFTER FABRICATION



QTY	NO.	NO. OF SLOTS	REV'D.
1	1	1	
2	2	2	

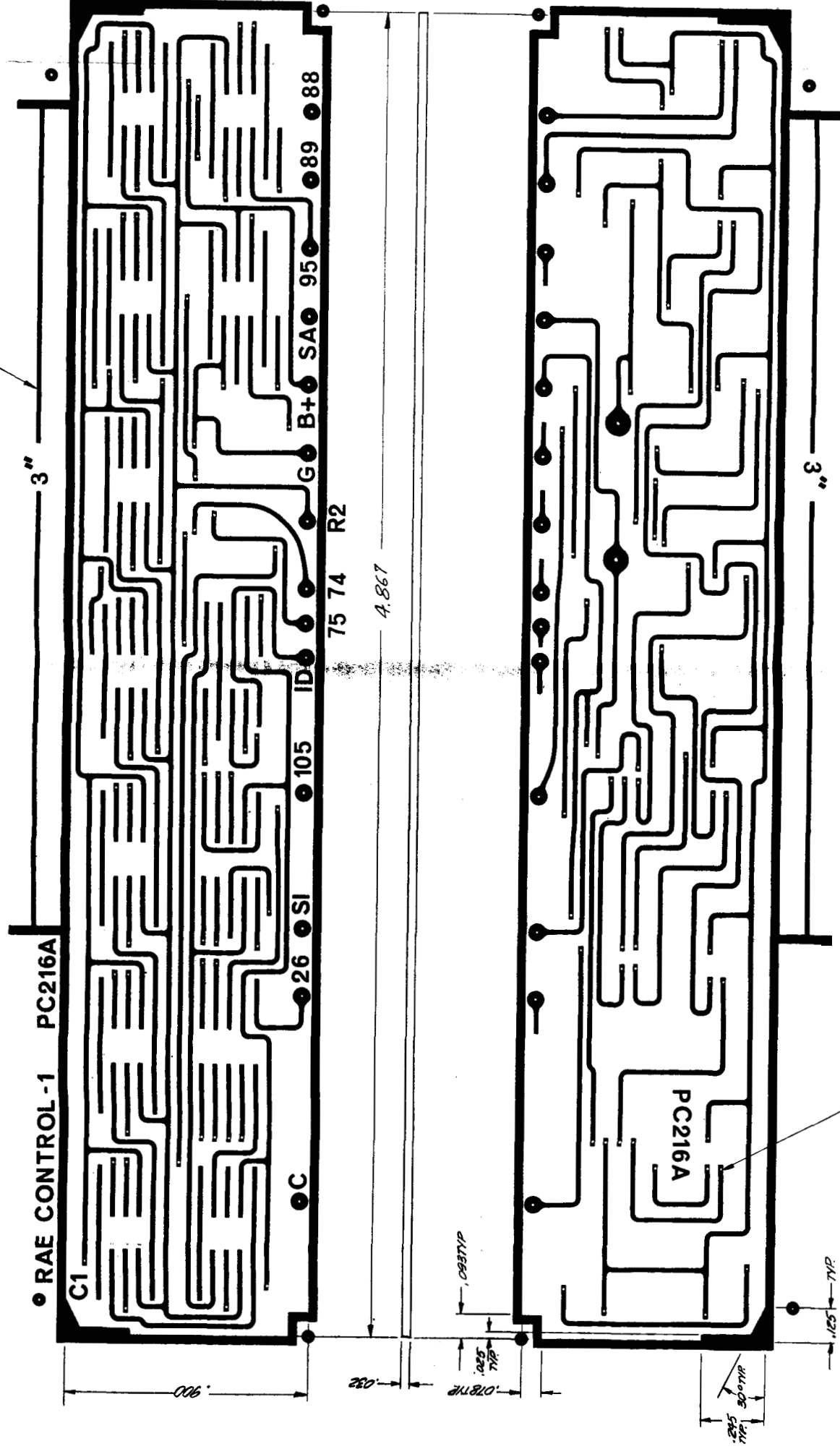


WT A 2158-D-479

RAE CONTAINER		NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		GD-RAE-1155-507	
SCALE		UNIT WT		CODE	
MATERIAL BRASS		NAME		DATE	
DO NOT SCALE THIS DRAWING		WTA		WTA	
REMOVE BURE, BREAK SHARP EDGES		CHECKED		APPROVED	
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES TOLERANCE ON DIMENSIONS ± .005		SCALE: 1/1		FINISH: AS SHOWN	
APPROVED BY: [Signature]		USED ON		APPLICATION	
NEXT ASSY.		APPROVED BY: [Signature]		APPROVED BY: [Signature]	

REV	DATE	BY	DESCRIPTION	APPRO

- REF. DIMENSIONS FOR REDUCTION.

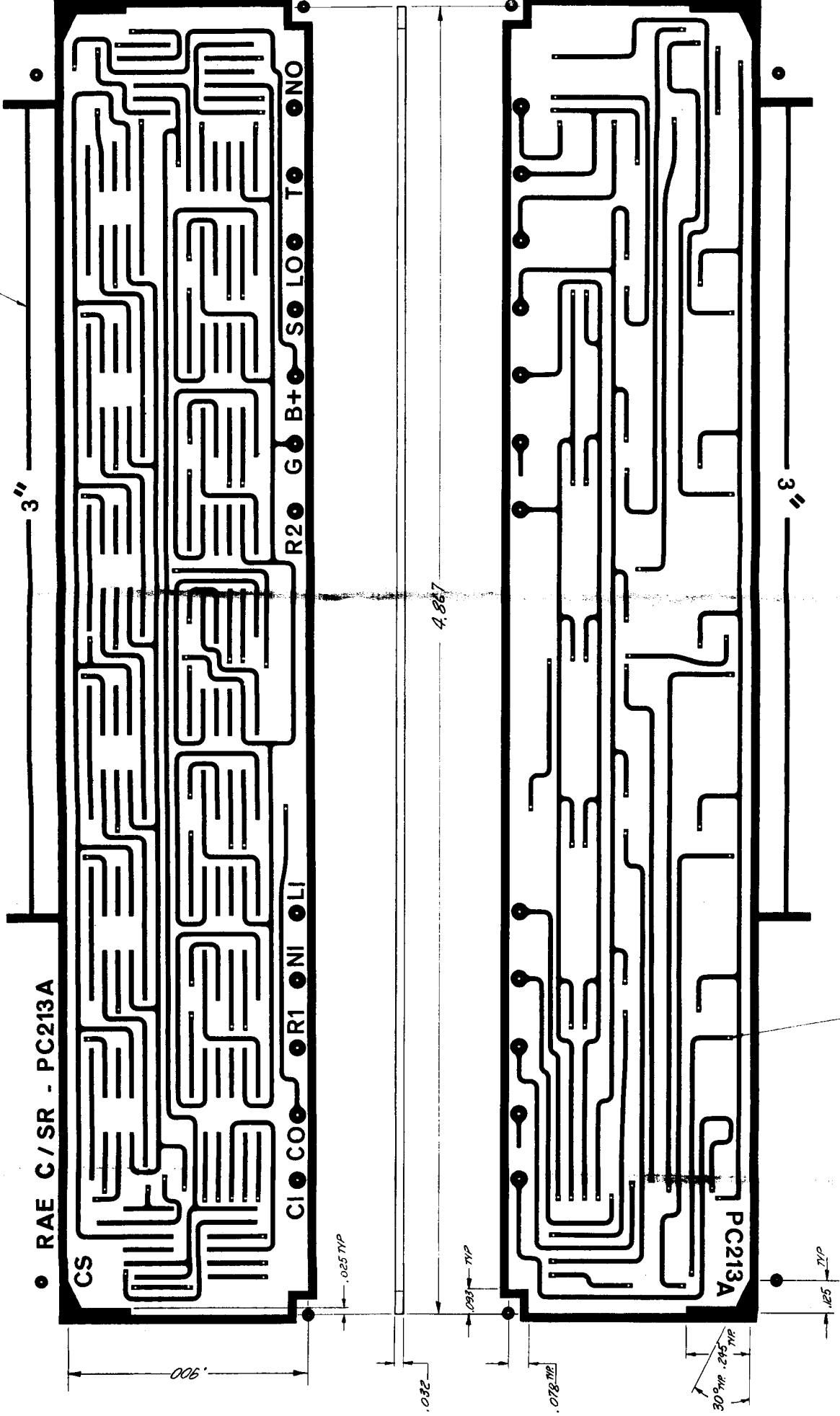


NOTES:

1. GLASS EPOXY LAMINATED SHEET, 2 OZ. COPPER CLAD, GOLD PLATED, TYPE FL GE .032 C2.
2. FABRICATE IN ACCORDANCE WITH GEORGE C. MARSHALL SPACE FLIGHT CENTER STANDARD MSFC-STD-154.

WT-58-D-423 NEXT ASSY.	APPLICATION USED ON	MATERIAL <i>SEE NOTE 1.</i>		NAME DESIGNER		INIT	DATE	WT-58-D-494	
		DO NOT SCALE THIS DRAWING REMOVE BURS, BEAK SHARP EDGES		WT-58				RAE	
		UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES		WT-58				PRINTED WIRING	
		TOLERANCES: FRACTIONS ± .005, DECIMALS ± .005, ANGLES ± 1°		WT-58				BOARD CONTROL	
		FINISH:		WT-58				CIRCUIT # 1	
		CLIENT:		WT-58				NATIONAL AERONAUTICS AND SPACE ADMINISTRATION 60000 SPACE FLIGHT CENTER GREENBELT, MARYLAND	
SCALE		UNIT WT		CODE		INSET		1 OF 1	
GD-RAE-1155-508									

REF DIMENSION FOR REDUCTION.

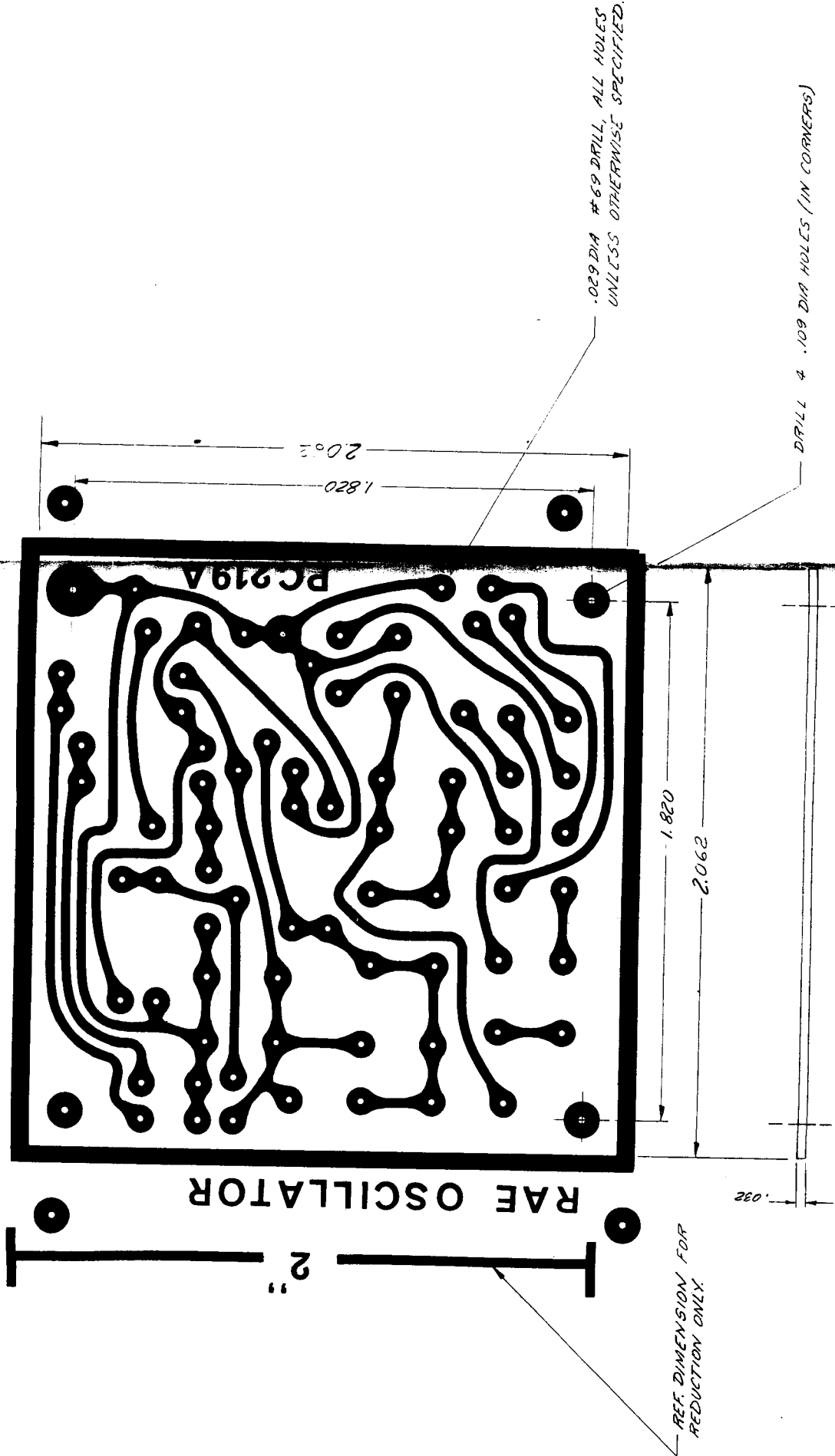


NOTES:

1. GLASS EPOXY LAMINATED SHEET, 2 OZ. COPPER CLAD, GOLD PLATED. TYPE FL Q5-032 C2.
2. FABRICATE P.C. BOARDS IN ACCORDANCE WITH GEORGE C. MARSHALL SPACE FLIGHT CENTER STANDARD MSFC-STD-154.

WTA 2158-D-488

NAME		INIT	DATE
DESIGNER			
DRAWN		WTA	
CHECKED			
APPROVED			
APPROVED			
SCALE		4:1	
FINISH			
CLIMB			
MATERIAL		SEE NOTE 1	
DO NOT SCALE THIS DRAWING			
REMOVE BURRS, BREAK SHARP EDGES			
UNLESS OTHERWISE SPECIFIED			
DIMENSIONS ARE IN INCHES			
TOLERANCES		DECIMALS ± .005, ANGLES ± 1°	
NEXT ASSY.		USED ON	
APPLICATION			
RAE		PRINTED WIRING BOARD COUNTER SHEET REGISTER	
NATIONAL AERONAUTICS AND SPACE ADMINISTRATION		GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND	
GD-RAE-1155-509		CODE	
WTA 2158-D-488		UNIT WT	
SHEET		OF 1	

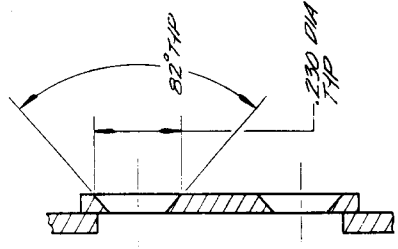
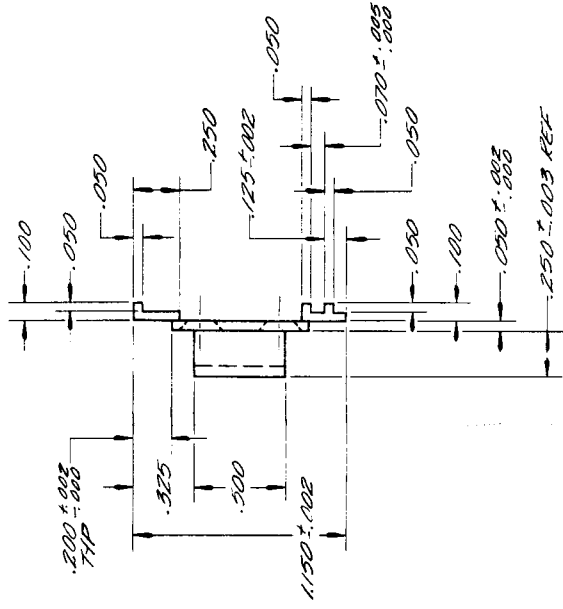
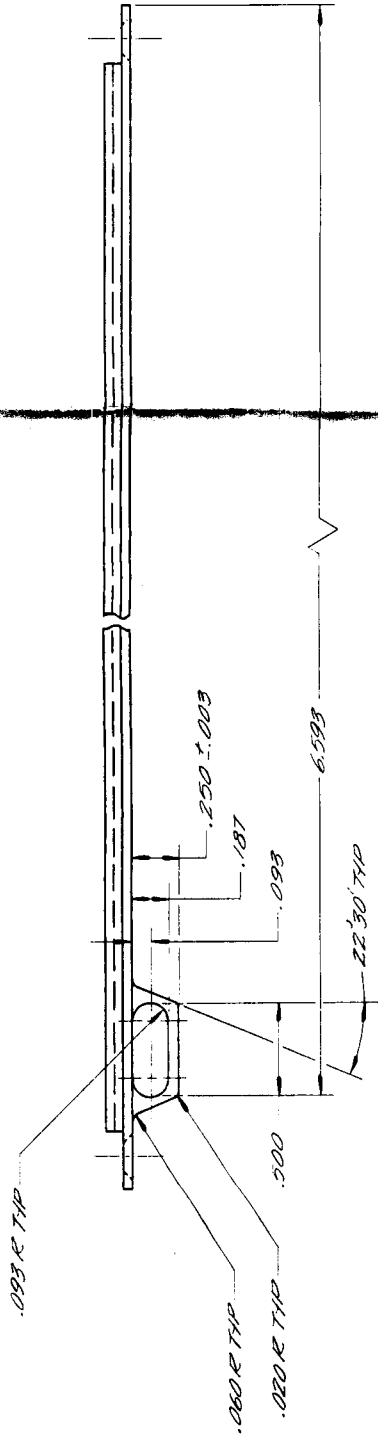


- NOTES:
1. MATERIAL - GLASS EPOXY LAMINATED SHEET, 2 OZ. COPPER CLAD, GOLD PLATED, TYPE PL GE. 092-02.
 2. FABRICATE IN ACCORDANCE WITH GEORGE C. MARSHALL SPACE FLIGHT CENTER STANDARD NSFC-STD-154.

WTA 2158-D-507

NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GODDARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		RAE PRINTED WIRING BOARD, RAE OSCILLATOR		GD-RAE-1155-510	
NAME	INIT	DATE	SCALE	UNIT	WT
DESIGNER					
DRAWN	WTA				
CHECKED					
APPROVED					
APPROVED					
MATERIAL SEE NOTE 1			DO NOT SCALE THIS DRAWING		
REMOVE BURS, BREAK SHARP EDGES			SCALE: 4:1		
UNLESS OTHERWISE SPECIFIED DIMENSIONS ARE IN INCHES			FINISH:		
TOLERANCE ON FRACTIONS ±			CLIENT:		
DECIMALS ±			ANGLES ±		
USED ON			APPLICATION		
2158-D-507			NEXT ASSY:		

NOTES: UNLESS OTHERWISE SPECIFIED:
1. FINISH: ALUMINUM SURF.
2. FINISH: ALUMINUM SURF.
ALL DIMENSIONS ARE IN INCHES
AND DECIMALS THEREOF



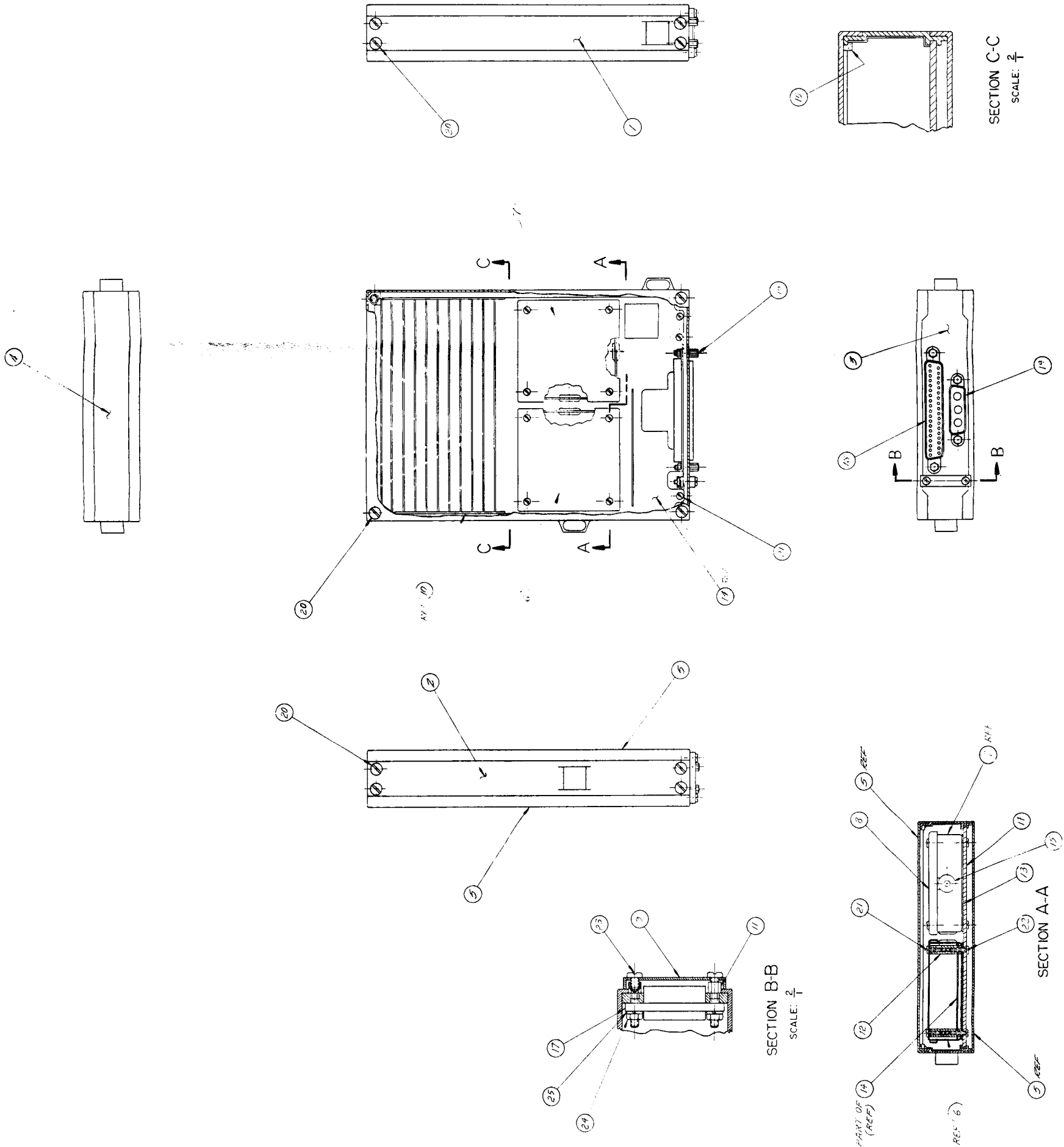
SECTION A A
SCALE: 4/1
TYP 2 PLACES

WTA 2153-D-414

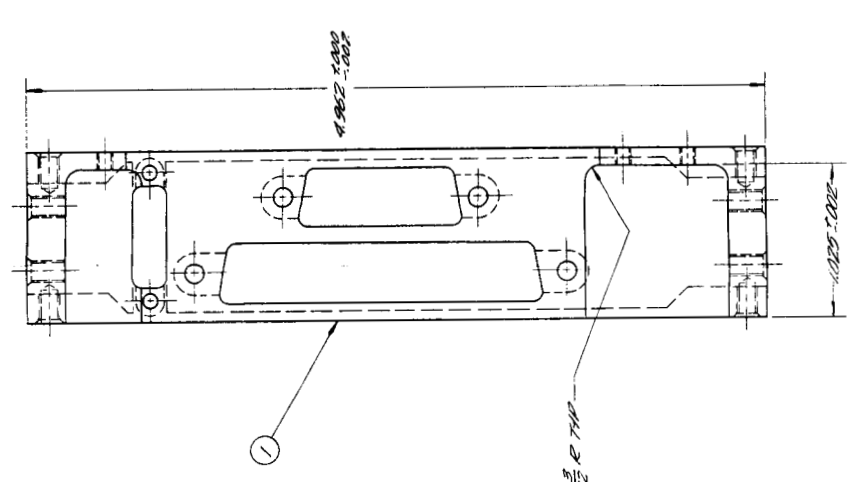
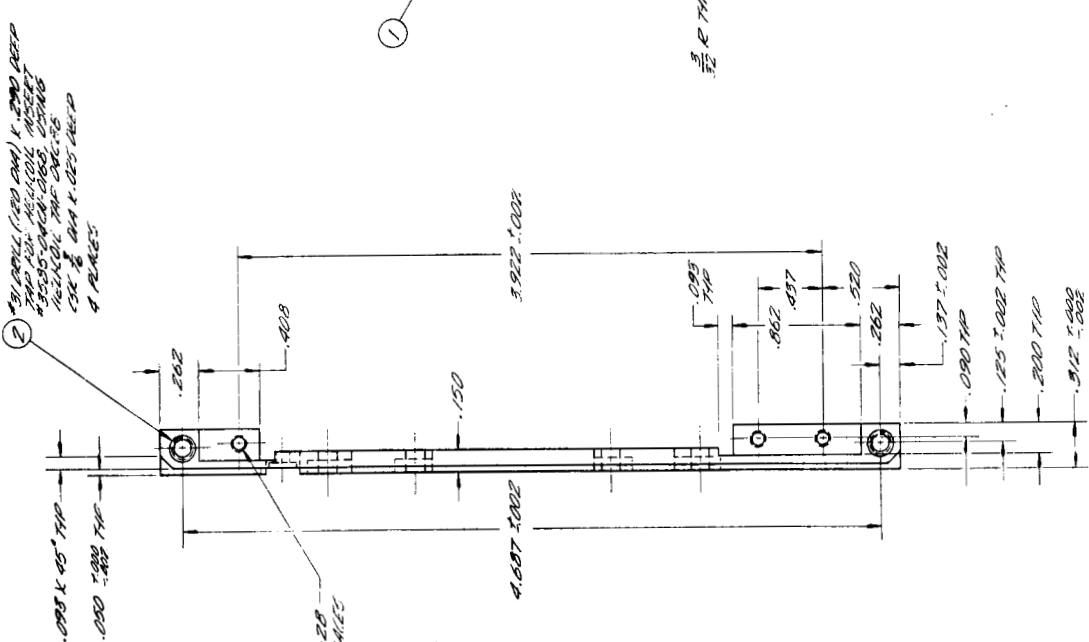
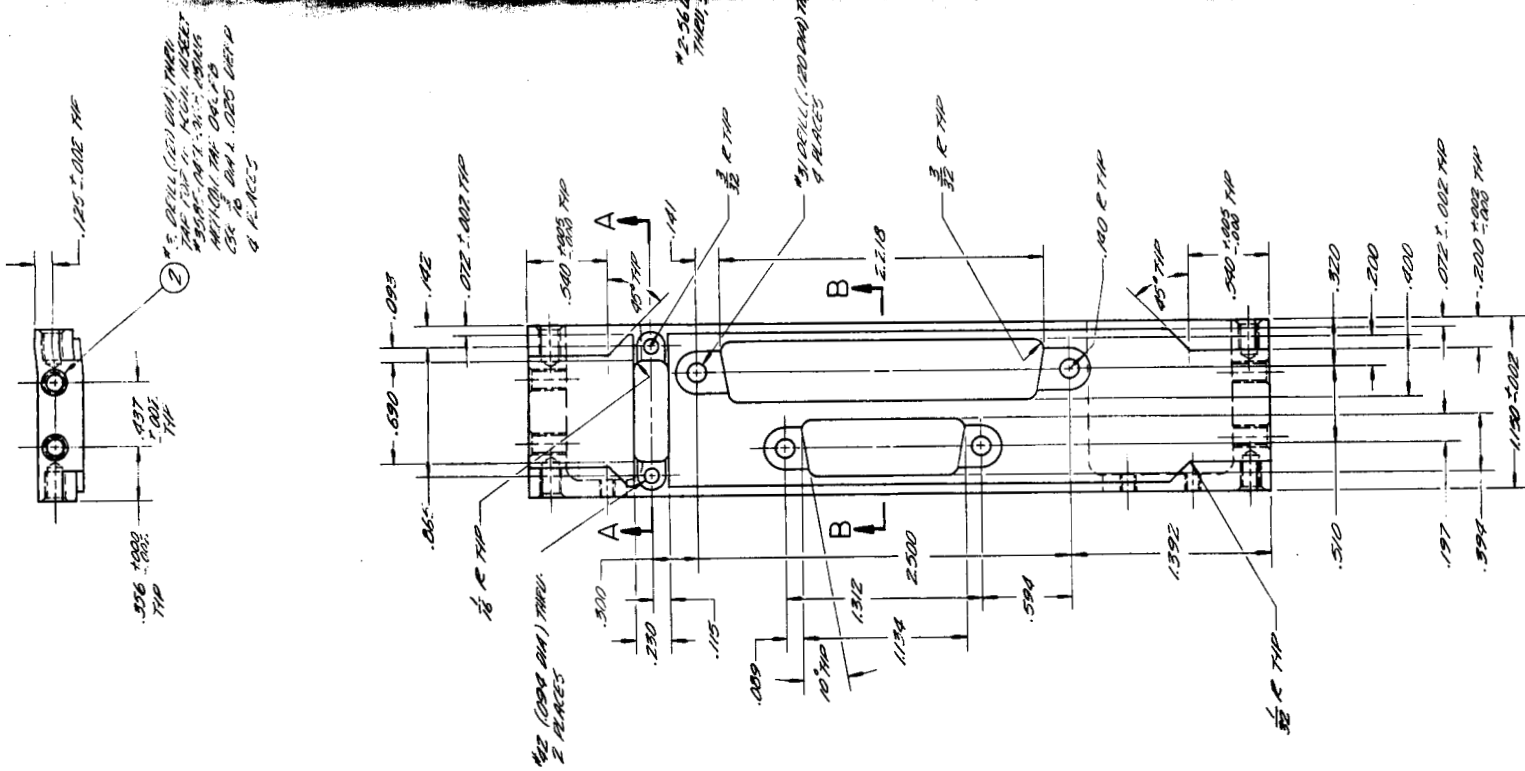
MATERIAL: ALUMINUM 1100-0		NAME: WTA	INIT: WTA	DATE: 12-3-78
DESIGNER: WTA		CHECKED: WTA	DATE: 12-3-78	
DO NOT SCALE THIS DRAWING		SCALE: 4/1	APPROVED: WTA	
REMOVE BURRS, BREAK SHARP EDGES		FINISH: 1100-0	APPROVED: WTA	
UNLESS OTHERWISE SPECIFIED		FINISH: 1100-0	APPROVED: WTA	
DIMENSIONS ARE IN INCHES		FINISH: 1100-0	APPROVED: WTA	
TOLERANCE ON FRACTIONS ± .001		FINISH: 1100-0	APPROVED: WTA	
DECIMALS ± .001		FINISH: 1100-0	APPROVED: WTA	
ANGLES ± 2°		FINISH: 1100-0	APPROVED: WTA	
APPLICATION: USED ON		FINISH: 1100-0	APPROVED: WTA	
NEXT ASSY: 2153-D-416		FINISH: 1100-0	APPROVED: WTA	
NATIONAL AERONAUTICS AND SPACE ADMINISTRATION		FINISH: 1100-0	APPROVED: WTA	
GODDARD SPACE FLIGHT CENTER		FINISH: 1100-0	APPROVED: WTA	
GREENBELT, MARYLAND		FINISH: 1100-0	APPROVED: WTA	
CODE: GD-RAE-1155-511		FINISH: 1100-0	APPROVED: WTA	
SHEET: 1 OF 1		FINISH: 1100-0	APPROVED: WTA	

NOTES:

1. SOLDER PER MASA QUALITY PUBLICATION
NPC 200-4



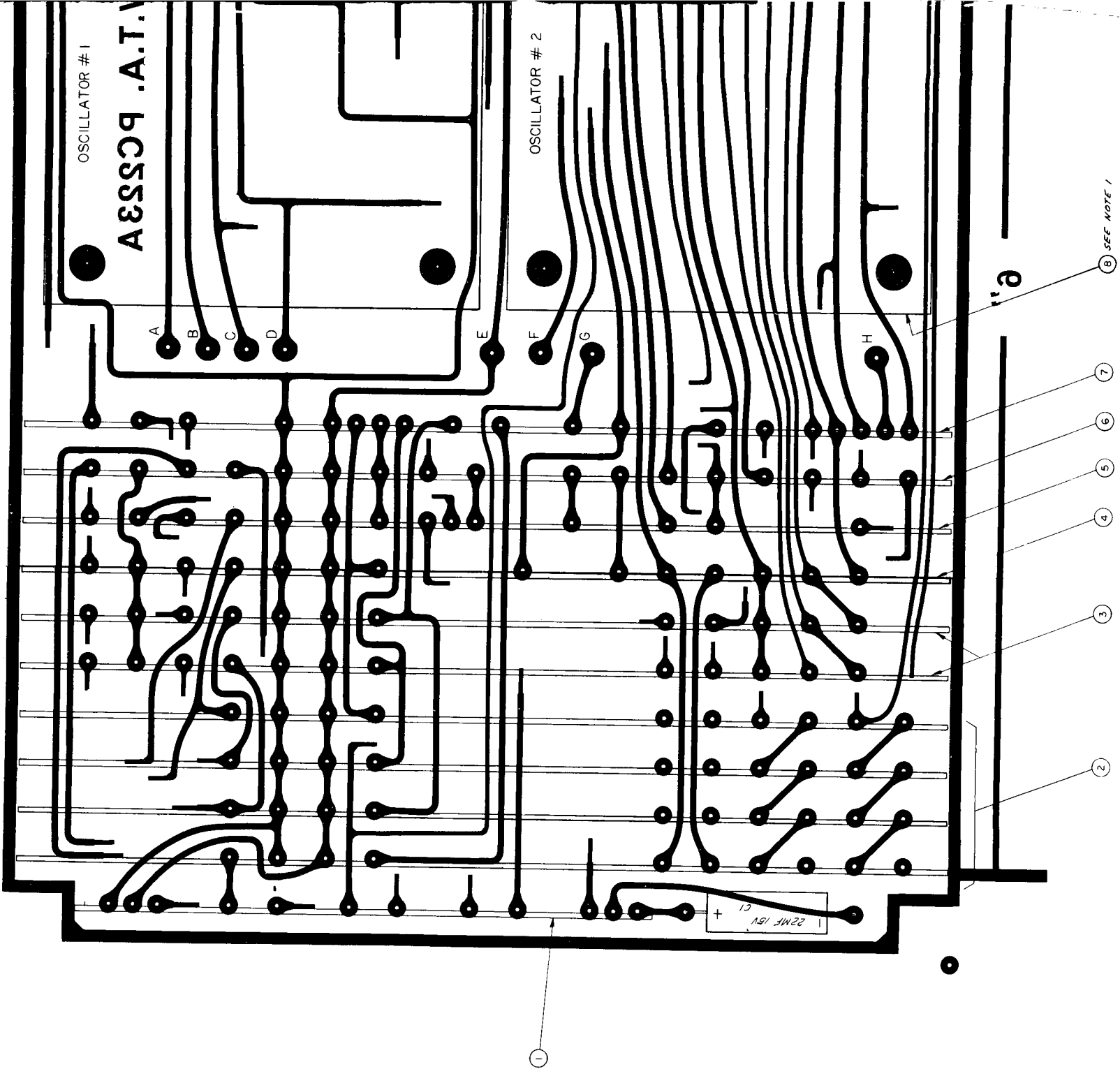
MATERIAL		NAME	QTY	DATE	REMARKS
25	2	WASHER LOCK, 5/16"			
26	2	NUT, HEX, 5/16"			
27	2	SCREW, MACH, PAN HD, 5/16"			
28	2	SCREW, MACH, PAN HD, 5/16"			
29	2	SCREW, MACH, PAN HD, 5/16"			
30	2	SCREW, MACH, PAN HD, 5/16"			
31	2	SCREW, MACH, PAN HD, 5/16"			
32	2	SCREW, MACH, PAN HD, 5/16"			
33	2	SCREW, MACH, PAN HD, 5/16"			
34	2	SCREW, MACH, PAN HD, 5/16"			
35	2	SCREW, MACH, PAN HD, 5/16"			
36	2	SCREW, MACH, PAN HD, 5/16"			
37	2	SCREW, MACH, PAN HD, 5/16"			
38	2	SCREW, MACH, PAN HD, 5/16"			
39	2	SCREW, MACH, PAN HD, 5/16"			
40	2	SCREW, MACH, PAN HD, 5/16"			
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94	2	SCREW, MACH, PAN HD, 5/16"			
95	2	SCREW, MACH, PAN HD, 5/16"			</

[illegible]

SECTION A-A

SECTION B-B

WTB 2158 F-176		NATIONAL AERONAUTICS AND SPACE ADMINISTRATION GOODARD SPACE FLIGHT CENTER GREENBELT, MARYLAND		GF-RAE-1155-513	
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WTB 2158 F-17					



OSCILLATOR # 1

A8SS33A

A B C D

OSCILLATOR # 2

E F

G

H

22MF 15V

SEE NOTE 1

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9 SEE NOTE 1

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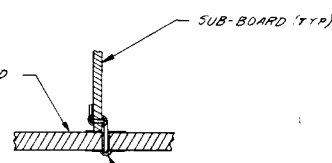
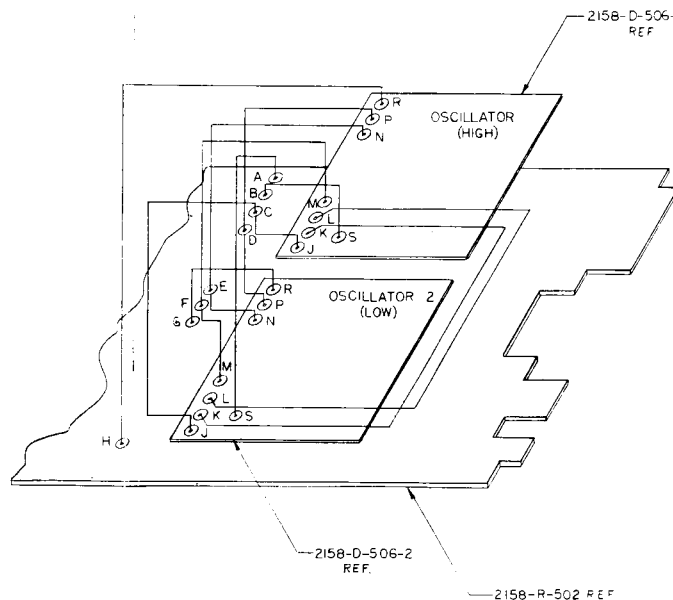
MAIN BOARD
(18) REF

REF. SCHEMATIC 2158-R-529

17

FROM POINT	WIRE DESCRIPTION	COLOR	APPROX. LENGTH	TO CONNECTOR PIN
1	MASSIVE GND	BRN	3"	J2 9-10
2	+6 VOLT	WHT/BLU	3"	J2 17-18
3	POWER GND	BLK	3"	J2 11-12
4	400 CPS CLOCK	WHT/ORN	3"	J2 23-24
5	+3.5 VOLTS	ORN	3"	J2 13-14
6	READ COMMAND	WHT/GRN	3"	J2 21-22
7	DATA OUT	WHT/YEL	3"	J2 25-26
8	SIGNAL GND	GRY	3"	J2 15-16
9	SHIFT COMMAND	WHT/BRN	3"	J2 19-20
10	5-VELD GND	WHT	3"	J2 7-8
11	TEMP MONITOR	VIO	3"	J2 3-4
12	+12 VOLTS	RED	3"	J2 1-2
13	-6 VOLTS	BLU	3"	J2 5-6

- NOTES:
1. LEAVE 8" MIN. LENGTH EXTENDING FROM OSCILLATOR TERMINALS TO INSURE SUFFICIENT WIRE FOR USE. PASS WIRES THRU GROMMETS PRESSED INTO HOLES IN CANS (2158-D-473). 1/2" TERMINALS LOCATED IN OSCILLATORS, ITEMS 8 & 9, SHOULD BE MOUNTED IN RESPECTIVE CANS AS SHOWN. SEE DRAWING 2158-F-486 FOR MECHANICAL ASSEMBLY.
 2. SOLDER PER NASA QUALITY PUBLICATION VQ-100-200-4
 3. FABRICATE IN ACCORDANCE WITH GEORGE C. HARRIS, JR., SPACE FLIGHT CENTER STANDARD ASS'Y KIT 100.



SECTION A-A

TYPICAL INSTALLATION OF SUB-BOARDS TO MAIN BOARD.

ITEM NO.	RECD	NOMENCLATURE	QTY	OR MFG PART NO.	MATERIAL	REMARKS	ZONE
18	1	PRINTED WIRING BOARD, MAIN		2158-C-502			
7	NR	WIRE, TETRAFLUOROETHYLENE COATED		#26 AWG	STRANDED	FOR INDIVIDUAL COLORED - SEE TABLE SUPPLIED BY...	
16	1	TEMP MONITOR					
15	1	CAPACITOR, 22 MFD 35V		SPRAGUE			
14	3	CAPACITOR, 22 MFD 15V		SPRAGUE			
13	1	DIODE, IN749A		T.I.			
12	2	RESISTOR 150Ω 1/4W 15%		MELWYN			
11	1	P.C. TEST LIMITER		2158-D-495			
10	1	WELDED MODULE		2158-D-527			
9	1	OSCILLATOR (HIGH)		2158-D-506-1			
8	1	OSCILLATOR (LOW)		2158-D-506-2			
7	1	P.C. ASSY. CONTROL #3		2158-D-504			
6	1	P.C. ASSY. CONTROL #2		2158-D-499			
5	1	P.C. ASSY. CONTROL #1		2158-D-493			
4	1	P.C. ASSY. N.C. COUNTER 31P		2158-D-489			
3	2	P.C. ASSY. COUNTER SHIFT R		2158-D-487			
2	4	P.C. ASSY. SHIFT REG.		2158-D-491			
1	1	P.C. ASSY. CONTROL #4		2158-D-497			
1	NO.	QTY	NAME	INIT	DATE		
LIST OF MATERIAL FOR ONE SUBJECT ITEM (WTA 2158-R-5)							
DRAWING NO'S INVOLVED:							
NEXT ASSY: 2158-F-486							
DO NOT SCALE THIS DRAWING							
TOLERANCES:							
FRACTIONS ± ANGLES ±							
3 PLACE DECIMALS ±							
2 PLACE DECIMALS ±							
SCALE: 4:1							
JOB NO:							
CLIENT:							
APPROVED							
SCALE							
UNIT BY							
CODE							
INVENT							
PAGE 1 OF 1							

RAE
PRINTED WIRING ASSY
MAIN

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